

MODEL SR400

GATED PHOTON COUNTER

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**Stanford Research Systems, Inc.
1290 D Reamwood Avenue
Sunnyvale, CA 94089 USA
Phone: (408) 744-9040**

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SAFETY AND PREPARATION FOR USE

**** CAUTION ****

This instrument may be damaged if operated with the LINE VOLTAGE SELECTOR set for the wrong ac line voltage or if the wrong fuse is installed.

LINE VOLTAGE SELECTION

The SR400 operates from a 100V, 120V, 220V, or 240V nominal ac power source having a line frequency of 50 or 60 Hz. Before connecting the power cord to a power source, verify that the LINE VOLTAGE SELECTOR card, located in the rear panel fuse holder, is set so that the correct ac input voltage value is visible.

Conversion to other ac input voltages requires a change in the fuse holder voltage card position and fuse value. Disconnect the power cord, open the fuse holder cover door and rotate the fuse-pull lever to remove the fuse. Remove the small printed circuit board and select the operating voltage by orienting the printed circuit board to position the desired voltage to be visible when pushed firmly into its slot. Rotate the fuse-pull lever back into its normal position and insert the correct fuse into the fuse holder.

LINE FUSE

Verify that the correct line fuse is installed before connecting the line cord. For 100V/120V, use a 1 Amp fuse and for 220V/240V, use a 1/2 Amp fuse.

LINE CORD

The SR400 has a detachable, three-wire power cord for connection to the power source and to a protective ground. The exposed metal parts of the instrument are connected to the outlet ground to protect against electrical shock. Always use an outlet which has a properly connected protective ground.

WARNING REGARDING USE WITH PHOTOMULTIPLIERS

The signal inputs may be damaged if a photomultiplier is used improperly with the photon counter. When left completely unterminated, a PMT will charge a cable to a few hundred Volts in a very short time. If this cable is connected to a signal input, the stored charge may damage the front-end transistors. To avoid this problem, provide a leakage path of about 100 K Ω to ground inside the base of the PMT to prevent charge accumulation.

CONNECTION TO OTHER INSTRUMENTS

All front panel BNC shields are connected to the chassis ground and to the power outlet ground via the power cord. Do not apply any voltage to either the shields or to the outputs. The outputs are not protected against connection to any potential other than ground.

SR400 SPECIFICATIONS

OVERVIEW

There are three complete counters identified as A, B, and T. All counters count up to 200 MHz.

Counter A can count INPUT 1 or the 10 MHz internal clock. Counter A is gated by AGATE, and has a count capacity of 10^9 .

Counter B can count INPUT 1 or INPUT 2. Counter B is gated by BGATE, has a count capacity of 10^9 and may be used as the preset counter to determine the count interval.

Counter T can count the internal 10 MHz clock, INPUT 2, or the gate trigger. This counter may be preset to determine the count period for up to 25 hours or 9×10^{11} counts, gates, or triggers. The timer has an accuracy of 25 ppm from 0 to 50°C.

SIGNAL INPUTS

Bandwidth:	dc to 300 MHz
Input impedance:	50 Ohms
Linear range:	± 300 mV (at input)
Input protection:	± 5 Vdc, 50 V for 1 μ s
Overload recovery:	5 ns for <10 μ s duration overload

DISCRIMINATORS

Counters A, B, and T have independent discriminators when counting the signal inputs. All discriminator levels may be set to a fixed level or scanned. A rear panel INHIBIT input (TTL active high) can inhibit the discriminators to stop the count.

Referenced to the signal inputs:

Discriminator range:	-300 mV to +300 mV.
Discriminator slope:	Rising or Falling
Resolution:	0.2 mV
Input offset voltage:	1 mV
Minimum pulse input:	10 mV
Pulse pair resolution:	5 ns.
DISC outputs:	NIM levels into 50 Ohms

TRIGGER INPUT

Impedance:	10 KOhms
Threshold:	± 2.000 Vdc in 1 mV steps
Slope:	Rising or Falling
Protection:	15 Vdc, 100 V for 1 μ s

GATE GENERATORS

There are two independent gates, AGATE and

BGATE, which enable the A and B counters. Both gates are triggered by the TRIGGER input. The gates may be fixed in time or scanned. The GATE outputs show the positions of the gates with respect to the discriminator outputs.

Insertion delay:	25 ns
Maximum delay:	999.2 ms
Minimum gate width:	5 ns
Maximum gate width:	999.2 ms or CW
Resolution:	0.1%, 1 ns minimum
Accuracy:	2 ns +1%
Jitter:	200 ps rms +100 ppm
Maximum trigger rate:	1 MHz
GATE view outputs:	NIM levels into 50 Ohms
GATE view error:	< 2 ns

COUNTING MODES

A and B for a period determined by T preSET and T input.

A \pm B for a period determined by T preSET and T input.

A for a period determined by B preSET and B input.

SCAN AND DWELL

The number of count PERIODS or data points in a scan may be set from 1 to 2000. The duration of one count period is determined by the preset condition.

The time between consecutive count periods is the DWELL TIME and can be set from 2 ms to 60 s. The DWELL output will be TTL high during the DWELL time. This output can be used to trigger external devices.

At the end of a scan (of 1 to 2000 count periods) counting may be programmed to STOP or START the scan over again.

The START key begins the first count period of the programmed scan.

The STOP key terminates the current count period and pauses the scan. If scanning, gates and disc levels are held at their last value. The STOP key pressed while in a paused condition will reset the scan and all scanned parameters will return to their start values. The START key pressed while paused resumes the scan by starting the next count period.

Specifications

The DWELL TIME may also be set to EXTERNAL. In this mode, count periods begin with the START key or EXTERNAL START INPUT (TTL rising edge). Count periods terminate with the preset condition, the STOP key, or the EXTERNAL STOP INPUT (TTL rising edge). A STOP key while not counting resets the scan. This allows completely variable dwell times and/or counting periods.

All count data is internally buffered for one scan. Data may be read over the computer interfaces during or after a scan.

DISPLAY MODE

CONTINUOUS: Displays current counter value
HOLD: Displays final count value

D/A OUTPUT

The front panel D/A Output is proportional to A, B, A-B, or A+B depending upon the counting mode selected. The D/A output is updated at the end of each count period.

There are two rear panel D/A outputs, PORT 1 and PORT 2. These outputs may be set or scanned from the front panel or via the computer interface.

Full scale: ± 10 Vdc
Resolution: 12 bits (5 mV)
Current rating: 10 mA
Output Impedance: < 1 Ohm
Accuracy: $0.1\% + 5$ mV

COMPUTER INTERFACE

IEEE-488 (GPIB) and RS232 (up to 19.2 kbaud).
Full instrument control and data transmission.

GENERAL

Dimensions: 16" x 13" x 3.5"
Weight: 10 lbs
Power: 35 Watts from 100, 120, 220, or 240 Vac
Warranty: One year parts and labor on materials and workmanship

MENU DISPLAY LIST

MODE

COUNT=A,B FOR T PRESET
A=INPUT 1
B=INPUT 2 B SET=1E3
T=10MHZ T SET=1E0 s
N PERIODS=2000 [at 1234]
AT N=STOP DWELL=1E0s
D/A=A
D/A RANGE=LOG
DISPLAY=CONTINUOUS

A GATE

AGATE=SCAN $\Delta=1.000 \mu\text{s}$
A DELAY=1.000 μs
A WIDTH=1.000 μs

B GATE

BGATE=SCAN $\Delta=1.000 \mu\text{s}$
B DELAY=1.000 μs
B WIDTH=1.000 μs

LEVEL

TRIG SLOPE=RISE
TRIG LVL=+2.000 V
A DISC SLOPE=FALL
A DISC=SCAN $\Delta=-1.0 \text{ mV}$
A DISC LVL=-10.0 mV
B DISC SLOPE=FALL
B DISC=SCAN $\Delta=-1.0 \text{ mV}$
B DISC LVL=-10.0 mV
T DISC SLOPE=FALL
T DISC=SCAN $\Delta=-1.0 \text{ mV}$
T DISC LVL=-10.0 mV
PORT1=SCAN $\Delta=+0.100 \text{ V}$
PORT1 LVL=+0.500 V
PORT2=SCAN $\Delta=+0.100 \text{ V}$
PORT2 LVL=+0.500 V

SETUP

LCD CONTRAST=12
STORE=3 EXECUTE=>
RECALL=3 EXECUTE=>

COM

GPIB ADDR=23
RS232 BAUD=9600
RS232 BITS=8
RS232 PARITY=NONE
RS232 WAIT=6
RS232 ECHO=OFF
DATA=

ABRIDGED COMMAND LIST

The parameter i is 0,1,or 2 to select counter A,B, or T

MODE

CM j Set COUNT to mode j; A,B(0), A-B(1), A+B(2) for preset T, or A FOR B preset(3).
CI i,j Set counter i to input j; 10 MHz(0), INP 1(1), INP 2(2), TRIG(3).
CP i, n Set counter i preSET to $1 \leq n \leq 9E11$.
NP m Set Number of PERIODS in a scan to $1 \leq m \leq 2000$.
NN Read current count period number or scan position.
NE j Set end of scan mode to mode j; START(1) or STOP (0).
DT x Set DWELL time to $2E-3 \leq x \leq 6E1$ s or EXTERNAL(0).
AS j Set D/A to source j; A(0), B(1), A-B(2), A+B(3).
AM j Set front panel D/A to RANGE j; LOG(0) or LINEAR (1-7).
SD j Set DISPLAY to mode j; CONTINUOUS(0) or HOLD(1).

LEVELS

TS j Set TRIG to SLOPE j; RISE(0) or FALL(1).
TL v Set TRIG LVL to $-2.000 \leq v \leq 2.000$ V.
DS i,j Set DISC i to SLOPE j; RISE(0) or FALL(1).
DM i,j Set DISC i to mode j; FIXED(0) or SCAN(1).
DY i,v Set DISC i scan step to $-0.0200 \leq v \leq 0.0200$ V.
DL i,v Set DISC i LVL to $-0.3000 \leq v \leq 0.3000$ V.
DZ i Read current DISC i LVL (during scan).
PM k, j Set PORT k (1 or 2) to mode j; FIXED(0) or SCAN(1).
PY k,v Set PORT k (1 or 2) scan step to $-0.500 \leq v \leq 0.500$ V.
PL k,v Set PORT k (1 or 2) LVL to $-10.000 \leq v \leq 10.000$ V.
PZ k Read current PORT k (1 or 2) LVL (during scan).

GATES

GM i,j Set GATE i to mode j; CW(0), FIXED(1), or SCAN(2).
GY i,t Set GATE i DELAY scan step to $0 \leq t \leq 99.92E-3$ s.
GD i,t Set GATE i DELAY to $0 \leq t \leq 999.2E-3$ s.
GZ i Read current GATE i DELAY position (during a scan).
GW i,t Set GATE i WIDTH to $0.005E-6 \leq t \leq 999.2E-3$ s.

FRONT PANEL

CS Count start, same as START key.
CH Count pause, same as STOP key while counting.
CR Count reset, same as STOP key pressed twice.
CK j Simulate key press j.
RR Rotate knob right (CW or UP) 1 step.
RL Rotate knob left (CCW or DOWN) 1 step.
SC Read cursor position. Left(0), right(1), or inactive(2).
MI j Set front panel to mode j; local(0), remote(1), locked-out(2). RS232 only.
MS string Display string on menu line.
MS Return menu line to normal display.
MD j,k Display line k of menu j.
MM Read menu number of display (j).
ML Read menu line of display (k).

Abridged Command List

STORE/RECALL

ST m Store instrument settings to location m (1 to 9).
RC m Recall instrument settings from location m (1 to 9).
RC 0 Recall default settings.

INTERFACE

CL Clear instrument.
SV m Set GPIB SRQ mask to $0 \leq m \leq 255$.
SS Read status byte.
SS j Read bit j(0-7) of status byte.
SI Read secondary status byte.
SI j Read bit j (0-2) of secondary status byte.
SW m Set RS232 character wait interval to $m \times 3.33 \text{ ms}$ $0 \leq m \leq 25$. RS232 only.
SE j,k,l,m Set RS232 terminator sequence to j,k,l,m (ASCII codes). RS232 only.
SE Clear RS232 terminator sequence to defaults. RS232 only.

DATA

QA Read last count in counter A.
QB Read last count in counter B.
QA m Read from scan buffer point m(1-2000) for counter A.
QB m Read from scan buffer point m(1-2000) for counter B.

EA Send entire counter A buffer.
EB Send entire counter B buffer.
ET Send entire counter A and B buffer.

FA Start scan and send N PERIODS data points from counter A.
FB Start scan and send N PERIODS data points from counter B.
FT Start scan and send N PERIODS data points from both counters.

XA Read current contents of counter A.
XB Read current contents of counter B.

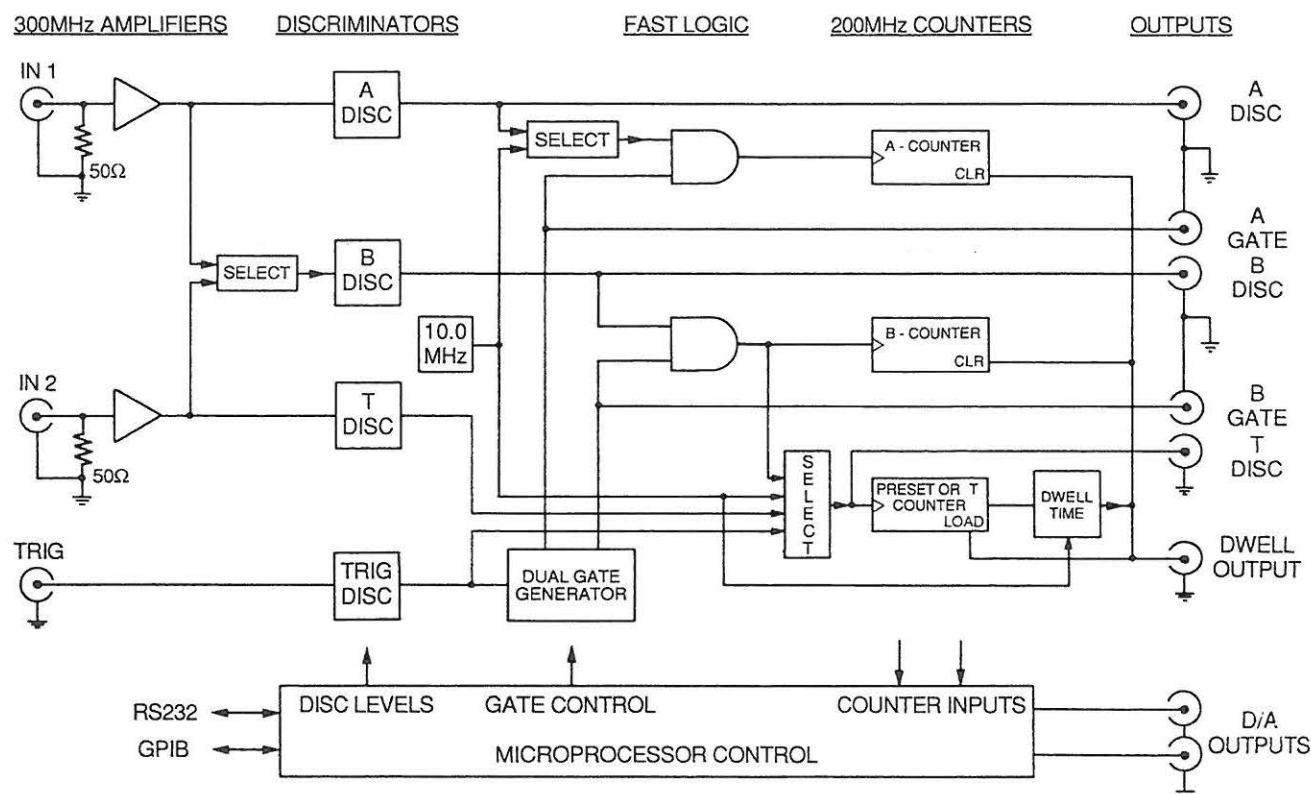
STATUS BYTE

Bit	Description
0	Parameter changed
1	Data ready
2	Scan finished
3	Counter overflow
4	Rate error
5	Recall error
6	Service request
7	Command error

SECONDARY STATUS BYTE

Bit	Description
0	Triggered
1	Inhibited
2	Counting
3-7	Unused

INSTRUMENT OVERVIEW



BLOCK DIAGRAM

SR400 OVERVIEW

The SR400 architecture is diagrammed above. There are three fast counters identified as counters A, B, and T. All three counters operate at rates up to 200 MHz. The input to each counter is selected from a number of sources, including the two analog signal inputs, the crystal timebase, and the external trigger input. Counter T is presettable and determines the measurement period. For pulsed experiments, counters A and B may also be synchronized to external events via the two independent gate generators. The gate generator provides gates from 5 ns to 1 s in duration with a delay from external trigger ranging from 25 ns to 1 s.

SIGNAL INPUTS AND DISCRIMINATORS

There are two independent analog signal inputs labelled INPUT 1 and INPUT 2. They are internally terminated into 50 Ω. The inputs can accept signals of either polarity up to ±300 mV and are pro-

ected to ±5V dc. Each input is followed by a dc to 300 MHz amplifier. This allows detection of pulses as small as 10 mV. If greater sensitivity is required, the SR440 remote amplifier is available. The SR440 can provide gains from 5 to 625 at bandwidths exceeding 200 MHz.

There are three discriminators, one for each counter. Each discriminator has selectable slope and a threshold programmable from -300 mV to +300 mV in 0.2 mV steps (referenced to the inputs). Pulse-pair resolution is 5 ns and pulses of either polarity may be detected. Each threshold may be programmed to scan in either direction with selectable step size. This can provide a pulse height analysis (PHA) output and is useful for choosing the PMT bias or discriminator levels.

COUNTING

Each counter's input may be selected from a number of sources. All multiplexing of counter inputs is

done internally to provide accurate timing and ease of operation. Input cables should rarely need to be swapped or disconnected to change measurement modes.

The actual inputs to the counters can be viewed as NIM level pulses from the DISC outputs on the front panel. The DISC pulses are negative going from 0 to -0.7 V. The falling edge is the active or counted edge regardless of the discriminator slope setting. The DISC outputs are very useful when adjusting discriminator thresholds or gate timing.

When the START key is pressed, a START command is received from the computer interface, or the EXTERNAL START input is pulsed, counter T is enabled to count. The COUNT PERIOD begins with the first pulse from T counter's input after START. The COUNT PERIOD enables counters A and B to count their respective inputs. If the A GATE and B GATE are CW or continuous, all pulses that occur during the COUNT PERIOD are counted. If the gates are enabled, only the pulses that occur during those gates which fall within the COUNT PERIOD are counted.

Counter T is presettable from 1 to 9E11. When counter T reaches the preset count, the COUNT PERIOD is terminated and counting is halted on all counters. When the input to counter T is the 10 MHz crystal timebase, the COUNT PERIOD is a fixed time interval from 100 ns to 25 hours. When the input is INPUT 2, the COUNT PERIOD is externally determined and can be as short as 5 ns and indefinitely long. This mode is used for source compensation or reciprocal measurements. If the input is TRIGGER, the COUNT PERIOD is a fixed number of gates (each trigger provides one gate). This results in a constant gate aperture time during each COUNT PERIOD. If the input is B PRESET, then counter T counts the B DISC output gated by B GATE. This is for gated ratio or reciprocal counting.

The SR400 may be programmed to cycle through 1 to 2000 COUNT PERIODS in a single scan. At the end of the programmed scan, the counters may be stopped or the scan may be restarted. If the end of scan mode is STOP, then after the last COUNT PERIOD in the scan, the counters are halted indefinitely. A counter RESET is required to re-arm the counters and a START command starts another scan. If the Number of PERIODS in the scan

is 1, then each START command results in one COUNT PERIOD. If N PERIODS is greater than 1, then consecutive COUNT PERIODS are separated by the DWELL time. The DWELL time may be set from 2 ms to 60 s. During the DWELL time, counting is disabled and data may be transferred or external parameters scanned. The DWELL OUTPUT is a TTL output which is high during the DWELL time. At the end of the DWELL time, the next COUNT PERIOD begins and the scan continues until N PERIODS have been measured. If the end of scan mode is START, one DWELL time separates the last COUNT PERIOD of the completed scan and the first COUNT PERIOD of the next scan.

Another DWELL mode is EXTERNAL. In this mode, the EXTERNAL START input enables counter T. The COUNT PERIOD begins with the first pulse to counter T after START. Counting then proceeds normally until the preset is reached. At the end of the preset COUNT PERIOD, the DWELL time is indefinite and the next COUNT PERIOD starts with the next EXTERNAL START pulse. In this way, preset COUNT PERIODS may be synchronized to external events. If externally determined COUNT PERIODS are desired, then the preset COUNT PERIOD can be set to a value much larger than actually required and the EXTERNAL STOP input used to terminate the COUNT PERIOD. The COUNT PERIOD then starts with the first pulse to counter T after EXTERNAL START and ends with EXTERNAL STOP. The next EXTERNAL START begins the next COUNT PERIOD in the scan. When the scan is finished and the scan end mode is STOP, a RESET from the front panel or computer interface is required to reset the scan. If the scan end mode is START, the next EXTERNAL START begins the first COUNT PERIOD of the next scan.

SCANNING

In all scan modes, a number of parameters may be scanned. These parameters are the three discriminator thresholds (Pulse Height Analysis), the two gate delays (Boxcar mode), and the two D/A output ports (X-axis of recorder, scope, or analog control of other apparatus).

After each COUNT PERIOD, each scanning parameter is adjusted by one step. All changes are made during the DWELL time so that all values are stable during the COUNT PERIODS. The scan limits are

determined by the start position, the step size, and the number of PERIODS in a scan. The range of a scan is N PERIODS times the step size. When the counters are reset, all scanned parameters return to their start positions.

GATING

The gate generators may be disabled by setting the gate modes to CW. In this mode, the gates are continuously open and counting is enabled solely by the COUNT PERIOD. This is appropriate for CW sources or long counting intervals. If the signal is pulsed or so weak that synchronous (lock-in) measurement is required, then the gate generators are required. The gates prevent pulses that occur outside the time interval of interest from being counted and improve signal to noise ratio. When the gates are scanned, the time behavior of the signal may be recovered.

The gates are triggered by the TRIGGER input. The TRIGGER threshold is adjustable from -2V to +2V and slope may be either RISE or FALL. The discriminated trigger pulse may be viewed through the T DISC output when the input to counter T is set to TRIG. This can be helpful when setting the trigger threshold.

The minimum insertion delay from trigger to gate is 25 ns. The additional delay may be programmed from 0 ns to 999.2 ms with a resolution of 1 part in 1000 or 1 ns, whichever is greater. The gate width is adjustable from 5 ns to 999.2 ms with the same resolution. The two gates are output as NIM pulses from the GATE outputs. The gates are open when the outputs are low. The gate outputs have a timing accuracy with respect to the discriminator outputs of 2 ns. This allows accurate timing of fast gates relative to the signal.

The two gates have independent modes, delays, and widths. This allows for a variety of measurements. For example:

1) "Boxcar" mode. The gates are used to enable the counters only when the signal is present. If the gates are scanned, then the time profile of the signal is recovered. Background subtraction can be included by using one gate for the signal and an equal width gate on the background, either before or after the signal. The signal gate may be scanned

while the background gate held fixed. Gated source compensation is also possible since counters A and B may count different sources at the same time or the same source at different times. In the compensation mode, the COUNT PERIOD is determined by the gated output of the B discriminator and the COUNT MODE is A FOR B PRESET.

2) "Lock-in" or "Chopped" mode. In this case, a cw signal is chopped on and off. This can be accomplished with a light chopper such as the SR540. The light chopper reference output is used to trigger the gates. The A GATE is positioned during the "open" cycle and counts signal plus noise. The B GATE is positioned during the "closed" cycle and only counts the noise. The counter outputs are subtracted to give only the signal. The widths of both gates should be equal and less than half of the chopper period. This prevents frequency jitter in the chopper from affecting the overlap of the gates with the open and close cycles of the chopper.

When counter T is preset and its input is set to TRIG, then each COUNT PERIOD consists of a constant number of gates. This is the usual mode of gated operation. However, a number of other modes are available. For example, if the input is set to 10 MHz, each COUNT PERIOD consists of the number of gates which occur during the COUNT PERIOD time interval. If the input is a signal input, then counting is both gated and source compensated. If the DWELL mode is EXTERNAL, an EXTERNAL START pulse starts counting and an EXTERNAL STOP pulse stops counting. The counters are enabled only during those gates which occurred between the two EXTERNAL pulses.

COUNT MODES

The length of a COUNT PERIOD is the time it takes counter T to reach its preset count. If N_t is the preset count in counter T, and R_a and R_t are the mean pulse rates of the inputs to counters A and T, then the COUNT PERIOD is

$$T = N_t / R_t$$

The number of counts in counter A will be

$$A = R_a T = R_a N_t / R_t$$

Rate:

If $R_t = 10$ MHz, then the COUNT PERIOD is a constant amount of time and

$$A = R_a (N_t / R_t) \sim R_a$$

and counter A measures the mean rate of its input.

Gated:

If R_t is the gate trigger rate and counter A is gated, then the COUNT PERIOD is N_t / R_t but the "effective" time the counter is enabled is $N_t W_a$ where W_a is the width of A gate and

$$A = R_a (N_t W_a) \sim R_a$$

In this case, counter A measures the mean rate during the A gate.

Source Compensated:

If R_t is INPUT 2, then the COUNT PERIOD is determined by the pulse rate of INPUT 2 and the T discriminator. Count A will be

$$A = N_t R_a / R_t \sim R_a / R_t$$

and is proportional to the ratio of the rates of counters A and T. If INPUT 2 monitors source intensity, then the COUNT PERIOD will adjust for fluctuations in the source. Counter A measures the ratio of its mean input rate and the mean rate of INPUT 2.

Reciprocal:

If R_t is INPUT 2, then the COUNT PERIOD is N_t / R_t . If R_a is 10 MHz, then counter A measures

$$A = R_a N_t / R_t = (R_a N_t) / R_t \sim 1 / R_t$$

and is proportional to the reciprocal of the count rate of INPUT 2. Note that each COUNT PERIOD is determined by N_t counts of INPUT 2, thus the measurement time is determined by the signal strength. Measurements of both strong and weak signals are made to the same accuracy for constant signal to noise ratio.

Counter B:

Counter B may be used in all of the above modes except Reciprocal. Since counter B is completely independent of counter A, 2 sources of counts may be measured simultaneously. These sources may be the same signal or 2 different signals; the sources may be discriminated at separate thresholds; they may be non-gated or gated; if gated, the 2 gates may be independently set. Almost any measurement can be made.

In addition, counter B has a preset mode called A FOR B PRESET. In this mode, the input to counter B is sent to counter T. The COUNT PERIOD is now determined by the B discriminator and the B gate. If R_a is INPUT 1, then counter A is source compensated by the gated input of counter B. This allows source compensation of gated signals and sources. If R_a is 10 MHz, then counter A measures the reciprocal of the rate of counter B's gated input.

OUTPUTS

The front panel can display counts up to $10^9 - 1$. Counters A and B can be displayed as separate counters or combined as A-B or A+B. When A FOR B PRESET is the count mode, then only A data is displayed.

The front panel D/A output provides an analog output proportional to A, B, A-B, or A+B depending on the counting mode. The scale may be logarithmic (1V/decade) or linear to any 3 digits.

If a chart recorder is used, the D/A output should drive the Y axis. The X axis can be recorder driven (strip chart mode) or the PORT1 or PORT2 outputs can be used to drive the X axis (scans). This latter method allows accurate determination of the X value of each point. If the DWELL output is used as the pen lift, the points will be unconnected.

If count rates are high and count periods short, then scans may be displayed on an X-Y scope. The D/A output is the Y drive and PORT1 or PORT2 is the X drive. The DWELL output should be used as the blanking pulse.

MEASUREMENT EXAMPLES

Example 1: Internally Triggered Gate Mode

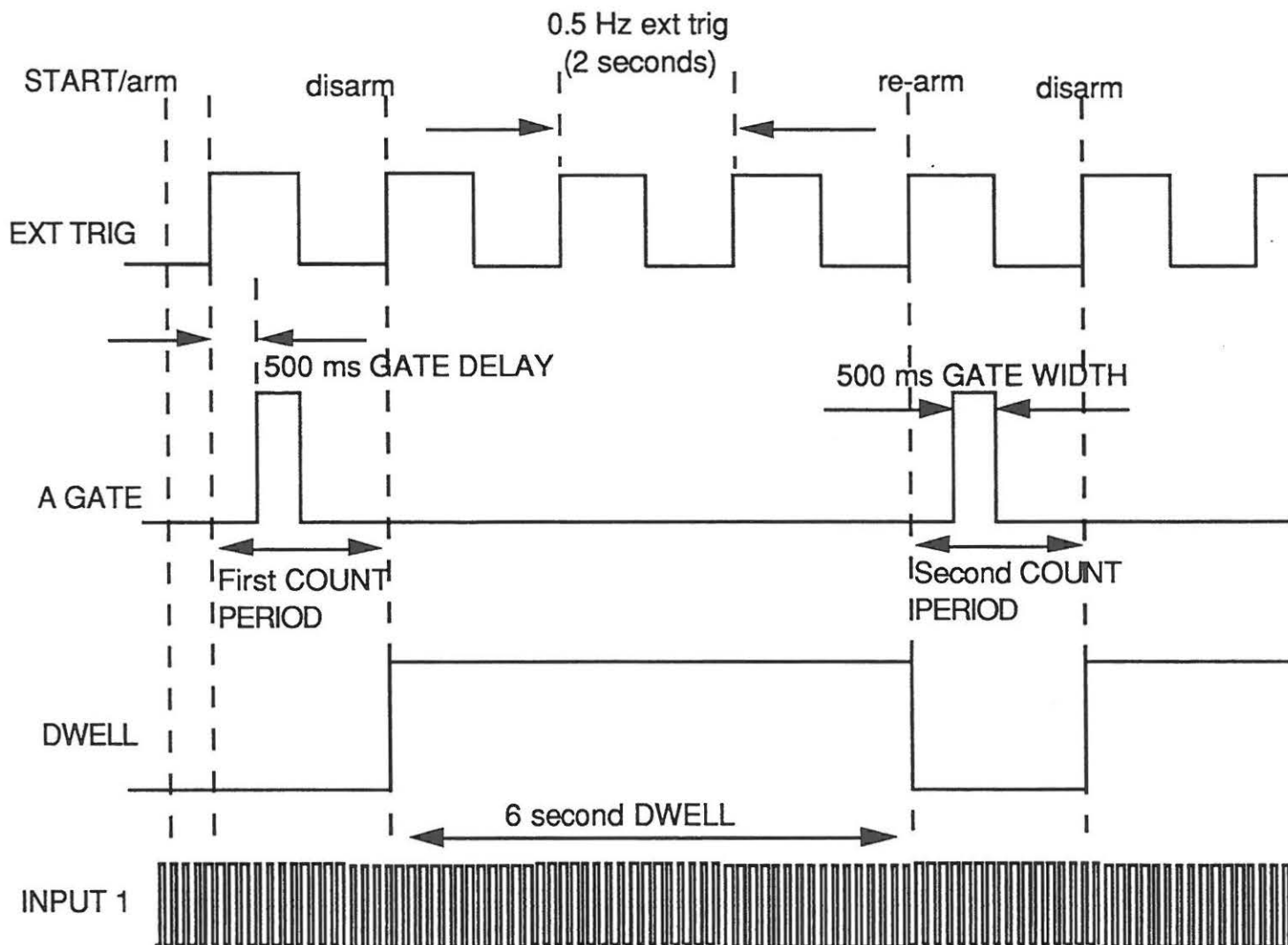
This example, described under the heading "QUICK TEST" on page 55 of this manual, is easy to perform and will quickly acquaint the user with the basic operation of the SR400.

Example 2: Externally Triggered Gate Mode

In this example, the SR400 is programmed to make 10 counting measurements of a 10 kHz pulse train. For each measurement, a gate will be delayed 500 ms from an external trigger pulse and will open for a 500 ms duration. Following the gate, there will be a 6 second dwell period* during which the counter will be disarmed. After the dwell period, the counter will reset to 0 counts, re-arm and repeat the measurement. After 10 measurement cycles, the SR400 will stop counting and disarm. You will need a 10 kHz signal source and a 0.5 Hz square wave generator for this example.

***It is important to note that the dwell period begins at the first external trigger pulse arriving after the gate is closed. Thus, even though there is only one gate opened for each measurement cycle, a minimum of two external triggers are needed: one to trigger the gate and another to begin the dwell period.**

The following timing diagram illustrates this example.



1. Turn the SR400 off. Hold down the STOP (RESET) button while turning the SR400 on. Wait three seconds and release the button.

2. Press the MODE key. Using the four arrow keys and the front panel knob, select the following parameters. Unless otherwise indicated, leave all other settings at their default values:

T = TRIG TSET = 1E0
N PERIODS = 10 [at 0]
AT N = STOP DWELL = 6E0 S

3. Press the A GATE key and select the following parameters:

A GATE = FIXED
A DELAY = 500 mS
A WIDTH = 500 mS

4. Press the LEVEL key and select the following parameters:

A DISC SLOPE = RISE
A DISC LVL = +150 mV

5. Using a function generator or a pulse generator, apply a 0.5 Hz, 5 V (TTL) signal to the TRIGGER INPUT on the SR400.

6. Using a second signal source, apply a 10 kHz square wave (or pulse train) with an amplitude of approximately 500 mV to the SIGNAL INPUT 1 on the SR400.

7. To view the COUNT PERIODS, press the MODE key and use the up/down arrow keys to display "N PERIODS = 10 [at 0] "

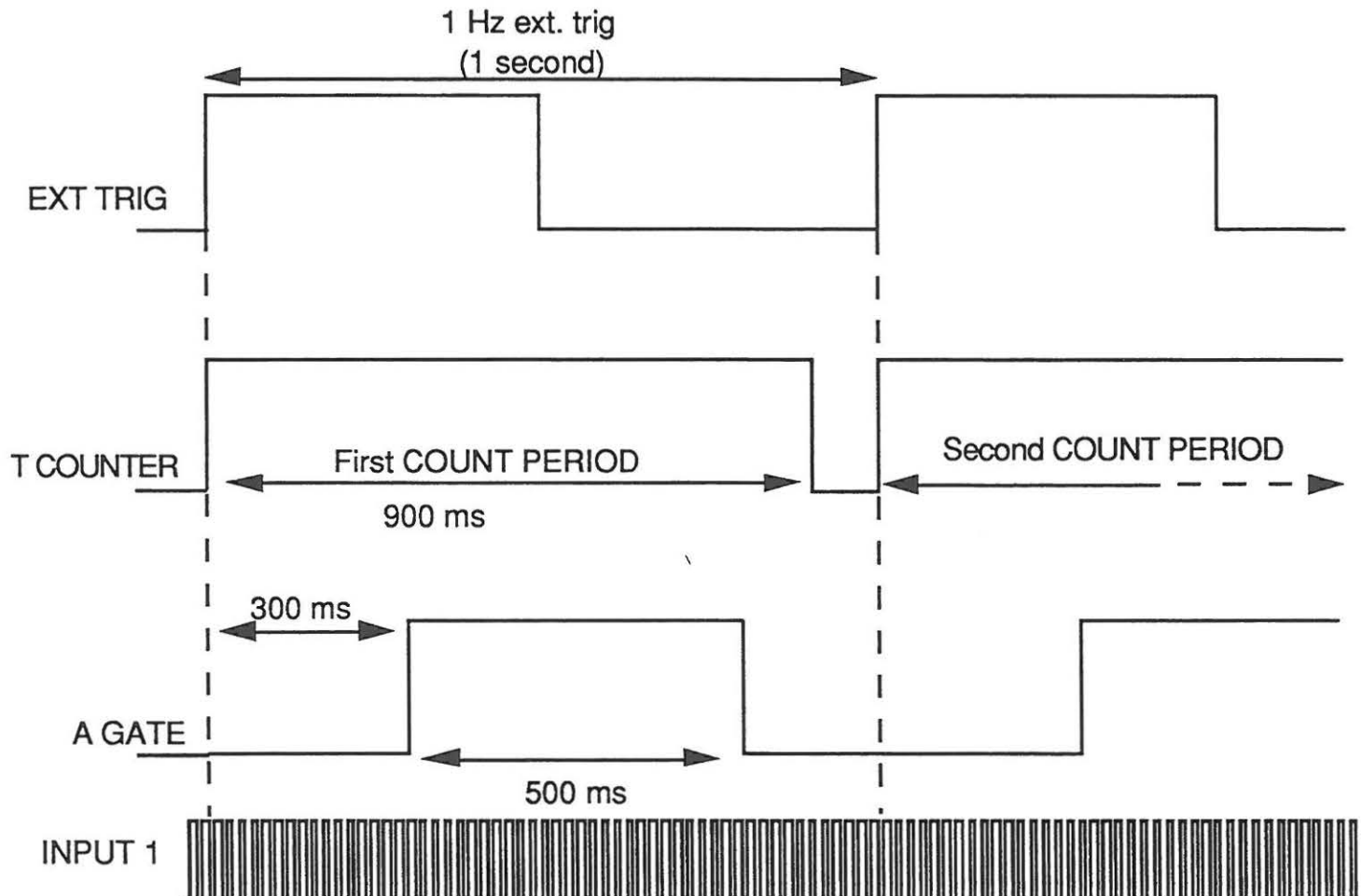
Arm the counter by pressing the START button. (Alternatively, a TTL pulse applied to the START BNC on the front panel will arm the counter.) Each measurement cycle should result in approximately 5,000 counts.

As an extension of the above example, the SR400 can be programmed to accumulate counts over numerous gates. This is often done to improve the signal to noise ratio of low light level experiments. To accumulate counts, set TSET (MODE menu) to the desired number of gates. For example, if you wish to accumulate over five gates, set TSET = 5E0. In this case, six external triggers are required for each measurement cycle: five to trigger the five gates and one to start the DWELL PERIOD. Any additional triggers received during the DWELL PERIOD are ignored.

Example 3: Externally Triggered Gate Mode

This example is a variation of Example 2, but requires no additional trigger pulses to begin the DWELL PERIOD. In this example, the SR400 is programmed to make 10 counting measurements of a 10 kHz pulse train. The counting gates, each 500 ms wide, are delayed 300 ms from external triggers arriving at a 1 Hz rate. In this case, the SR400 requires only one trigger per COUNT PERIOD, however the trigger configuration is a bit trickier than the previous example.

The following timing diagram illustrates this example.



To avoid the triggered dwell period, the T counter is logically ANDed with the A gate. This is accomplished by letting the T counter count the SR400's internal 10 MHz oscillator signal for a period extending out beyond the A gate, but not exceeding the trigger repetition period (1 second in this case). Mathematically, this is expressed:

$$A \text{ DELAY} + A \text{ WIDTH} < TSET < (1 / \text{Trigger Rate})$$

Furthermore, DWELL must be set to EXTERNAL in the MODE menu. This enables the SR400 to reset and begin each COUNT PERIOD upon receiving a TTL pulse at the START BNC input.

1. Turn the SR400 off. Hold down the STOP (RESET) button while turning the SR400 on. Wait three seconds and release the button.

2. Press the MODE key. Using the four arrow keys and the front panel knob, select the following parameters. Unless otherwise indicated, leave all other settings at their default values:

T = 10 MHz TSET = 9E-1 s

N PERIODS = 10 [at 0]

AT N = STOP DWELL = EXTERNAL (Turn the knob counterclockwise until "DWELL = EXTERNAL" is displayed.)

3. Press the A GATE key and select the following parameters:

A GATE = FIXED

A DELAY = 300 mS

A WIDTH = 500 mS

4. Press the LEVEL key and select the following parameters:

A DISC SLOPE = RISE

A DISC LVL = +150 mV

5. Using a function generator or a pulse generator, apply a 1 Hz, 5 V (TTL) signal to the EXTERNAL START input on the SR400. Use a BNC 'T' connector at the EXTERNAL START input and extend a cable from the EXTERNAL START input to the TRIG input.

6. Using a second signal source, apply a 10 kHz square wave (or pulse train) with amplitude of approximately 500 mV to SIGNAL INPUT 1 on the SR400.

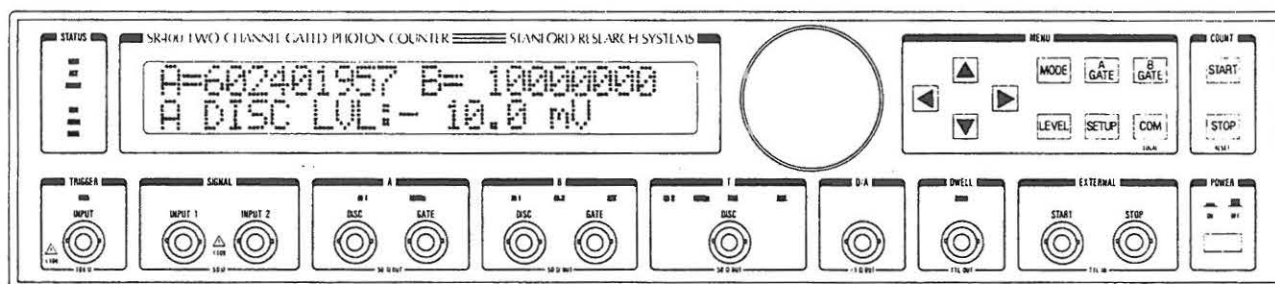
7. To view the COUNT PERIODS, press the MODE key and use the up/down arrow keys to display "N PERIODS = 10 [at 0] "

8. Arm the SR400 by pressing the STOP (RESET) button.

Each counting period should result in approximately 5,000 counts. After 10 counting measurements the SR400 will disarm. To re-arm, press the STOP (RESET) button.

GUIDE TO OPERATION

FRONT PANEL SUMMARY



POWER BUTTON

The SR400 is turned on by depressing the POWER button. All instrument settings are stored in nonvolatile memory and are retained when the power is turned off. The model, firmware version, and serial numbers of the unit will be displayed briefly when the power is turned on. The recall status is also displayed. If recall is OK, then the instrument returned to the settings in effect when the power was last turned off. If recall is ERR, then the stored settings were lost and the default settings are used. If recall is DEFAULT, then the instrument is resetting to default conditions. This is done if the STOP (RESET) key is held down when the power is turned on. The default setup is listed in the next chapter.

LIQUID CRYSTAL DISPLAY

The 2 line by 24 character LCD is the user interface for data display and front panel programming operations. The top line always shows count data. The lower line is the menu line and displays one line of setup information. The LCD is backlit by an electroluminescent panel for convenient operation in low light level conditions. When in the SETUP menu, the contrast of the LCD may be adjusted for optimum viewing.

MENU KEYS

All operating parameters of the SR400 are grouped into six function menus. The six menu keys select the group of functions to be programmed. The four arrow keys select individual parameters within each menu. The selected parameter is displayed on the menu line of the

display.

The MODE menu selects the counting mode, counter inputs, counter presets, scan mode and length, dwell time, analog output, and display mode. The LEVEL menu adjusts the discriminator, trigger, and D/A levels. The A GATE and B GATE menus program the gate generators. The SETUP menu provides STORE and RECALL of 9 different instrument setups and adjustment of the LCD contrast. The COM menu sets up the computer interface parameters and displays received data. Detailed descriptions of the menus are provided at the end of this section.

SPIN KNOB

All parameters are altered using the spin knob. Whenever the menu line has an active or flashing cursor, the displayed parameter can be adjusted with the knob.

START, STOP, AND RESET

The two count keys, START and STOP, are used to start, pause, and reset the counters.

A scan can be started only if the counters are in the reset condition; all scanned parameters are at their start values and the counters are reset to zero. In general, pressing the STOP key twice in succession puts the counters into the reset condition.

START begins the first COUNT PERIOD of the programmed scan. If the DWELL time is programmed, then the scan will continue automatically. A STOP will pause the scan; the current COUNT PERIOD is terminated and ignored. A START while

paused will resume the scan. A STOP while paused resets the scan. If the DWELL time is EXTERNAL, then a STOP during the COUNT PERIOD terminates the COUNT PERIOD and advances the scan. START begins the next COUNT PERIOD in the scan. A START during a COUNT PERIOD or programmed DWELL period is ignored.

A STOP while in a dwell period or pause resets the scan; all scanned parameters return to their start values and the counters are reset. In general, 2 consecutive presses of the STOP key will reset the counters.

The SCAN led is on when a scan is in progress. The SCAN led is off when the scan is paused or the counters are reset.

INSTRUMENT STATUS

Six LED's indicate the instrument status. REM is on when the front panel is locked out, ACT flashes when there is activity on the computer interfaces, and ERR flashes when there is a computer interface error. COUNT is on during each COUNT PERIOD, OVRN indicates that a counter has exceeded $10^9 - 1$ counts during a COUNT PERIOD, and RATE flashes whenever a gate delay or width exceeds the trigger period minus 1 μ s. The RATE led also flashes if the trigger rate exceeds 1 MHz, even if the gates are in the CW mode.

COUNTER LED'S

The source LED's in the counter sections indicate what each counter's input is. The counter inputs are output through the DISC outputs. The SET LED's indicate which counter, B or T, is the preset counter. The B SET LED is on when the mode is A FOR B PRESET, the T SET LED is on otherwise.

INPUTS

The rising or falling edge of the TRIGGER input triggers the gate generators. The input impedance is 10 K Ω and the minimum pulse width is 10 ns. The trigger level is adjustable from -2V to +2V with either slope. The minimum pulse amplitude is .1 V.

The SIGNAL INPUTS, 1 and 2, are 50 Ω analog inputs. Photomultiplier or preamplifier outputs are connected to these inputs. These inputs are linear between -300 mV to +300 mV and protected to \pm

5 Vdc.

EXTERNAL START is a TTL input with the rising edge active. The minimum pulse width is 20 ns. When the DWELL time is EXTERNAL, the COUNT PERIOD begins with the rising edge of START and ends with the preset condition or the rising edge of STOP. When the DWELL time is programmed, EXT START begins the scan.

EXTERNAL STOP is a TTL input with the rising edge active. The minimum pulse width is 20 ns. When the DWELL time is EXTERNAL, EXT STOP ends each count period. Reset must be accomplished via the STOP key or computer interface. When the DWELL time is programmed, EXT STOP resets the scan.

OUTPUTS

The DISC outputs are 50 Ω , NIM level, outputs showing the actual inputs to the three counters. These outputs are active low (-.7 V) and counting occurs on the falling edge. The outputs are low when the signal inputs exceed their threshold levels with the correct slope.

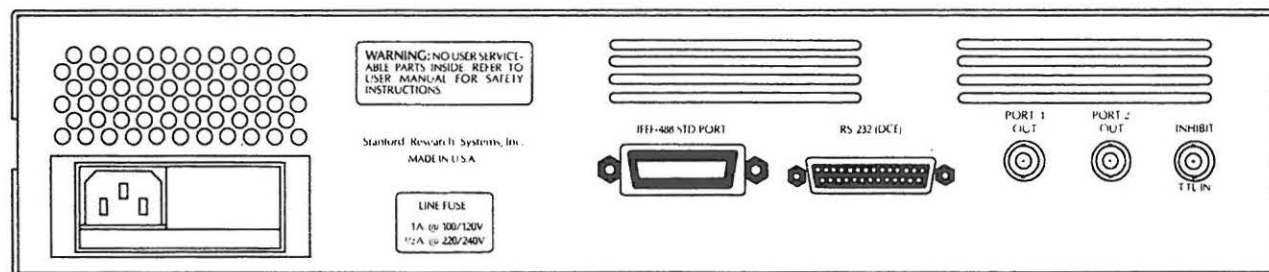
Cables attached to the DISC outputs should be kept far from the cables which carry the SIGNAL inputs. The 0.7 V DISC pulses can be picked up by the SIGNAL cables and discriminated. This causes the discriminators to oscillate. This is not a problem if the threshold is greater than 10 mV. If smaller thresholds are required, careful arrangement of the cables is necessary, or the DISC output should just be terminated at the front panel.

The GATE outputs are 50 Ω , NIM level, outputs showing the actual gates to counters A and B. The GATE outputs are low (-.7 V) during the gate. The timing accuracy between the GATE outputs and DISC outputs is 2 ns.

The D/A analog output provides a -10V to +10V output proportional to the count data with linear or logarithmic scaling. Output impedance is < 1 Ω and the maximum current output is 10 mA.

The DWELL output is a TTL level output which is high whenever the counters are paused. The rising edge may be used to trigger changes in external parameters or data acquisition.

REAR PANEL SUMMARY



POWER ENTRY MODULE

The power entry module is used to fuse the line, select the line voltage, and block high frequency noise from entering or exiting the instrument. Refer to the first page of this manual for instructions on selecting the correct line voltage and fuse.

IEEE-488 STD PORT

The 24 pin IEEE-488 connector allows a computer to control the SR400 via the IEEE-488 or GPIB instrument bus. The address of the instrument is set from the front panel using the COM menu.

RS232 CONNECTOR

The RS232 interface connector is configured as a

DCE. The baud rate, parity, character length, and echo mode are programmed from the front panel using the COM menu.

INPUT

INHIBIT is a TTL input which disables the A,B, and Tdiscriminators when it is high. The minimum pulse width is 10 ns.

OUTPUT

The PORT1 and PORT2 D/A analog outputs provide -10 V to +10V, user programmable outputs. The outputs may be scanned to drive the X axis of a recorder or scope during scans. Output impedance is $< 1 \Omega$ and the maximum current output is 10 mA.

FRONT PANEL MENUS

The lower line of the LCD display is the menu line. Menu lines are grouped into six menus. To select a new group or menu, press a new menu key. The menu line that will be displayed is the most recently active menu line of the new menu.

Each menu contains several lines. Each line may consist of 1 or 2 displayed parameters. The DOWN and UP keys change which menu line within a menu will be displayed. The DOWN key displays the next menu line and the UP key, the previous line. Pressing either key many times in succession will scroll through all of the menu lines within a menu.

If the menu line contains two parameters, the LEFT and RIGHT keys select between the two. On some lines, the right hand parameter is active only for certain values of other parameters. On these lines, the right hand parameter may be inactive.

The spin knob is used to modify the active parameter. A flashing cursor precedes the value which the knob will alter.

The following section details each menu line. The lines are listed in the order in which the DOWN key scrolls through each menu. For parameters with limited choices, all selections are shown. For numerical parameters, example values are shown.

MODE MENU

COUNT= A,B FOR T PRESET
A-B FOR T PRESET
A+B FOR T PRESET
A FOR B PRESET

This line selects the data display and the preset counter. The upper line of the LCD is the count display. There are four displays available; A and B separately, A-B, A+B, or only A, depending upon the COUNT selection. All count data is right justified with leading zeroes off. The preset counter is either T or B. The appropriate SET LED will be on indicating which counter is preset. Note that the choice of inputs to the counters determines whether the count data measures rate, ratio, or reciprocal.

Changing this line causes the counters to reset.

A= INPUT 1
10 MHz

The input to counter A can be either the internal 10 MHz timebase or the signal at INPUT 1 discriminated by the A discriminator. The selected counter input can be viewed at the A DISC output. The 10 MHz source is used for reciprocal measurements or for quick testing of the counter operation. The appropriate A counter source LED is also lit.

Changing this line during a COUNT PERIOD may result in a bad data point.

B= INPUT 2 BSET= ****
INPUT 1 1E0
9E11

The input to counter B can be either the signal at INPUT 1 or INPUT 2 discriminated by the B discriminator. The selected counter input can be viewed at the B DISC output. The appropriate B counter source LED is lit. Changing the counter input during a COUNT PERIOD may result in a bad data point.

If counter T is preset, the B SET parameter is inactive (indicated by the **** on the display). If counter B is preset (ratio or reciprocal measurement), B SET is the B preset count from 1E0 to 9E11. Changing B SET during a scan causes the counters to pause; any COUNT PERIOD in progress is terminated and ignored. A START begins the next COUNT PERIOD with the new preset value.

T= 10 MHz TSET= 1E0 s
INPUT 2 1E7
TRIG 1E3

The input to counter T can be the internal 10 MHz timebase (rate measurement), the signal at INPUT 2 discriminated by T discriminator (ratio or reciprocal measurement), or the TRIGGER input discriminated by the trigger discriminator (constant gate aperture). The selected counter input can be viewed at the T DISC output. The appropriate T counter source LED is lit.

If counter B is preset, this line is inactive (indicated by the **** on the display). If counter T is preset, both parameters on this line are active. T SET is the preset value of counter T from 1E0 to 9E11. If

the T input is 10 MHz, then T SET is displayed as real time = T SET cycles of 10 MHz. Changing T SET during a scan causes the counters to pause; any COUNT PERIOD in progress is terminated and ignored. A START begins the next COUNT PERIOD with the new preset value.

N PERIODS=2000 [at 1234]

This line programs the number of COUNT PERIODS in one scan from 1 to 2000. N PERIODS determines the length of the internal data buffer and the range of any scanned parameters. The quantity in brackets is the number of completed COUNT PERIODS in the current scan.

Changing N PERIODS during a scan changes the length of the current scan. Reducing N PERIODS past the current scan position will cause an end of scan condition.

AT N= STOP DWELL= 1E0 s
START 2E-3 s
EXTERNAL

This line sets the end of scan mode and the DWELL time. When a scan reaches N PERIODS, the counters may be paused or restarted. AT N=STOP causes an end of scan pause. All scanned parameters are held at their final values. A counter reset is required before another scan can begin. If N PERIODS is 1, then each START results in one COUNT PERIOD. If N PERIODS is greater than 1, then consecutive COUNT PERIODS are separated by the DWELL time. During the DWELL time, counting is disabled and scanned parameters updated. The DWELL OUTPUT is TTL high during the DWELL time.

If AT N=START, the counters will wait one DWELL time at the end of each scan, during which all scanned parameters return to their start positions, and then another scan begins. This is used for continuous data taking with the measurement rate determined by the COUNT PERIOD plus the DWELL TIME.

The DWELL time may be programmed from 2E-3 s to 6E1 s (1 minute) or set to EXTERNAL. The internal DWELL time is based on the 10 MHz timebase and may have 200 ns of indeterminacy. Changing the DWELL time during a scan causes the counters to pause; any COUNT PERIOD in progress is

terminated and ignored.

Setting the DWELL time below 2E-3 s results in an EXTERNAL dwell. In this mode, the START key, EXT START, or remote start command starts the COUNT PERIOD. The COUNT PERIOD lasts until the preset condition is reached. The following DWELL time is indefinite; the next COUNT PERIOD begins with another START. A STOP key, EXT STOP, or remote stop command during the COUNT PERIOD terminates the COUNT PERIOD. Thus by programming the preset COUNT PERIOD much longer than required, the COUNT PERIOD may be determined by START and STOP. Scanning proceeds as normal.

NOTE: If AT N=STOP, the counters will pause at the end of a scan. A reset (STOP key, EXT STOP, or remote reset command) is required to rearm the counters before another scan can begin. In general, to reset the counters at any time, the STOP key is pressed twice in succession.

D/A= A
B
A+B
A-B

The D/A output is proportional to the count data on the top line of the LCD. When the COUNT is A-B, A+B, or A only, the D/A output can only follow the data display. In these modes, the cursor is inactive and the knob does nothing. If the COUNT is A,B FOR T PRESET, then the D/A output can follow either A or B.

D/A RANGE= LOG
##765####

The D/A output scale can be either logarithmic or linear. If the output is logarithmic, the scale is 1V per decade with 1 count being 0V, 0 counts being -1V, and negative counts being -2V. If the scale is linear, the output is proportional to 3 digits of the count. These digits are shown while the ignored digits are represented by #s.

The D/A output is updated at the end of each COUNT PERIOD. If the D/A output is driving a chart recorder, the end of scan mode should be STOP or the pen reset will mar the graph. If the DWELL output is connected to the recorder pen lift, data

will be plotted as unconnected points, otherwise, the discrete points will be connected.

**DISPLAY= CONTINUOUS
HOLD**

The count data on the upper line of the LCD can be updated in two ways. CONTINUOUS mode updates the display 20 times a second during the COUNT PERIOD. Data is held during the DWELL time at the final value reached during the previous COUNT PERIOD. HOLD updates the display only at the end of each COUNT PERIOD.

CONTINUOUS is useful when COUNT PERIODS are long, HOLD is better when COUNT PERIODS are short.

LEVEL MENU

**TRIG SLOPE= RISE
FALL**

The gate generator may be triggered on either the rising or falling edge of the TRIGGER input.

TRIG LVL=+2.000 V

The gate TRIGGER threshold level may be adjusted from -2.000 V to +2.000V. The discriminated trigger pulses may be viewed through the T DISC output when counter T is preset and its input is TRIG. This can be helpful in selecting a trigger level. The minimum pulse amplitude required to trigger is 100 mV.

**A DISC SLOPE= RISE
FALL**

When A counter's input is INPUT 1, A discriminator may be triggered by rising or falling edges of the INPUT 1 signal. If the signal pulses are negative going, then the slope should be FALL. If the signal pulses are positive, then the slope should be RISE. In either case, the A DISC output is always active low; i.e., a discriminated rising edge of the signal will be output as a negative going NIM pulse at the A DISC output. The falling edges of the A DISC output are counted by counter A.

**A DISC= FIXED
SCAN Δ= -1.0 mV**

The threshold level of A discriminator may be FIXED or SCANNed. If the level is fixed, the right hand parameter is not displayed. If the level is scanned, then the right hand parameter is the step size. At the end of each count period in a scan, the level is incremented by one step size. The step size can be varied from -20.0 mV to + 20.0 mV. The range of a scan is N PERIODS times the step size. The starting level is set in the next menu line.

The scan mode may be changed during a scan. If it is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start level immediately.

The step size may also be changed during a scan. The new step size will be in effect at the end of the next count period. The scan is not reset or paused.

**A DISC LVL= -10.0 mV
: -11.0 mV**

The threshold level of A discriminator may be adjusted from -300 mV to + 300 mV. This level is the fixed threshold when A DISC is FIXED and the starting level when A DISC is SCANNed.

When the A DISC mode is SCAN and a scan is in progress, this line displays the current position of the A DISC level. The cursor is inactive and replaced by a ':'. The spin knob cannot adjust the threshold while it is scanning. When the mode is SCAN, the starting level may only be adjusted when the counters are in reset.

**B DISC SLOPE= RISE
FALL**

B discriminator may be triggered by rising or falling edges of the INPUT 1 or INPUT 2 signal. If the signal pulses are negative going, then the slope should be FALL. If the signal pulses are positive, then the slope should be RISE. In either case, the B DISC output is always active low; i.e., a discriminated rising edge of the signal will be output as a negative going NIM pulse at the B DISC output. The falling edges of the B DISC output are counted by counter B.

B DISC= FIXED SCAN Δ = -1.0 mV

The threshold level of B discriminator may be FIXED or SCANNed. If the level is FIXED, the right hand parameter is not displayed. If the level is scanned, then the right hand parameter is the step size. At the end of each count period in a scan, the level is incremented by one step size. The step size can be varied from -20.0 mV to + 20.0 mV. The range of a scan is N PERIODS times the step size. The starting level is set in the next menu line.

The scan mode may be changed during a scan. If it is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start level immediately.

The step size may also be changed during a scan. The new step size will be in effect at the end of the next count period. The scan is not reset or paused.

B DISC LVL = -10.0 mV : -11.0 mV

The threshold level of B discriminator may be adjusted from -300 mV to + 300 mV. This level is the fixed threshold when B DISC is FIXED and the starting level when B DISC is SCANNed.

When the B DISC mode is SCAN and a scan is in progress, this line displays the current position of the B DISC level. The cursor is inactive and replaced by a ':'. The spin knob cannot adjust the threshold while it is scanning. When the mode is SCAN, the starting level may only be adjusted when the counters are in reset.

T DISC SLOPE= RISE FALL

When T counter's input is INPUT 2, T discriminator may be triggered by rising or falling edges of the INPUT 2 signal. If the signal pulses are negative going, then the slope should be FALL. If the signal pulses are positive, then the slope should be RISE. In either case, the T DISC output is always active low; i.e., a discriminated rising edge of the signal will be output as a negative going NIM pulse at the T DISC output. The falling edges of the T DISC output are counted by counter T.

T DISC= FIXED SCAN Δ = -1.0 mV

The threshold level of T discriminator may be FIXED or SCANNed. If the level is FIXED the right hand parameter is not displayed. If the level is scanned, then the right hand parameter is the step size. At the end of each count period in a scan, the level is incremented by one step size. The step size can be varied from -20.0 mV to + 20.0 mV. The range of a scan is N PERIODS times the step size. The starting level is set in the next menu line.

The scan mode may be changed during a scan. If it is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start level immediately.

The step size may also be changed during a scan. The new step size will be in effect at the end of the next count period. The scan is not reset or paused.

T DISC LVL = -10.0 mV : -11.0 mV

The threshold level of T discriminator may be adjusted from -300 mV to + 300 mV. This level is the fixed threshold when T DISC is FIXED and the starting level when T DISC is SCANNed.

When the T DISC mode is SCAN and a scan is in progress, this line displays the current position of the T DISC level. The cursor is inactive and replaced by a ':'. The spin knob cannot adjust the threshold while it is scanning. When the mode is SCAN, the starting level may only be adjusted when the counters are in reset.

PORT1= FIXED SCAN Δ =+0.100 V

The level of the rear panel PORT1 output may be FIXED or SCANNed. If the level is scanned, then the right hand parameter is the step size. At the end of each count period in a scan, the level is incremented by one step size. The step size can be varied from -0.500 V to +0.500 V. The range of a scan is N PERIODS times the step size. The starting level is set in the next menu line.

The scan mode may be changed during a scan. If it

is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start level immediately.

The step size may also be changed during a scan. The new step size will be in effect at the end of the next count period. The scan is not reset or paused.

PORT1 = +0.500 V
: +0.600 V

The level of the rear panel PORT1 output may be adjusted from -10 V to +10 V. This level is the fixed output when PORT1 is FIXED and the starting level when PORT1 is SCANNed.

When the PORT1 mode is SCAN and a scan is in progress, this line displays the current position of the PORT1 level. The cursor is inactive and replaced by a ':'. The spin knob cannot adjust the level while it is scanning. When the mode is SCAN, the starting level may only be adjusted when the counters are in reset.

PORT2= FIXED
SCAN Δ=+0.100 V

The level of the rear panel PORT2 output may be FIXED or SCANNed. If the level is scanned, then the right hand parameter is the step size. At the end of each count period in a scan, the level is incremented by one step size. The step size can be varied from -0.500 V to +0.500 V. The range of a scan is N PERIODS times the step size. The starting level is set in the next menu line.

The scan mode may be changed during a scan. If it is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start level immediately.

The step size may also be changed during a scan. The new step size will be in effect at the end of the next count period. The scan is not reset or paused.

PORT2 = +0.500 V
: +0.600 V

The level of the rear panel PORT2 output may be adjusted from -10 V to +10 V. This level is the fixed output when PORT2 is FIXED and the starting level when PORT2 is SCANNed.

When the PORT2 mode is SCAN and a scan is in progress, this line displays the current position of the PORT2 level. The cursor is inactive and replaced by a ':'. The spin knob cannot adjust the level while it is scanning. When the mode is SCAN, the starting level may only be adjusted when the counters are in reset.

GATE MENU

The A GATE and B GATE are identical in operation. The discussion below describes the A GATE menu. The B GATE menu is similar, the only difference being that A GATE enables A counter and B GATE enables B counter.

In order to provide 1 part in 1000 settability, gate delay, width, and scan step times have variable resolution depending upon the 4 most significant digits of the time.

<u>4 most significant digits</u>	<u>resolution in 4th digit</u>
1000 to 2048	1 (1001, 1002, 1003, ...)
2048 to 4096	2 (2048, 2050, 2052, ...)
4096 to 8192	4 (4096, 4100, 4104, ...)
8192 to 9992	8 (8192, 8200, 8208, ...)

Below 1.000 μs, the resolution is 1 ns.

For example, near 10.00 μs, the allowed times are 9.984 μs, 9.992 μs, 10.00 μs, 10.01 μs, etc. In each case, the resolution is 1 part in 1000 or better.

Note that gates may be missed or truncated while gate delays or widths are changed. Changing gate parameters while counting may therefore cause bad data to be collected. While scanning, all changes to the gate delay are made during the dwell time and do not affect the count data. The GATE outputs may show missed gates during the dwell time.

A GATE=CW
FIXED
SCAN $\Delta=1.000 \mu\text{s}$

The A gate enables the A counter in three modes. CW is continuous gating; counter A is enabled during the entire count period. FIXED delay and width enables the counter only during those gates which occur during the count period. SCANNing increments the gate delay between count periods during a scan.

The gate may be viewed through the A GATE output. The counter is enabled when the output is NIM low. If the mode is CW, the output is always low, otherwise, it is low only during the gate.

If the gate delay is scanned, then the right hand parameter is the step size. At the end of each count period in a scan, the delay is incremented by one step size. The step size can be varied from 0 to 99.92 ms. The range of a scan is N PERIODS times the step size. The starting value of the delay is set in the next menu line.

The scan mode may be changed during a scan. If it is changed to SCAN, the delay begins scanning from the start position on the next count period. If it is changed to FIXED, the delay returns to the start position immediately.

The step size may also be changed during a scan. The new step size will be in effect at the end of the next count period. The scan is not reset or paused.

A DELAY=****
=1.000 μs
: 2.000 μs

This line is inactive if the A GATE is CW. Even when this is the case, the RATE LED will flash if the trigger frequency exceeds 1 MHz.

If the A GATE mode is FIXED or SCAN, then the delay of A GATE from trigger may be adjusted from 0 to 999.2 ms. This is in addition to the fixed 25 ns insertion delay. This delay is the fixed delay when A GATE is FIXED and the starting delay when A GATE is SCANNed.

The delay can be as long as the time between triggers minus 1 μs . This is independent of the gate width. Delays longer than the trigger period minus

1 μs will result in missed gates and the RATE led will flash.

When the A GATE mode is SCAN and a scan is in progress, this line displays the current position of the A GATE delay. The cursor is inactive and replaced by a ':'. The spin knob cannot adjust the delay while it is scanning. When the mode is SCAN, the starting delay may only be adjusted when the counters are in reset.

Because of the variable resolution of the delay, the scan step may be less than the resolution of the current delay. In this case, the delay is incremented every n count periods where n is the number of scan steps needed to exceed the resolution. In all cases, the delay only steps from one allowed value to another and is as close to the desired value as will be allowed by the resolution.

A WIDTH= ****
1.000 μs

This line is inactive if A GATE is CW. Even when this is the case, the RATE LED will flash if the trigger frequency exceeds 1 MHz.

If the A GATE mode is FIXED or SCAN, then the width of A GATE may be adjusted from 0.005 μs to 999.2 ms. The width can be as long as the time between triggers minus 1 μs . This is independent of the gate delay. Widths longer than the trigger period minus 1 μs will result in missed gates and the RATE led will flash. This allows gates from one trigger to overlap the next trigger. When used with a light chopper reference as the trigger, the phase of the gates may be adjusted over nearly 360°.

B GATE=CW
FIXED
SCAN $\Delta=1.000 \mu\text{s}$

See A GATE menu above.

B DELAY=1.000 μs
: 2.000 μs

See A DELAY menu above.

B WIDTH=1.000 μs

See A WIDTH menu above.

SETUP MENU**LCD CONTRAST=12**

The LCD contrast may be adjusted for optimum viewing. Larger values increase the contrast.

STORE=3 EXECUTE=>

9 instrument setups may be stored. Parameters in the SETUP and COM menus are not stored by this command. SETUP and COM parameters are stored on power down as part of the current setup only.

Use the spin knob to select a storage location, 1 through 9. The right arrow key will store the setup.

RECALL=3 EXECUTE=>

9 user stored instrument setups and 1 default setup may be recalled. User locations 1 through 9 and DEFault are the setups available. Parameters in the SETUP and COM menus are not altered by this command.

Use the spin knob to select a storage location, 1 through 9 or DEFault. The right arrow key will recall the setup.

COM MENU

Communication parameters in this menu should not be altered while the computer interface is active. Changing any line except for DATA will reset the computer interfaces and communication buffers. Data in this menu is retained on power down as part of the current setup.

GPIB ADDR=23

The instrument's GPIB address can set from 1 to 30.

RS232 BAUD= 19200
9600
..
300

The instrument's RS232 baud rate can be set to any standard value from 300 to 19200 baud.

RS232 BITS= 8
7

The RS232 character length can be 7 or 8 bits. 8 bits is standard.

RS232 PARITY= NONE
EVEN
ODD

The RS232 parity can be even, odd, or none.

RS232 WAIT=6

The SR400 can be programmed to wait between each character it sends over the RS232 interface. This is to accomodate computers with slow computer interfaces. The wait interval is programmed in units of 3.3 ms and can be set from 0 to 25. The default value is 6 and should be set to a smaller value whenever possible.

RS232 ECHO= OFF
ON

To control the SR400 from a terminal, the ECHO feature is turned ON. This causes the unit to echo back commands received over the RS232 port. In this mode, the SR400 will send line-feeds in addition to carriage returns with each value returned and will also send the prompts 'OK>' and '??>' to indicate that the previous command was either processed correctly or contained an error.

When controlling the SR400 from a computer, the ECHO feature should be turned OFF.

DATA=

This line is the data window. The last 254 characters received over the computer interface can be played back to help find errors. The spin knob scrolls forward and backward through the window. Unrecognized control characters are displayed as 'cc'. Carriage return and line-feed are shown as 'cr' and 'lf'.

The window always returns to the most recent data whenever this line is first displayed or a new character is received. Changing any COM parameters clears the communication buffers and clears this window.

INSTRUMENT SETUPS

The SR400 Photon Counter is a very flexible instrument capable of many modes of operation. This section illustrates some of the measurements that can be made.

DEFAULT SETUP / POWER ON CLEAR

If the STOP key is held down when the power is turned on, the instrument settings will be set to the defaults shown below instead of the settings in effect when the power was turned off.

<u>Menu Line</u>	<u>Setting</u>
COUNT=	A,B FOR T PRESET
A=	INPUT 1
B=	INPUT 2
B SET=	1E3 (inactive)
T=	10 MHz
T SET=	1E0 s (1E7 counts)
N PERIODS=	1
AT N=	STOP
DWELL=	1E0 s
D/A=	A
D/A RANGE=	LOG
DISPLAY=	CONTINUOUS
TRIG SLOPE=	RISE
TRIG LVL=	+2.000 V
A DISC SLOPE=	FALL
A DISC=	FIXED ($\Delta=+ 0.0$ mV)
A DISC LVL=	- 10.0 mV
B DISC SLOPE=	FALL
B DISC=	FIXED ($\Delta=+ 0.0$ mV)
B DISC LVL=	- 10.0 mV
T DISC SLOPE=	FALL
T DISC=	FIXED ($\Delta=+ 0.0$ mV)
T DISC LVL=	- 10.0 mV
PORT1=	FIXED ($\Delta=+ 0.000$ V)
PORT1 LVL=	+ 0.000 V
PORT2=	FIXED ($\Delta=+ 0.000$ V)
PORT2 LVL=	+ 0.000 V
A GATE=	CW ($\Delta=0.000$ μ s)
A DELAY=	0.000 μ s
A WIDTH=	0.005 μ s
B GATE=	CW ($\Delta=0.000$ μ s)
B DELAY=	0.000 μ s
B WIDTH=	0.005 μ s

LCD CONTRAST= 12

STORE=	1
RECALL=	DEF
GPIB ADDR=	23
RS232 BAUD=	9600
RS232 BITS=	8
RS232 PARITY=	NONE
RS232 WAIT=	6
RS232 ECHO=	OFF

In addition, the RS232 termination sequence will be set to a carriage return and the GPIB SRQ mask reset to zero.

If the message RECALL ERROR is displayed on power up without the STOP being held down, then the instrument is ignoring the setup retained in memory. This can be due to a low battery. In this case, the instrument returns to the default setup.

The default setup may also be recalled using the RECALL menu line or via the remote CL command or a GPIB DCL or SDC command. In these cases, the SETUP and COM menus are not affected.

QUICK CHECK

Unplug all cables from the unit and recall the default setup by holding the STOP key down while turning the unit on. The message RECALL DEFAULT should appear on the display for ~3 s. Press the DOWN key once to go to the A=INPUT 1 menu line. Rotate the knob until A=10 MHz appears on the menu line. To start a count period, press the START key once. The data display should show counting in counter A for 1 second and stop at A= 10000000. Counter A will have counted the internal 10 MHz timebase for 1 second and reached 10,000,000 counts. To reset the counter, press the STOP key. The data display should reset to 0.

Now press the DOWN key 3 more times to display the N PERIODS line. Rotate the knob until N PERIODS=10. Start a scan by pressing the START key. The counter should cycle through 10 count periods separated by the dwell time (1 s). The display shows the current scan position in brackets. Reset the scan using the STOP key.

Start another scan with the START key but press

the STOP key before all 10 count periods are completed. This pauses the scan at the last completed count period. START will resume the scan and STOP will reset the scan.

In general, if the START key doesn't start a count period it is because the counters are paused at the end of a scan. In this case, the SCAN led will be off and the N PERIODS display shows that the programmed number of points has been taken. A STOP is required to reset the scan.

Pressing the STOP key twice in succession always resets the counters.

2 PHOTON PILE UP CORRECTION

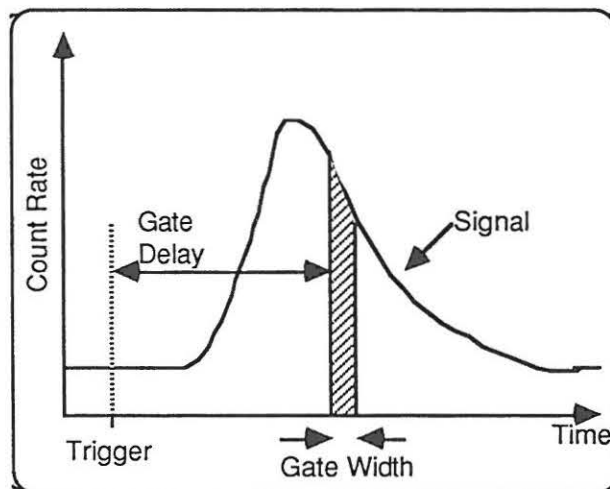
If the pulse rate from the PMT is high enough, significant number of 2 photon pulses will occur. These pulses are larger than single photon pulses and result when 2 photons hit the cathode at or near the same time. A single discriminator level will count 1 and 2 photon pulses equally giving an erroneous count. By using 2 discriminators, the 2 photon pulses can be counted correctly.

Set both A and B inputs to INPUT 1 and apply the signal to INPUT 1. Select the COUNT mode A+B FOR T PRESET. The count will now be the number of pulses that exceed the A DISC level plus the number that exceed the B DISC level. Set the A DISC level below the 1 photon level and the B DISC level above the 1 photon level but below the 2 photon level. 1 photon pulses will be counted only by A while 2 photon pulses will be counted by both A and B. A+B will be the number of 1 photon pulses plus twice the number of 2 photon pulses which is the correct number of photons.

GATED PHOTON COUNTING

When the photons of interest result from a source such as a pulsed discharge or laser, gated photon counting may be required to discriminate against photons which occur outside the time interval of interest.

Assume that the pulse rate is a function of time as shown below. The time scale may be nanoseconds up to hundreds of milliseconds. The A counter will be used to count the signal via INPUT 1.

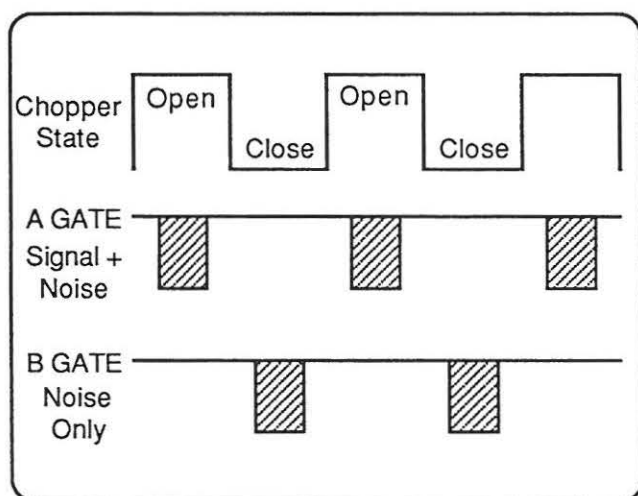


By using A GATE, shown as the shaded region, to enable the counter, the count rate versus time may be measured. Only those photons which occur within the gate time are counted. By scanning the gate between count periods, the entire time behavior is recovered. Set the T input to TRIG to enable an equal number of gates during each count period.

If background subtraction is required, the B counter can be used along with the A counter. The B input is INPUT 1 and the B DISC is set to the same level as A DISC. Adjust B GATE to be equal in width to A GATE and position it at a FIXED delay either before the signal or after the signal to count the background rate. Using the COUNT mode A-B FOR T PRESET, the background counts will be subtracted from the signal counts.

SYNCHRONOUS OR "CHOPPED" PHOTON COUNTING

In noisy environments or when the signal count rate is very low, a mechanical light chopper may be used to improve signal to noise. This is very similar to lock-in detection of analog signals. Two gated counters are used, one counts the signal plus background during the chopper open state and the other counts the background during the chopper closed state. The difference between the two counters is the signal. If A counter is the signal counter and B counter the background counter, the two gates should be timed as shown below.



The two gate widths need to be the same width. The gates are triggered by the chopper reference signal. The gate delays are used to position the gates correctly. Since the gate delay can be as large as the trigger period (minus $1\ \mu\text{s}$), the phase of the gates relative to the trigger can be adjusted to cover the open and closed cycles of the chopper. The gate widths should be adjusted to be less than the open and closed periods of the chopper. This avoids chopper frequency drift or jitter from disturbing the gate/chopper overlap. Also, if the chopper aperture does not result in a sharp turn on or turn off of the signal, the transition region between open and closed can be avoided.

The COUNT mode A-B FOR T PRESET is used with T input = TRIG. This displays the difference between the two channels and ensures that equal numbers of open and closed gates are counted during each count period.

REMOTE PROGRAMMING

The SR400 Photon Counter is remotely programmable via both RS232 and GPIB (IEEE-488) interfaces. It may be used with laboratory computers or simply with a terminal. All front panel features (except power) may be controlled and read via the computer interfaces. The SR400 also has two programmable analog output ports available to provide general purpose control voltages.

COMMUNICATING WITH THE SR400

Before using either the RS232 or GPIB interface, the appropriate communication parameters need to be set. These parameters are set from the front panel in the COM menu. If the GPIB interface is being used, only the GPIB address of the instrument needs to be set. If the RS232 interface is being used, the baud rate, number of bits in a character, and parity need to be set. The RS232 wait interval programs the delay between characters sent from the SR400. The delay is the wait value times 3.3 ms. For most computers this can be set to 0. Some terminal emulators may require a delay. The RS232 echo mode should be OFF if the SR400 is connected to a computer. It may ON if connected to a terminal.

COMMAND SYNTAX

Communications with the SR400 use ASCII characters. Commands to the SR400 may be in either UPPER or lower case.

A command to the SR400 consists of two command letters, arguments or parameters if necessary, and an ASCII carriage return (<cr>) or line-feed (<lf>) or both. The different parts of the command do not need to be separated by spaces. Spaces can improve the readability of your programs and, if included in a command, they will be ignored. If more than one parameter is required by a command, the parameters must be separated by a comma. Examples of commands are:

CM 1 <cr>	set the counting mode to A-B FOR T PRESET
CI 0,1 <cr>	set the input to counter A to INPUT 1
NP 5E2 <cr>	set N PERIODS to 500
GD 0,1.2E-6 <cr>	set A gate delay to 1.2 μ s
PL 1, -3.4E-1 <cr>	set the PORT1 output to -0.34 V

Multiple commands may be sent on a single line. The commands must be separated by a semicolon (;) character. The commands will not be executed until the terminating carriage return is sent.

An example of a multiple command is:

CM 1; CI 0,1; GD 0,1.2E-6 <cr>

It is not necessary to wait between commands. The SR400 has a command input buffer of 256 characters and processes the commands in the order received. Likewise, the SR400 has an output buffer (for each interface) of 256 characters. If a buffer overflows, the message "DATA BUFFER OVERFLOW" appears on the LCD display for 5 s and all buffered data is erased.

In general, if a command is sent without parameters, it is interpreted as a request to read the value of the associated function or setting. Values returned by the SR400 are sent as a string of ASCII characters terminated usually by carriage return, line-feed. For example, after the above command is sent, the following read commands would generate the responses shown below.

<u>Command</u>	<u>Response from SR400</u>
CM	1<cr><lf>
CI 0	1<cr><lf>
GD 0	1.2E-6<cr><lf>

The choice of terminating characters sent by the SR400 is determined by which interface is being used and whether the 'echo' feature is in use. The terminating sequence for the GPIB interface is always <cr><lf>. The default sequence for RS232 is <cr> when the echo mode is off, and <cr><lf> when the echo mode is on. The terminating sequence for the RS232 interface may be changed using the SE command.

Note that the terminating characters are sent with each value returned by the SR400. Thus, the response to the command string CM;CI0;GD0<cr> while using the RS232 non-echo mode would be 1<cr>1<cr>1.2E-6<cr>.

In general, it is good programming practice to receive the response from one query command before sending another command.

FRONT PANEL STATUS LED's

The ACT LED flashes whenever the SR400 is sending or receiving characters over the computer interfaces.

The ERR LED flashes whenever an error has occurred, such as, an illegal command has been received, a parameter is out of range, or a communication buffer has exceeded 240 characters.

The REM LED is on whenever the SR400 is programmed to be in the remote (front panel locked out) state.

RS232 ECHO AND NO ECHO OPERATION

In order to allow the SR400 to be operated from a terminal, an echo feature has been included which causes the unit to echo back commands received over the RS232 port. This feature is enabled by setting the RS232 ECHO mode to ON in the COM menu. In this mode, the SR400 will send line-feeds in addition to carriage returns with each value returned and will also send the prompts 'OK>' and '??>' to indicate that the previous command line was either processed correctly or contained an error. Operating the SR400 from a terminal is an ideal way to learn the commands and responses before attempting to program a computer to control the SR400. When the unit is controlled by a computer, the echo feature should be turned OFF to prevent the sending of spurious characters which the computer is not expecting.

TRY-OUT WITH AN ASCII TERMINAL

Before attempting any detailed programming with the SR400, it is a good idea to try out the commands using a terminal. Connect a terminal with an RS232 port to the RS232 connector on the rear panel of the SR400. A 'straight' RS232 cable is required since the SR400 is a DCE and the terminal

is a DTE. After connecting the terminal, hold down the STOP key while turning the unit on. This causes the SR400 to assume its default settings so that the following discussion will agree with the actual responses of the SR400. Set the baud rate, parity, and character length to match the terminal using the COM menu. The RS232 echo mode should be changed to ON. Now turn the unit off and on again. The ACT LED will flash for a second and the sign-on message will appear on the terminal. Following the message, the prompt 'OK>' will be displayed. This indicates that the SR400 is ready to accept commands. The sign on message is only sent if the RS232 echo mode is ON.

Type the letters 'CM' followed by a carriage return (CM<cr>). The SR400 responds by sending to the terminal the character 0 indicating that the counting mode is A,B FOR T PRESET. In general, a command with no arguments or parameters reads a setting of the unit. Now type the command MD1,2<cr>. The menu line of the display should now read A=INPUT 1. Type the command CI0,0<cr> to change the counter A input to 10 MHz. The menu line should now read A=10MHz.

Now use the command CS<cr> to start counting. After one second, counter A should have 10000000 displayed (10 MHz for 1 second). The data from counter A can now be read. Type the command QA<cr> and the response should be the counter A value.

Attach a dc voltmeter to the PORT1 D/A output on the rear panel. The range should allow for 10V readings. The voltage at this output can be set using the PL command. Type PL1,5.0<cr> and the output will change to 5.0V. To read this back to the terminal, just type PL1<cr>. When setting the voltage, the value may be sent as an integer (5), real (5.000), or floating point (0.500E1) number.

At this point, the user should experiment with a few of the commands. A detailed command list follows.

DETAILED COMMAND LIST

The two letters in each command sequence specify the command. The rest of the sequence consists of parameters. Multiple parameters are separated by a comma. Those parameters shown in { } are optional while those without { } are required.

Variable i is an integer which selects a counter:

i	Counter
0	A
1	B
2	T

Variables i,j,k,l, and m are integers and must be expressed in integer format.

Variable n is a real number ≥ 1 with one significant digit (i.e. 20 or 5E5).

Variable x is a real number with one significant digit (i.e. 2E-3 or .002).

Variables n,x,t, and v are real numbers in integer, real, or floating point format.

MODE

CM { j }

If j is included, the CM command sets the counting mode. If j is absent, the counting mode is returned.

j	Counting Mode
0	A,B FOR T PRESET
1	A-B FOR T PRESET
2	A+B FOR T PRESET
3	A FOR B PRESET

Setting the counter mode also performs a counter reset.

CI i { ,j }

The CI command sets and reads the counter inputs. i = 0, 1, or 2 selecting counter A,B, or T and is required. If j is included, then the input to counter i is changed to source j. If j is absent, the input to counter i is returned.

j	Input
0	10 MHz
1	INPUT 1
2	INPUT 2
3	TRIG

Counter	Allowed Inputs
A	10 MHz, INPUT1
B	INPUT 1, INPUT 2
T	10 MHz, INPUT 2, TRIG

Changing a counter input while counting may result in bad count data.

CP i { ,n }

The CP command sets and reads the counter presets. If i = 1 then counter B is selected, if i = 2 counter T is selected. The parameter i is required. If n is included, then counter i is preset to n where n is a value from 1 to 9E11. n may be expressed in any format but must be greater than or equal to 1 and only the most significant digit is used. For example, "CP2,10" , "CP2,1E1" , "CP2,0.1E2" , "CP2,12" all set T SET to 1E1. Changing a counter preset while counting causes the counters to pause. A counter start command or keypress resumes counting. If n is absent, then the preset value of counter i is returned. In the above example, the string "1E1" is returned. Note that reading and setting the preset values of a counter not currently in the preset mode is allowed.

NP { m }

If m is absent, the NP command returns the programmed number of periods in a scan (N PERIODS). If m is included, the number of periods is set to the value m. The parameter m must be an integer from 1 to 2000. Changing the number of periods during a scan is allowed. If the new value exceeds the current scan position, then the scan will continue to the new number of periods. If the new value is lower than the current scan position, then the scan ends at the next end of count period.

NN

The NN command reads the scan position. The scan position is the number of count periods or data points completed so far during the current scan and is shown in the brackets on the N PERIODS menu line. The NN command is a read only command.

NE { j }

If $j = 0$, the NE command sets the scan end mode (AT N) to STOP. If $j = 1$, the mode is set to START. The scan end mode may be changed at any time. If j is absent, the scan end mode is returned.

DT { x }

The DT command reads and sets the dwell time. If x is absent, the dwell time is returned. If x is included, the dwell time is set to x seconds where $2E-3 \leq x \leq 6E1$. If $x = 0$, the dwell is set to EXTERNAL. Only the most significant digit of x is used so "DT 2E-3", "DT .0022", "DT 2.2E-3" all set the dwell to 2 ms.

AS { j }

The AS command reads and sets the front panel D/A output source. If j is absent, the D/A source is returned. The parameter j may only be included if the counting mode is A,B FOR T PRESET. In this case, $j=0$ or 1 selects A or B. In all other modes, the D/A output follows the count display. The source may be read in all modes.

j	Output source
0	A
1	B
2	A-B
3	A+B

AM { j }

If j is included, the AM command sets the D/A output scale. If j is absent, the D/A output scale is returned.

j	Output scale
0	Log (1V/decade)
1	987#####
2	#876#####
3	##765####
4	###654###
5	####543##
6	#####432#
7	#####321

SD { j }

If $j = 1$, the SD command sets the count display mode to HOLD. If $j = 0$, the display mode will be CONTINUOUS. If j is absent, the display mode is returned.

LEVELS

TS { j }

If $j = 1$, the TS command sets the gate trigger slope to FALL or negative. If $j = 0$, the gate trigger slope is set to RISE or positive. If j is absent, the gate trigger slope is returned.

TL { v }

If v is absent, the TL command returns the gate trigger level. When v is included, the trigger level is set to v Volts where $-2.000 \leq v \leq 2.000$. The resolution is .001 V.

DS i { ,j }

The DS command sets and reads the discriminator slopes. $i = 0, 1$, or 2 designates discriminator A,B, or T and is required. If j is included, the selected discriminator slope is set to RISE, $j=0$, or FALL, $j=1$. If j is absent, the slope of discriminator i is returned.

DM i { ,j }

The DM command sets and reads the discriminator modes. $i = 0, 1$, or 2 designates discriminator A,B, or T and is required. If j is included, the selected discriminator is set to FIXED, $j=0$, or SCAN, $j=1$. If j is absent, the mode of discriminator i is returned. The mode may be changed during a scan. If it is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start value immediately.

DY i { ,v }

The DY command sets and reads the discriminator scan step size. $i = 0, 1$, or 2 designates discriminator A,B, or T and is required. If v is included, the selected discriminator scan step is set to v Volts where $-0.0200 \leq v \leq 0.0200$. The resolution is 0.0002 V (0.2 mV). If v is absent, the selected discriminator scan step is returned. The scan step may be set and read even if the selected discriminator is in the FIXED mode. Changing the scan step during a scan will affect the level for the next count period and all those that follow.

DL i { ,v }

The DL command sets and reads the discriminator levels. $i = 0, 1$, or 2 designates discriminator A,B, or T and is required. If v is included, the selected discriminator level is set to v Volts where $-0.3000 \leq v \leq 0.3000$. The resolution is 0.0002 V (0.2 mV). If v is absent, the selected discriminator level is re-

turned. If the discriminator mode is SCAN, the DL command sets and reads the starting value of the level. Changing the level during a scan will cause the level to be set to the new value on the next count period. Note that the scan is not reset. Reading the level during a scan always returns the starting value, not the current value.

DZ i

The DZ command returns the level of discriminator *i* during a scan. *i* = 0, 1, or 2 designates discriminator A, B, or T and is required. If discriminator *i* is in the SCAN mode, its current level during a scan is read. If the counters are in reset or the discriminator is in the FIXED mode, the DZ command returns the starting value for the level (same as DL command). The DZ command is a read only command.

PM i { ,j }

The PM command sets and reads the rear panel PORT1 and PORT2 D/A output modes. *i*=1 or 2 to select PORT1 or PORT 2 and is required. If *j* is included, the mode is set to FIXED, *j*=0, or SCAN, *j*=1. If *j* is absent, the mode is returned. The mode may be changed during a scan. If it is changed to SCAN, the level begins scanning from the start value on the next count period. If it is changed to FIXED, the level returns to the start value immediately.

PY i { ,v }

The PY command sets and reads the PORT1 and PORT2 scan steps. *i*=1 or 2 to select PORT1 or PORT 2 and is required. If *v* is included, the scan step is set to *v* Volts where $-0.500 \leq v \leq 0.500$. The resolution is 0.005 V (5 mV). If *v* is absent, the scan step is returned. The scan step may be set and read even if the selected port is in the FIXED mode. Changing the scan step during a scan will affect the level for the next count period and all those that follow.

PL i { ,v }

The PL command sets and reads the PORT1 and PORT2 output levels. *i*=1 or 2 to select PORT1 or PORT 2 and is required. If *v* is included, the output level is set to *v* Volts where $-10.000 \leq v \leq 10.000$. The resolution is 0.005 V (5 mV). If *v* is absent, the output level is returned. If the Port output mode is SCAN, the PL command sets and reads the starting value of the level. Changing the level during a scan will cause the level to be set to the new value on the next count period. Note that the scan is not reset. Reading the level during a

scan always returns the starting value, not the current value.

PZ i

The PZ command returns the level of the PORT1 and PORT2 outputs during a scan. *i*=1 or 2 to select PORT1 or PORT 2 and is required. If the selected port is in the SCAN mode, its current level during a scan is read. If the counters are in reset or the output is in the FIXED mode, the PZ command returns the starting value for the level (same as PL command). The PZ command is a read only command.

GATES

GM i { ,j }

The GM command sets and reads the gate modes. *i* = 0 or 1 designates A gate or B gate and is required. If *j* is included, the selected gate is set to CW, *j*=0, or FIXED, *j*=1, or SCAN, *j*=2. If *j* is absent, the mode of gate *i* is returned. The mode may be changed during a scan. If it is changed to SCAN, the delay begins scanning from the start position on the next count period. If it is changed to FIXED, the delay returns to the start position immediately. If it is changed to CW, the gate immediately changes to continuously on.

GY i { ,t }

The GY command sets and reads the gate delay scan step. *i* = 0 or 1 designates A gate or B gate and is required. If *t* is included, the selected gate scan step is set to *t* seconds where $0 \leq t \leq 99.92E-3$. The resolution is variable depending upon the 4 most significant digits of *t*.

4 most significant digits	resolution in 4th digit
1000 to 2048	1 (1001,1002,1003,...)
2048 to 4096	2
4096 to 8192	4
8192 to 9992	8 (8192,8200,8208,...)

For example, near 10.00 μ s, the allowed values of *t* are 9.984 μ s, 9.992 μ s, 10.00 μ s, 10.01 μ s, etc. Below 1.000 μ s, the resolution is 1 ns. Almost any value of *t* may be sent and the SR400 will round to the nearest allowed value. If *t* is absent, the selected gate scan step is returned. The scan step may be set and read even if the selected gate is not in the SCAN mode. Changing the scan step during a scan will affect the gate position for the next count period and all those that follow.

GD i { ,t }

The GD command sets and reads the gate delays. $i=0$ or 1 designates A gate or B gate and is required. If t is included, the selected gate delay is set to t seconds where $0 \leq t \leq 999.2E-3$. The resolution is variable depending upon the 4 most significant digits of t .

4 most significant digits	resolution in 4th digit
1000 to 2048	1 (1001,1002,1003,...)
2048 to 4096	2
4096 to 8192	4
8192 to 9992	8 (8192,8200,8208,...)

For example, near $10.00 \mu s$, the allowed values of t are $9.984 \mu s$, $9.992 \mu s$, $10.00 \mu s$, $10.01 \mu s$, etc. Below $1.000 \mu s$, the resolution is $1 ns$. Almost any value of t may be sent and the SR400 will round to the nearest allowed value. If t is absent, the selected gate delay is returned. The gate delay may be set and read regardless of the gate mode. If the mode is CW, the gate remains continuous. If the mode is SCAN, the GD command sets and reads the starting position of the gate delay. Changing the delay during a scan will cause the delay to be set to the new value on the next count period. Note that the scan is not reset. Reading the delay during a scan always returns the starting delay, not the current delay.

GZ i

The GZ command returns the gate delay position during a scan. $i = 0$ or 1 designates A gate or B gate and is required. If gate i is in the SCAN mode, its current delay position during a scan is read. If the counters are in reset or the gate is not in the SCAN mode, the GZ command returns the starting delay position (same as GD command). The GZ command is a read only command.

GW i { ,t }

The GW command sets and reads the gate widths. $i = 0$ or 1 designates A gate or B gate and is required. If t is included, the selected gate width is set to t seconds where $0.005E-6 \leq t \leq 999.2E-3$. The resolution is variable depending upon the 4 most significant digits of t .

4 most significant digits	resolution in 4th digit
1000 to 2048	1 (1001,1002,1003,...)
2048 to 4096	2
4096 to 8192	4
8192 to 9992	8 (8192,8200,8208,...)

For example, near $10.00 \mu s$, the allowed values of t are $9.984 \mu s$, $9.992 \mu s$, $10.00 \mu s$, $10.01 \mu s$, etc. Below $1.000 \mu s$, the resolution is $1 ns$. Almost any value of t may be sent and the SR400 will round to the nearest allowed value. If t is absent, the selected gate width is returned. The gate width may be read and set regardless of the gate mode. If the mode is CW, the gate remains continuous. If the mode is FIXED or SCAN, the gate width is adjusted immediately.

NOTE

Changing the gate delay or width may cause some gates to be missed or shortened. Changing gate parameters while counting may therefore cause bad data to be collected. It is recommended that the counters be paused before changing gate values and then started again.

FRONT PANEL

CS

Sending the CS command has the same effect as pressing the START key. The CS command may be sent at any time but only has an effect if the counters are paused or in reset. If the DWELL TIME is EXTERNAL, the CS command will start a count period. A CH command may be sent to terminate the count period.

CH

Sending the CH command has the same effect as pressing the STOP key. The CH command may be sent at any time. If received while scanning, the scan will pause. If received while paused, the counters will reset. If the DWELL TIME is EXTERNAL, the CH command may be used to terminate count periods.

CR

Sending the CR command resets the counters. All scanning parameters return to their start values and previously buffered scan data is lost. The CR command has the same effect as pressing the STOP key twice in succession.

CK j

The CK command simulates pressing a front panel key. j is a required parameter from 0 to 13 . The CK command may be sent at any time and is not restricted by the remote/local state of the front panel.

j	key
0	DOWN CURSOR
1	RIGHT CURSOR
2	LEVEL
3	SETUP
4	COM
5	STOP
6	LOCAL
7	RESET
8	LEFT CURSOR
9	UP CURSOR
10	MODE
11	AGATE
12	BGATE
13	START

RR

The RR command simulates rotating the knob right (clockwise or up) one step. The displayed parameter will be altered if its cursor is active.

RL

The RL command simulates rotating the knob left (counterclockwise or down) one step. The displayed parameter will be altered if its cursor is active.

SC

The SC command reads the cursor position. "0" is returned for left, "1" for right, and "2" for inactive. To move the cursor, use the CK command.

MI j

The MI command sets the front panel inhibit mode. j is a required parameter from 0 to 2.

j	Mode
0	Local: all front panel controls are operative
1	Remote: front panel controls are not operative. The LOCAL (COM) key returns the status to local.
2	Lock-out: front panel controls are not operative. No key returns the status to local. Another MI command is needed to return to local.

The MI command may only be sent via the RS232 interface. The GPIB interface uses REN, LLO, and GTL to set the front panel mode.

MS { string }

The MS command writes user messages on the menu line of the LCD display. String may be up to 24 characters long with spaces represented by underscores. For example, the command "MS HELLO_THERE" puts the message "HELLO THERE" on the display. The display will return to normal when any active key is pressed. It will also return to normal if the current menu line needs to be updated during a scan. The MS command without a string returns the display to normal immediately.

MD j,k

The MD command changes the menu display to line k of menu j. The allowed combinations of j and k are shown below. Both j and k are required parameters.

j,k	
1,1	COUNT
1,2	A
1,3	B
1,4	T
1,5	N PERIODS
1,6	AT N
1,7	D/A OUT
1,8	D/A RANGE
1,9	DISPLAY
2,1	A GATE
2,2	A DELAY
2,3	A WIDTH
3,1	B GATE
3,2	B DELAY
3,3	B WIDTH
4,1	TRIG SLOPE
4,2	TRIG LVL
4,3	A DISC SLOPE
4,4	A DISC MODE
4,5	A DISC LVL
4,6	B DISC SLOPE
4,7	B DISC MODE
4,8	B DISC LVL
4,9	T DISC SLOPE
4,10	T DISC MODE
4,11	T DISC LVL
4,12	PORT1 MODE
4,13	PORT1 LVL
4,14	PORT2 MODE
4,15	PORT2 LVL

<u>j,k</u>	
5,1	GPIB ADDR
5,2	RS232 BAUD
5,3	RS232 BITS
5,4	RS232 PARITY
5,5	RS232 WAIT
5,6	RS232 ECHO
5,7	DATA
6,1	LCD CONTRAST
6,2	STORE
6,3	RECALL

MM

The MM command reads the menu number of the display (j in the above table). The MM command is a read only command.

ML

The ML command reads the menu line of the display (k in the above table). The ML command is a read only command.

INTERFACE

CL

The CL command resets the unit to its default state. The default front panel settings are listed in the Instrument Setups section. Parameters in the SETUP and COM menus are not changed. In addition, the SRQ mask is cleared, and the RS232 terminating sequence is reset to the proper defaults.

The command and output buffers are cleared by the CL command. Therefore, it is bad practice to use the CL command before all previous commands have been processed and all responses have been received. There should never be any other commands on the same line as a CL command.

SS { j }

The SS command reads the status byte. (See the following section for a definition of the status byte). j designates one bit, 0-7, of the status byte. If j is included, the designated bit of the status byte is returned ("0" or "1"). The bit which is read is then reset. If j is absent, the value of the entire byte is returned (0-255) and all status bits are then reset.

SI { j }

The SI command reads the secondary status byte. (See the following section for a definition of the secondary status byte). j designates one bit, 0-2,

of the byte. If j is included, the designated bit of the secondary status byte is returned ("0" or "1"). The bit which is read is then reset. If j is absent, the value of the entire byte is returned (0-7) and all secondary status bits are then reset.

SV { m }

If m is included, the SV command sets the GPIB SRQ (service request) mask to the value m (0-255). If m is absent, the value of the SRQ mask is returned. The SV command may only be sent via the GPIB interface.

SW { m }

The SW command sets and reads the RS232 character wait interval. If m is included, the SR400 will wait m times 3.3 ms between characters sent over the RS232 interface. This allows slow computer interfaces to keep up. m can range from 0 to 25. If m is absent, the wait value is returned. The SW command may only be sent via the RS232 interface.

SE { j,k,l,m }

The SE command sets the RS232 end-of-record characters sent by the SR400 to those specified by the decimal ASCII codes (0-127) j,k,l, and m. If no argument is included, the end-of-record sequence returns to the default (a carriage return), otherwise, up to four characters may be specified. The end-of-record required by the SR400 when receiving commands is not affected. The SE command may only be sent via the RS232 interface.

STORE/RECALL

ST m

The ST command stores the instrument settings to location m = 1 to 9. Parameters in the SETUP and COM menus are NOT stored by this command. The SETUP and COM parameters are stored on power down as part of the current setup.

RC m

The RC command recalls instrument settings from location m = 0 to 9. If m = 0, the default settings are recalled (see the Instrument Setups section for a list of the default settings). In all cases, the SETUP and COM parameters are not altered by this command. The RC command also resets the counters. The menu display returns to the menu displayed when the instrument settings were stored. This is very often the STORE menu.

POLLED DATA**QA****QB**

The QA command reads the most recent complete data point from counter A, QB reads the most recent data point from counter B. QA and QB commands should only be sent after checking the Data Ready status bit. This bit is set at the end of each complete count period and signals the availability of valid data. The Data Ready status bit is reset after it is read. Sending QA or QB commands without polling the Data Ready status can cause data points to be read multiple times. Note that QA and QB do not reset the Data Ready status.

If data is not ready, the QA and QB commands return -1. If counter B is preset, QB returns -1.

QA m**QB m**

When m is included in a QA or QB command, buffered data from scan point m is read. m = 1 is the first data point in the scan and m = N PERIODS is the last data point. If m > 2000 or m = 0, an error occurs. If counter B is preset, then QB m returns -1.

If m > last completed point of the current scan, then -1 is returned. Therefore during a scan, QA 1 can be sent until a positive count is returned. At this point, the QA 2 command should be sent until a positive count is returned. All data from the scan can be read in this manner. This is the fastest way to read data during a scan while still maintaining full handshaking.

Data from the scan buffer can also be read after the scan is completed. The Scan Done status bit can be polled to determine when the scan finishes. The Scan Done status bit is set at the end of a scan only if the scan end mode is STOP. The Scan Done status bit is reset after it is read. To prevent ambiguity, the status byte should be cleared before starting a scan and then polled to determine when the scan is finished.

Whenever scan data is to be read, the scan end mode should be STOP and a new scan started only after reading the desired data from the previous scan.

If m is absent or between 1 and 2000, data is always returned from a Q command. The value -1 is reserved for errors.

CONTINUOUS DATA**EA****EB****ET**

The E commands dump the scan buffers at the end of a scan. The E commands may only be sent while paused at the end of a scan. Data will be sent continuously starting with data point 1 and ending with data point N PERIODS. Each data point is separated by an end-of-record sequence. If an E command is used with the RS232 interface, the controlling computer interface should be interrupt driven or have fast data communication routines since the data transmission is limited only by the baud rate and the character wait interval.

The EA command dumps the counter A buffer, EB dumps the counter B buffer, and ET dumps both counters. EB and ET generate errors if counter B is preset. Data from the ET command is ordered A1,B1,A2,B2,... with each data point separated by the terminator sequence.

FA**FB****FT**

The F commands start a new scan. Data is returned at the end of each count period. Data continues to be returned until all N PERIODS points are completed and sent. If data points accumulate faster than they can be sent over the computer interface, they are buffered and sent as fast as the interface allows. F commands may only be sent when the counters are in reset.

The FA command starts the scan and returns the next N PERIODS data points from counter A. FB returns points from counter B and FT returns points from both counters. FB and FT generate errors if counter B is preset. Data from the FT command is ordered A1,B1,A2,B2,... with each data point separated by the terminator sequence.

NOTE

Scan buffers are reset at the start of each scan. When the counters reset, data from the previous scan is lost and any actively transmitting E or F command is terminated. Thus, programs may hang if the counters are reset or a new scan is started before all data points from the previous scan are received. To prevent this, the scan end mode should be STOP and a

new scan initiated only after receiving N PERIODS data points. An E command sent while a previous E command is still executing generates an error.

EXAMINE DATA

XA

XB

The X commands read the counter contents regardless of the count state. An X command sent while counting returns the current counter contents. If the counters are in the dwell cycle, paused, or reset, the counter value will be zero. The display may still be showing the most recently completed count but the counters are already internally reset. The X command is generally only useful for long count periods and/or slow count rates. If the count rate is faster than the interface speed, then the returned count values may be in error.

The XA command reads the contents of counter A. XB reads the contents of counter B. XB returns -1 if counter B is preset.

STATUS BYTE

The SR400 maintains an 8-bit status register which the user may read to obtain information on the unit's status. The status byte is read using the SS command which returns the value of the byte in ASCII coded decimal. The returned status byte reflects all of the status conditions which have occurred since the last time the byte was read. After the status byte has been read, it is cleared. Thus, the status byte should be read initially to clear all previous conditions (especially after a power up or after settings have been changed).

The definitions for each bit of the status byte are given below:

Bit 0

Parameter Changed from front panel. When this bit is set, it indicates that the front panel knob has been rotated and a setting has been altered.

Bit 1

Data Ready. This bit is set at the end of each count period to indicate that counter data is available.

Bit 2

Scan Finished. This bit is set at the end of a scan if

the scan end mode is STOP to indicate that the scan buffer is available for reading. This bit is not set if the scan end mode is START.

Bit 3

Counter Overrun. This bit is set whenever counter A or B exceeds or equals 10^9-1 counts.

Bit 4

Rate Error. This bit is set whenever a gate is missed. This can occur if a gate delay or width exceeds the trigger period minus 1 μ s.

Bit 5

Recall Error. This bit is set if a recall from a stored setting detects an error in the recalled data. If an error is found, the instrument setup is not altered.

Bit 6

SRQ. This bit is always zero when polled via the SS command. This bit is set in the status byte which is read via the GPIB serial poll following a service request generated by the SR400.

Bit 7

Command Error. This bit is set when an illegal command is received.

SECONDARY STATUS BYTE

The secondary status byte is actually only a 3 bit register. The secondary status byte is read using the SI command and returns a value from 0 to 7 in ASCII. The secondary status byte operates the same as the status byte above except that it cannot generate service requests.

The definitions for each bit of the status byte are given below:

Bit 0

Triggered. This bit is set when the gate generator is triggered.

Bit 1

Inhibited. This bit is set when the discriminators are inhibited.

Bit 2

Counting. This bit samples the state of the counters. If counting is in progress when this bit is polled, a 1 is returned, otherwise a 0 is returned.

ERRORS/DATA WINDOW

Whenever a parameter out of range or an unrecognized command error occurs, bit 7 of the status byte is set and the ERR LED flashes. In addition, any commands remaining on the current command line (up to the next <cr>) are lost. To aid in determining the cause of the error, the last 254 characters received by the SR400 may be viewed on the LCD display using the DATA line of the COM menu. The knob may be used to scroll through the command buffer. The display always returns to the most recent data whenever the menu line is brought up or a new character is received. Any command that clears the communication buffers also clears this display window. This includes changing any COM parameters, resetting the instrument, or a buffer overflow.

TROUBLE-SHOOTING INTERFACE PROBLEMS

If you are having difficulty getting your computer to communicate with the SR400 look to the following sections on the RS232 and GPIB interfaces for some tips specific to your particular interface.

An ASCII terminal is a valuable aid for debugging interface problems. You can use it to:

- 1) become familiar with the SR400's command structure
- 2) substitute a human for the SR400 by using a null modem cable (to make the DTE a DCE) and attaching the terminal to the port to which you would normally have connected the SR400. This allows you to test your program's responses to inputs which you provide from the terminal.

COMMON HARDWARE PROBLEMS INCLUDE:

- 1) The RS232 or GPIB cables are not properly attached.
- 2) The COM parameters are incorrect for the RS232 interface or the GPIB address is not set correctly (Make sure the RS232 echo is off when using the RS232 interface with a computer.)

- 3) Your computer requires an RS232 control line to be asserted, but your cable does not pass it between the SR400 and the computer, or, your computer is not asserting the DTR line on the RS232 .

COMMON SOFTWARE PROBLEMS INCLUDE:

- 1) You have sent the wrong command to ask for data from the SR400. Your program will wait forever for a response which is not going to come. This may not be your fault; we have seen Microsoft's Interpreted Basic on the IBM PC occasionally send a curly bracket (ASCII 253) when it was supposed to have sent a carriage return (ASCII 13).
- 2) Your computer's baud rate has been changed and no longer matches the SR400's baud rate.
- 3) The initial command sent to the SR400 was invalid due to a garbage character received during power-up, or, the first character in your computer's UART is garbage, also due to power-up. It is good practice to send a few carriage returns to the SR400 when your program begins, and have your program clear-out the UART at the start of your program.
- 4) The SR400 is not sending the correct 'end-of-record' marker for your computer. For example, it appears that Microsoft's Rev 3.2 FORTRAN on the IBM PC under DOS 2.1 requires two carriage returns for an end-of-record marker. The SE command can be used to set the SR400 end-of-record marker to 2 carriage returns. [The end-of-record marker is that sequence which indicates that the response is complete. From the keyboard, a single carriage return is the end-of-record marker.]
- 5) Answers are coming back from the SR400 too fast, overwriting the end-of-record markers, and causing the computer to hang, waiting for a complete response. In this case, the SW command can be used to slow down the response time of the SR400 preventing overwriting.
- 6) Answers are coming back from the SR400 too slowly due to the RS232 wait interval being too large. Use the SW0 command at the beginning of your program to speed up the transmission from the SR400.

SR400 WITH THE RS232 INTERFACE

The RS232 is a popular serial interface standard for bit serial communication. Despite the existence of the standard there are many permutations of control lines, baud rates, and data formats.

DATA COMMUNICATIONS EQUIPMENT (DCE)

The SR400 is configured as DCE so that it may be connected directly to a terminal or IBM PC. If the SR400 is to be interfaced with another DCE device, a special cable (sometimes referred to as a 'null modem' cable) is required. To use the RS232 interface you must set the COM parameters to match your computer's baud rate, parity, and number of bits.

RS232 ECHO

The RS232 ECHO should be OFF whenever the SR400 is interfaced with a computer. This prevents unwanted characters from being sent back to the computer after every command. The echo should be ON only if a terminal (or terminal emulator program) is being used.

WAIT COMMAND

The SR400 normally waits until the RS232 'Clear to Send' control line (CTS) is asserted before sending characters. However, some computers do not set and reset the CTS line, possibly causing the SR400 to send data when the computer is not ready to read it. The SR400 may be 'slowed down' using the SW command. Sending 'SWn' causes the unit to wait $n \times 3.3$ ms before sending each character over the RS232 interface. The command SW0 sets the wait interval to zero and results in the fastest transmission. The wait interval is set to 6 (20 ms) on power-up reset (STOP key held on power on).

TERMINATION SEQUENCES

The default RS232 termination characters are sufficient to interface with most computers, however, it will occasionally be necessary to send special terminating sequences to fit the requirements of some computers. This can be done with the SE command. The format for the command is:

SE {n1,n2,n3,n4}

where n1, n2, n3, and n4 are decimal values between 0 and 255 corresponding to the ASCII codes of the desired termination characters. For instance, if the desired termination sequence is an asterisk, (ASCII 42), two carriage returns, (ASCII 13), and a line-feed, (ASCII 10), the appropriate command is:

SE 42,13,13,10

If an NP command is sent requiring an answer of 100 (N PERIODS = 100), the SR400 would respond with the string

100*<cr><cr><lf>

Up to four terminating characters may be specified by the SE command. If no arguments are sent with the SE command, the terminating sequence returns to the default (echo on: <cr><lf>; echo off: <cr>). The SE command does not affect the terminating character (<cr>) required at the end of commands received by the SR400. It also does not affect the terminating sequence sent with data over the GPIB interface.

EXAMPLE PROGRAMS

The following section lists several example programs which use the RS232 port to control the SR400. All of the programs do the same thing, only the language is different. The programs set up the SR400 to count 10 MHz on counter A for 1 second. The counter data is read and the PORT1 output is ramped.

**PROGRAM EXAMPLE 1:
IBM PC, BASIC, VIA RS232**

In this example, the IBM PC's serial port (known as COM1:) is used to communicate with the SR400. Only two wires between the IBM PC's COM1: port and the SR400 are needed (pins #2 & #3 of the 25 pin RS232 connector), but pins 5,6,8, and 20 should be connected together on the connector at the IBM end.

```
10 'EXAMPLE PROGRAM TO READ THE SR400 COUNTER DATA AND RAMP THE PORT 1
20 'ANALOG OUTPUT USING IBM BASIC AND THE COM1: RS232 PORT.
30 '
40 '
50 'SET THE COM MENU OF THE SR400 FOR RS232 BAUD=9600
60 '                RS232 BITS=8
70 '                RS232 PARITY=NONE
80 '                RS232 WAIT=6
90 '                RS232 ECHO=OFF
100 '
110 '
120 OPEN "COM1:9600,N,8,2,CS,DS,CD" AS #1
130 '
140 'SETUP COM1: PORT TO 9600 BAUD, NO PARITY, 8 DATA BITS, 2 STOP BITS,
150 'IGNORE CTS (CLEAR TO SEND), DSR (DATA SET READY),
160 'AND CD (CARRIER DETECT).
170 '
180 PRINT #1,"      " 'CLEAR COM1: BY SENDING SPACES
190 PRINT #1,"RC0" 'RESET SR400
200 FOR I=1 TO 500: NEXT I 'WAIT FOR RESET TO FINISH
210 X=0 'INIT PORT1 VALUE TO ZERO
220 PRINT #1,"CI0,0" 'SET COUNTER A INPUT TO 10 MHZ
230 '
240 '
250 PRINT #1,"CR;CS" 'RESET COUNTER;START NEW COUNT PERIOD
260 '
270 PRINT #1,"QA1" 'READ COUNTER A VALUE
280 INPUT #1,N1 'INTO VARIABLE N1
290 IF N1=-1 THEN GOTO 270 'IF COUNT PERIOD NOT DONE, LOOP AND TRY AGAIN
300 PRINT "COUNTER A = "; N1; 'ELSE IF COUNT FINISHED, PRINT RESULT
310 '
320 X = X + .005 'INCREMENT PORT1 OUTPUT BY 5 MV
330 IF X>10 THEN X=0 'RESET RAMP IF AT 10 VOLTS
340 PRINT #1,USING "PL1,##.###"; X 'SET PORT1 OUTPUT VOLTAGE
350 PRINT "PORT1 = "; X 'PRINT PORT1 VALUE
360 '
370 GOTO 250 'LOOP FOREVER
```

PROGRAM EXAMPLE 2: IBM PC, MICROSOFT FORTRAN V4.0, VIA RS232

Machine language routines to interface to the COM1: RS232 port are provided in the file RS232.OBJ found on the SR400 demo disk. These routines allow for simple interfacing to the SR400 at 19.2 kbaud from FORTRAN programs.

To use these routines, the file 'for232.inc' (also on the demo disk) must be 'included' in the FORTRAN source.

Only two wires between the IBM PC's COM1: port and the SR400 are needed (pins #2 & #3 of the 25 pin RS232 connector), but pins 5,6,8, and 20 should be connected together on the connector at the IBM end.

```
$include:'for232.inc'
```

```
c      for232.inc must be included to call subroutines in RS232.OBJ
      link with RS232.OBJ (on demo disk)
```

```
c      RS232.OBJ defines:
```

```
c      init
c      initializes COM1: to 19.2 kbaud
```

```
c      txstr (str)
c      transmits string str to COM1:
c      str is a string terminated with '$'
```

```
c      rxstr (str)
c      fills str with string received from COM1:
c      str must be declared with length of 15 or greater
```

```
c      If an error occurs, nocom is called.
c      Nocom should be a FORTRAN subroutine in your program.
```

```
      program ex5
```

```
      character*20 str1,str2
```

```
c      Example program to read the SR400 counter data and ramp the PORT1
c      analog output using Microsoft FORTRAN v4.0 and the COM1: port.
```

```
c      Set the SR400 to 19200 baud, RS232 ECHO=OFF,RS232 BITS=8,PARITY=NONE
c      initialize COM1: port to 19.2 kbaud
c      call init
```

```
c      reset the SR400
c      call txstr('RC0$')
```

```
c      set character wait interval to zero and counter A to 10 MHz
c      call txstr('C10,0;SW0$')
```

```
c      reset PORT1 value to zero
c      x=0.0
```

SR400 with the RS232 Interface

```
c      reset and start new count period
10     call txstr('CR;CS$')

c      read counter A data into string str1
20     call txstr('QA1$')
       call rxstr(str1)

c      convert into integer variable n1
       read(str1,1000) n1
1000   format (bn,i15)

c      if data not ready yet, loop and try again
       if (n1.lt.0) goto 20

c      ramp PORT1 by 5 mV
       x = x + .005
       if (x.gt.10) x = 0.0
c      make PORT1 command string
       write(str2,2000) x
2000   format('PL1,',f7.3,'$')
       call txstr(str2)

c      print data
       write(*,3000) n1,x
3000   format(' Counter A = ',l15,'PORT1 = ',F7.3)

c      and loop forever
       goto 10

       end

c      *****
       subroutine nocom

c      in case of timeout error, this routine runs.
c      put your error handler here.

       write (*,*) char(7)
       write (*,*) ' RS232 Error'

       stop
       end
```

PROGRAM EXAMPLE 3: IBM PC, MICROSOFT C V4.0, VIA RS232

Machine language routines to interface to the COM1: RS232 port are provided in the file RS232.OBJ found on the SR400 demo disk. These routines allow for simple interfacing to the SR400 at 19.2 kbaud from C programs.

To use these routines, the large model must be used. Compile with the /AL switch and link with RS232.OBJ.

Only two wires between the IBM PC's COM1: port and the SR400 are needed (pins #2 & #3 of the 25 pin RS232 connector), but pins 5,6,8, and 20 should be connected together on the connector at the IBM end.

```
#include <stdio.h>
```

```
/* Compile with >MSC program name/AL;
   link with RS232.OBJ (on demo disk)
```

```
RS232.OBJ defines:
```

```
init           initializes COM1: to 19200 baud
```

```
txstr(str)     sends string str to COM1:
char *str;     str must terminate with '$' char
```

```
rxstr(str)     fills str with string received from COM1:
char *str;     str must be declared with a length of greater than 15
```

```
If an error occurs, your procedure nocom() is called.
nocom() must be a C procedure in your program.
```

Example program to read the SR400 counter data and ramp the PORT1 analog output using Microsoft C v4.0 (large model) and the COM1: port.

Set the SR400 for 19200 baud, ECHO=OFF,BITS=8,PARITY=NONE

```
*/
```

```
main()
```

```
{
char result[20], portstr[20];
long n1;
float x;
```

```
init();           /* init COM1: port to 19.2 kbaud */
txstr("rc0$");    /* reset SR400 */
txstr("ci0,0;sw0$"); /* counter A to 10 MHz */
                  /* set SR400 RS232 wait interval to zero */
```

```
x = 0;           /* init PORT1 value */
```

```

while (1)
{
    txstr("cr;cs$"); /* reset and start new count period */

    do
    {
        txstr("qa1$"); /* read counter a data */
        rxstr(result); /* into string result */

        /* scan result for long integer */
        if (!(sscanf(result, "%D", &n1))) nocom();
    }

    while (n1<0);

    x += 0.005; /* increment PORT1 level */
    if (x>10) x=0; /* reset if at 10 Volts */
    sprintf(portstr, "pl1,%f$",x); /* make PORT1 command string
    txstr(portstr); /* send command */

    /* print answer */
    printf ("Counter A = %ld   PORT1 = %f\n",n1,x);
}

/* ***** */
nocom() /* error routine goes here */
{
    printf("RS232 Error\n");
    putch(7);
    exit();
}

```


SR400 WITH THE GPIB INTERFACE

Before using the GPIB interface you must set the GPIB address in the COM menu.

GPIB CAPABILITIES

The GPIB capabilities of the SR400 consistent with IEEE standard 488 (1978) are shown in the table below. Also shown are the responses of the SR400 to some standard commands.

<u>Code</u>	<u>Function</u>
SH1	Source handshake capability
AH1	Acceptor handshake capability
T5	Basic Talker, Serial Poll, Unaddressed to talk if addressed to listen
L4	Basic Listener, Unaddressed to listen if addressed to talk
SR1	Service request capability
PP0	No parallel poll capability
DC1	Device Clear capability
RL1	REN,LLO, GTL implemented
DT1	GET implemented

SR400 Response to GPIB Commands

<u>Mnemonic</u>	<u>Command Response</u>
DCL	Device Clear Same as CL command
SDC	Selected Device Clear Same as CL command
SPE	Serial Poll Enable Send Status Byte after service request
GET	Group Execute Trigger Same as CS command

The SR400 will respond to commands only if it is addressed to listen and the Remote Enable line (REN) is asserted. When this occurs, the front panel goes to the REMote state; the knob and all of the keys except for the COM/LOCAL key are disabled. The COM/LOCAL key will return the unit to the LOCAL state. The COM/LOCAL key may be locked out by asserting the Local-Lockout state (LLO).

SERIAL POLLS AND SERVICE REQUESTS

The SR400 can be programmed to generate a service request (SRQ) to the GPIB controller every time a given status condition occurs. This is done using the SV{j} command. The mask value, j (0-255), is the SRQ mask byte. The mask byte is always logically AND'ed with the status byte. If the result is non-zero, the SR400 generates an SRQ and leaves the status byte unchanged until the controller performs a serial poll to determine the cause of the service request. For example, if we want to generate an SRQ at the end of a scan, we need an SRQ mask byte equal to 00000100 binary, or 4 decimal ("SV4" command). The byte 00000100 binary corresponds to the status byte with the 'scan finished' status bit set. When the next end of scan occurs, an SRQ will be generated. The serial poll will return a status byte showing SRQ and scan finished (01000100 binary or 68 decimal). When a service request is generated, the corresponding bit in the SRQ mask byte is reset. This will prevent an uncontrolled stream of service requests from continually interrupting the controller. To re-enable the service request, the controller must again set the service request mask with the SV command.

Serial Polls performed without a service request return all status conditions which have occurred since the last time the status byte was cleared.

Note that the serial poll does not clear the status byte. To clear the status byte use the SS command.

EXAMPLE PROGRAMS

The following section lists several example programs which use the GPIB port to control the SR400. All of the programs do the same thing, only the language is different. The programs set up the SR400 to count 10 MHz on counter A for 1 second. The counter data is read and the PORT1 output is ramped.

**PROGRAM EXAMPLE 4:
IBM PC, MICROSOFT BASIC, VIA GPIB**

This program requires the Capital Equipment Corporation GPIB card for the IBM PC. It has firmware in ROM to interface high level languages to the GPIB.

Subroutine calls in Microsoft BASIC are done to memory locations specified by the name of the

subroutine. The address is relative to the segment address specified by the DEF SEG statement preceding the CALL.

In this program, the CEC card's ROM starts at C000H, the system controller's address is 21, and the SR400 is at GPIB address 23.

```

10 'EXAMPLE PROGRAM TO READ THE SR400 COUNTER DATA AND RAMP THE PORT1
20 'ANALOG OUTPUT USING IBM PC BASIC AND THE CAPITAL EQUIPMENT CORP.
30 'GPIB INTERFACE CARD
40 '
50 '
60 'SET THE COM MENU GPIB ADDR=23
70 '
80 DEF SEG = &HC000                                'BASE ADDRESS OF CEC CARD
90 INIT=0: TRANSMIT=3: SEND=9: ENTER=21            'ADDRESSES OF CEC FIRMWARE ROUTINES
100 ADDR%=21: SYS%=0                                'CONTROLLER ADDRESS
110 SR400%=23                                        'SR400 ADDRESS
120 '
130 'STRING DEFINITIONS
140 IN$="IFC UNT UNL REN MTA LISTEN 23 SDC"        'CLEAR SR400
150 CI$="CI0,0"                                     'COUNTER A INPUT=10 MHZ
160 CR$="CR;CS"                                     'RESET AND START NEW COUNT
170 QA$="QA1"                                       'READ COUNTER A DATA
180 P1$="PL1,"                                     'SET PORT1 LEVEL
190 '
200 '
210 CALL INIT(ADDR%,SYS%)                           'INIT CEC CARD
220 CALL TRANSMIT(IN$,STATUS%): GOSUB 480           'RESET SR400
230 FOR I=1 TO 500: NEXT I                          'WAIT FOR RESET TO FINISH
240 '
250 X=0                                              'INIT PORT1 VALUE
260 '
270 CALL SEND(SR400%,CI$,STATUS%): GOSUB 480        'SET COUNTER A INPUT
280 '
290 'START NEXT COUNT PERIOD
300 CALL SEND(SR400%,CR$,STATUS%): GOSUB 480
310 '
320 CALL SEND(SR400%,QA$,STATUS%): GOSUB 480        'READ DATA
330 ANS$=SPACE$(20)                                'CLEAR RESULT STRING
340 CALL ENTER(ANS$,LENGTH%,SR400%,STATUS%): GOSUB 480 'READ COUNTER DATA
350 N1=VAL(ANS$)                                    'INTO VARIABLE N1
360 IF N1<0 THEN GOTO 320                          'IF COUNT PERIOD NOT DONE, TRY AGAIN
370 PRINT "COUNTER A = "; N1;                      'ELSE PRINT VALUE
380 '
390 X = X+.005                                       'INCREMENT PORT1 VALUE
400 IF X>10 THEN X=0                                'RESET PORT1 AT 10 VOLTS
410 PRINT " PORT1 = "; X

```

```
420 P$=P1$ + STR$(X)           'MAKE PORT1 COMMAND STRING
430 CALL SEND(SR400%,P$,STATUS%):GOSUB 480
440 '
450 GOTO 300                     'LOOP FOREVER
460 '
470 '
480 'CHECK STATUS OF LAST GPIB COMMUNICATION FOR ERRORS
490 IF STATUS%=0 THEN RETURN      'STATUS OK
500 PRINT "GPIB ERROR STATUS CODE = ";STATUS% 'ELSE ERROR
510 STOP
```

**PROGRAM EXAMPLE 5:
IBM PC, MICROSOFT C V4.0, VIA GPIB**

This program requires the Capital Equipment Corporation GPIB card for the IBM PC. It has firmware in ROM to interface high level languages to the GPIB.

The header file MS-C488.H contains definitions of all of the CEC firmware routines which can be called from C. Note that the order of arguments in

the calls are reversed from those used by BASIC.

To use these routines, the large model must be used. Compile with the /AL/Ze switch and link with GPIB-L.OBJ (on the CEC disk).

```
/* Compile with >MSC program name/AL/Ze;
   link with GPIB-L.OBJ (on CEC disk)
   include MS-C488.H
```

MS-C488.H defines all of the CEC firmware routines so that they can be called from a C program. Remember that the order of the arguments is REVERSED from the BASIC calls described in the manual.

Example program to read the SR400 counter data and ramp the PORT1 analog output using Microsoft C v4.0 (large model) and the Capital Equipment Corp. GPIB interface card.

Set the SR400 for GPIB ADDR=23.

```
*/
```

```
#include <ms-c488.h>
#include <stdio.h>
```

```
char result[80];
```

```
main()
```

```
{
  int i,status,sr400=23;
  int my_address=21, system_controller=0;
  char cmd[80], portstr[40];
  long n1;
  float x;
```

```
  initialize(&system_controller, &my_address);      /* init GPIB controller */
```

```
  /* clear SR400 */
  transmit(&status, "IFC UNT UNL REN MTA LISTEN 23 SDC");
  Statcheck(status);
  for (i=0;i<1000;i++);      /* wait for clear to finish */
```

```
  /* set counter A to 10 Mhz */
  TxGpib(sr400,"C10,0");
```

```
  x = 0;      /* init PORT1 value */
```

```

while (1)
{
    TxGpib(sr400,"CR;CS");      /* reset and start new count period */

    do
    {
        TxGpib(sr400,"QA1");      /* read counter a data */
        GetGpib(sr400);           /* into result string */

        /* scan result for long integer */
        if (!(sscanf(result, "%D", &n1)))
            Statcheck(255);
    }

    while (n1<0);                /* until count period done */
    x += 0.005;                  /* increment PORT1 level */
    if (x>10) x=0;               /* reset if at 10 Volts */
    sprintf(portstr, "PL1,%6.3f",x); /* make PORT1 command string */
    TxGpib(sr400,portstr);        /* send command */

    /* print answer */
    printf ("Counter A = %ld  PORT1 = %f\n",n1,x);
}

/* ***** */
TxGpib(address,command)      /* transmit a command to addressed device */
int address;
char *command;
{
    int status;
    char t_string[80];

    sprintf(t_string, "UNT UNL MTA LISTEN %d DATA '%s' 10",address,command);
    transmit(&status, t_string);
    Statcheck(status);
}

/* ***** */
GetGpib(address)      /* get result string from addressed device */
int address;
{
    char r_string[40], temp[80];
    int length, status;

    sprintf(r_string, "UNT UNL MLA TALK %d",address);
    transmit(&status, r_string);      /* make device a talker */
    Statcheck(status);
    strcpy(temp, "");                /* clear result string */
    receive(&status, &length, temp); /* get answer */
    Statcheck(status);
    strcpy(result, temp);            /* move answer into global result string */
}

```

```
/* ***** */
Statcheck(status)      /* check GPIB status */
int status;
{
    if (status)
    {
        printf("GPIB error: status = %d",status);
        exit();
    }
}
```


TROUBLESHOOTING

To start, make sure that the power entry module on the rear panel is set for the ac line voltage in your area and that the correct fuse is installed. The selected ac voltage may be seen through the window on the power entry module. Verify that the line cord is plugged all the way into the power entry module and that the power button on the front panel is pressed in.

LCD CONTRAST

If there are no characters on the LCD, or the contrast is very poor, go to the LCD CONTRAST line of the SETUP menu and use the spin knob to adjust the contrast. Larger values increase the contrast.

FAN

The fan in the SR400 is temperature controlled. When the unit is cold, the fan is off. When the internal temperature reaches about 30° C, the fan starts to blow. Do not block the vents in the chassis or the unit may not operate properly.

FRONT PANEL TEST

The keyboard, knob, display, and LED indicators may be tested in the front panel test mode. Turn the unit on while holding down the COM key to enter the front panel test. The LCD should display the message FRONT PANEL TEST. Pressing a key will cause the LCD to display the name of the key on both lines of the display. If the keys do not work properly, check to make sure that there are no keys stuck in the down position. Turning the knob will light each LED one at a time in succession across the front panel.

POWER ON RESET

If the instrument turns on with odd combinations of LED's on, a garbled message on the LCD, or is unresponsive to the keyboard or knob, then the memory contents may have been corrupted causing the instrument to "hang". To remedy the situation, turn the unit off and hold down the STOP key while turning the unit back on. This causes the unit to reinitialize the memory and assume the default setup.

QUICK TEST

Unplug all cables from the unit and recall the default setup by holding the STOP key down while turning the unit on. The message RECALL DEFAULT should appear on the display for ~3 s. Press the DOWN key once to go to the A=INPUT 1 menu line. Rotate the knob until A=10 MHz appears on the menu line. To start a count period, press the START key once. The data display should show counting in counter A for 1 second and stop at A= 10000000. Counter A will have counted the internal 10 MHz timebase for 1 second and reached 10,000,000 counts. To reset the counter, press the STOP key. The data display should reset to 0.

Now press the DOWN key 3 more times to display the N PERIODS line. Rotate the knob until N PERIODS=10. Start a scan by pressing the START key. The counter should cycle through 10 count periods separated by the default dwell time (1 s). The display shows the current scan position in brackets. Reset the scan using the STOP key.

Start another scan with the START key but press the STOP key before all 10 count periods are completed. This pauses the scan at the last completed count period. START will resume the scan and STOP will reset the scan.

In general, if the START key doesn't start a count period it is because the counters are paused at the end of a scan. In this case, the SCAN led will be off and the N PERIODS display shows that the programmed number of points has been taken. A STOP is required to reset the scan.

Pressing the STOP key twice in succession always resets the counters.

COUNTING

There are several reasons why a counter may not seem to be counting correctly.

1) Counting is not enabled. The COUNT led does not turn on when the START key is pressed. This is because the counters need to be reset. Use the STOP key to reset the counters. Verify that the preset count period is long enough so that a non-

zero count can occur.

2)The counter input is set to the wrong source or input. Check the counter input menu line and the counter discriminator output for the correct signal. If the preset counter has no input, then the count period will never terminate.

3)The counter is gated. If the count led is on and the source is correct but no counting occurs, the counter gate may be disabling the counter. Check that the gate is in CW mode if no gating is needed or, if gating is required, that the trigger is connected and discriminating correctly. The gate may be viewed through the GATE output. The output is low when the gate enables the counter.

4)Incorrect gate position. When nanosecond gate delays or widths are used, timing accuracy is critical. Gate timing should be verified by viewing the DISC and GATE outputs on an oscilloscope. Be sure to use equal length cables so as not to introduce timing skew. The timing accuracy of the outputs from the SR400 is 2 ns.

5)Bad gate triggering. The trigger input may not be discriminated correctly. Check the trigger slope and level. The discriminated trigger pulses may be viewed through the T DISC output when the input

to counter T is TRIG. Double triggering or the wrong slope can cause erroneous gates.

6)Incorrect discriminator levels or slopes. Verify that the discriminator settings are correct. Use the DISC output to check the discriminator output pulses on an oscilloscope.

7)Signal inputs are overloaded. Check that the inputs are $\leq \pm 300$ mV or, if the signal exceeds 300 mV, that the overloads do not last longer than 10 μ s. If the overload lasts longer than 10 μ s, the input amplifiers will not recover for about 200 ns.

8)Discriminators are inhibited. Check that the INHIBIT input on the rear panel is not held at TTL high. Leave the INHIBIT input unconnected when not in use.

9)EXT STOP is resetting the counters. If the DWELL TIME is programmed (i.e. not EXTERNAL) an EXT STOP will reset the counters. In general, EXT STOP should be disconnected when the DWELL TIME is **NOT** EXTERNAL.

10)The DWELL TIME is EXTERNAL and EXT START and STOP are disconnected or connected to the wrong signal sources.

CALIBRATION AND REPAIR

Most of the calibration parameters are determined by a computer aided calibration procedure after burn-in at the factory. These calibration parameters are stored in the permanent memory of each unit.

This section details the calibration of the signal inputs which, because of their high speed, require some hand adjustment. In addition, the repair procedure for the signal inputs is discussed.

**** WARNING ****

The calibration procedure requires adjusting the instrument with power applied and so there is a risk of personal injury or death by electric shock. Do not remove the covers from the shielded power supply when the power is on!

Please be careful.

SIGNAL INPUT OFFSET

To calibrate the input offset, connect a voltmeter to a signal input. P101 adjusts INPUT 1 and P121 adjusts INPUT 2. These potentiometers are located behind the signal inputs. Adjust the correct potentiometer for a reading of $0 \pm .1$ mV on the meter.

SIGNAL INPUT REPAIR

The diodes in the overload protection circuit can be damaged by excessive voltages at the inputs. To check for damage, measure for 0.6 V across both diodes in D101 and D121. D102, D103, D122, and D123 can be tested by removing them from the circuit and then checking for degradation of the amplifiers pulse response as each diode is reconnected. The amplifier outputs can be viewed with a fast oscilloscope across R201 and R212. Replacement diodes are available from the factory.

Performing these adjustments or replacements can result in the discriminator levels being slightly miscalibrated. The discriminator offsets may be a few mV larger after repair or adjustment.

USING PHOTOMULTIPLIER TUBES

PHOTOMULTIPLIER SELECTION FOR PHOTON COUNTING

Photomultiplier Tubes (PMT's) are high-gain, low noise light detectors. They can detect single photons over a spectral range of 180 to 900 nm. Windowless PMT's can be used from the near UV through the X ray region, and may also be used as particle detectors.

Photons which strike the PMT's photocathode eject an electron by the photoelectric effect. This electron is accelerated toward the first dynode by a potential of 100 to 400 Vdc. Secondary electrons are ejected when the electron strikes the first dynode, and these electrons are accelerated toward the second dynode. The process continues, typically for 10 dynodes, each providing an electron gain of about 4, to produce 1,000,000 electrons which are collected by the anode. If these electrons arrive in a 5 ns pulse into a 50 Ohm load, they will produce a 1.6 mV pulse. These pulses may be amplified and counted.

GEOMETRY

There are two basic geometries for photomultiplier tubes: head-on and side-on types. The head-on type has a semitransparent photocathode, and a linear array of dynodes. The head-on types offer large photocathodes with uniform sensitivity, and lower noise. These PMT's must be operated at a higher voltage, and are usually larger and more expensive than the side-on types. Side-on types have an opaque photocathode and a circular cage of dynodes.

SPECTRAL RESPONSE

There are a variety of materials which are used as photocathodes: the workfunction of the photocathode will determine the spectral response (and will influence the dark count rate) of the PMT. For photon counting, the figure of merit is the "quantum efficiency" of the PMT. A 10% quantum efficiency indicates that 1 in 10 photons which strike the photocathode will produce a photoelectron -- the rest of the incident photons will not be detected. The quantum efficiency is a function of wavelength, so select the PMT for the best quantum efficiency over the wavelength region of interest.

GAIN AND RISETIME

It is important to select a PMT with sufficient gain, and short enough risetime, to produce a detectable pulse for counting. In addition, the risetime is an important figure of merit to determine the maximum count rate for the tube.

The criteria for a "detectable pulse" depends on the electrical noise environment of your laboratory, and the noise your preamplifier. In laboratories with Q-switched lasers or pulsed discharges, it is difficult to reduce the noise on any coax cable below a few millivolts. A good, wide bandwidth preamplifier (such as the SR400) will have about 1.5 nV per root Hertz, or about 25 uV rms over its 300 MHz bandwidth. Peak noise will be about 2.5 times the rms noise, and so it is important that the PMT provide pulses of at least 100 uV amplitude.

Use manufacturer's specifications for the current gain and risetime to estimate the pulse amplitude from the PMT:

$$\text{Amplitude (mV)} = 4 \times \text{Gain (in millions)} / \text{Risetime (in ns)}$$

This formula assumes that the electrons will enter a 50 Ohm load in a square pulse whose duration is twice the risetime. (Since the risetime will be limited to 1.2 ns by the 300 MHz bandwidth of the preamplifier, do not use risetimes less than 1.5 ns in this formula.)

The current gain of a PMT is a strong function of the high voltage applied to the PMT. Very often, PMT's will be operated well above the high voltage recommended by the manufacturer, and so at substantially higher current gains (10x to 100x above specs). There are usually no detrimental affects to the PMT so long as the anode currents are kept well below their rated values.

Conclusions: Select a PMT with a risetime < 3 ns and a current gain > 5 million.

DARK COUNTS

PMT's are the quietest detectors available. The primary noise source is thermionic emission of electrons from the photocathode and from the first few dynodes of the electron multiplier. PMT housings

which cool the PMT to about -20°C can dramatically reduce the dark count (from a few kHz to a few Hz). The residual counts arise from radioactive decays of materials inside the PMT and from cosmic rays.

PMT's which are specifically designed for photon counting will specify their noise in terms of the rate of output pulses whose amplitudes exceed some fraction of a pulse from a single photon. More often, the noise is specified as an anode dark current. Assuming the primary source of dark current is thermionic emission from the photocathode, the dark count rate is given by :

$$\text{Dark Count (in kHz)} = 6 \times \text{Dark Current (in nA)} / \text{Gain (in millions)}$$

PMT BASE DESIGN

PMT bases which are designed for general purpose applications are not appropriate for photon counting. General purpose bases will not allow high count rates, and often cause problems such as double counting and poor plateau characteristics. A PMT base with the proper high voltage taper, bypassing, snubbing, and shielding is required for premium photon counting performance.

CAUTION: Lethal High Voltages are used in PMT applications. Use extreme caution when working with these devices. Only those experienced with high voltage circuits should attempt any of these procedures. Never work alone.

DYNODE BIASING

A PMT base provides bias voltages to the PMT's photocathode and dynodes from a single negative high voltage power supply. The simplest design

consists of a resistive voltage divider, as shown in Figure A.

In this configuration the voltage between each dynode, and so the current gain at each dynode, is the same. Typical current gains are 3 to 4, and so there will typically be 3 to 4 electrons leaving the first dynode, with a variance of about 2 electrons. This large relative variance (due to the small number of ejected electrons) gives rise to a large variations in the pulse height of the detected signal. Since statistical fluctuations in pulse height are caused by the low gain of the first few stages of the multiplier chain, increasing the gain of these stages will reduce pulse height variations and so improve the plateau characteristics of the PMT. To increase the gain of the first few stages, the resistor values in the bias chain are tapered to increase the voltage in the front end of the multiplier chain. The resistor values are slowly tapered so that the electrostatic focusing of electrons in the multiplier chain is not adversely affected.

Current for the electron multiplier is provided by the bias network. Current drawn from the bias network will cause the dynode potentials to change, and so change the tube gain. This problem is of special concern in lifetime measurements. The shape of exponential decay curves will be changed if the tube gain varies with count rate. To be certain that this is not a problem, repeat the measurement at half the original intensity.

The problem of gain variation with count rate is avoided if the current in the bias network is about 20 times the output current from the PMT's anode.

Example: If a PMT is operated so that it gives 20 mV pulses of 5 ns duration into a 50 Ohm cable, then the average current at 50 MHz count rate will

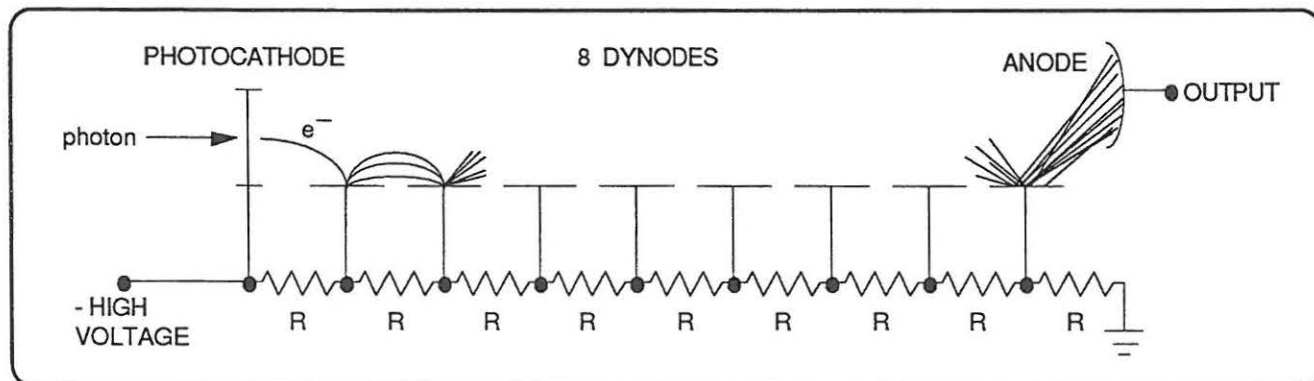


FIGURE A: RESISTIVE DIVIDER PMT BASE

be 0.1 mA. If the bias resistors are chosen so that the chain current is $20 \times 0.1 \text{ mA} = 2 \text{ mA}$, then the PMT's gain will remain constant vs. count rate. If this PMT is operated at 2500 vdc, then the power dissipated in this base is 5 Watts.

There are a few other methods to avoid this problem which do not require high bias currents. These methods depend on the fact that the majority of the output current is drawn from the last few dynodes of the multiplier.

(1) Replace the last few resistors in the bias chain with Zener Diodes. As long as there is some reverse current through a Zener, the voltage across the diodes is nearly constant. This will prevent the voltage on these stages from dropping as the output current is increased.

(2) Use external power supplies for the last few dynodes in the multiplier chain. This approach dissipates the least amount of electrical power since the majority of the output current comes from lower voltage power supplies, however it is the most difficult to implement.

(3) If the average count rate is low, but the peak count rate is high, then bypass capacitors on the last few stages may be used to prevent the dynode voltage from dropping. (Use $20\times$ the average output current for the chain current.) For a voltage drop of less than 1%, the stored charge on the last bypass capacitor should be $100\times$ the charge output during the peak count rate. For example, the charge output during a 1 ms burst of a 100

MHz count rate, each with an amplitude of 10 mV into 50 Ohms and a pulse width of 5 ns, is 0.1 uC . If the voltage on the last dynode is 200 Vdc, then the bypass capacitor for the last dynode should have a value given by:

$$C = 100 \text{ Q/V} = 100 \times 0.1 \text{ uC} / 200 \text{ V} = 0.05 \text{ uF}$$

The current from higher dynodes is smaller so the capacitors bypassing these stages may be smaller. Only the final four or five dynodes need to be bypassed, usually with a capacitor which has half the capacitance of the preceding stage. To reduce the voltage requirement for these capacitors, they are usually connected in series. (See Figure B)

By passing the dynodes of a PMT may cause high frequency ringing of the anode output signal. This can cause multiple counts for a single photon. The problem is significantly reduced by using small resistors between the dynodes and the bypass capacitors, as shown in Figure B.

SNUBBING

Snubbing refers to the practice of adding a network to the anode of the PMT to improve the shape of the output pulse for photon counting applications. This 'network' is usually a short piece of 50 Ohm coax cable which is terminated into a resistor of less than 50 Ohms. There are four important reasons for using a snubber network:

(1) Without some dc resistive path between the anode and ground, anode current will charge the

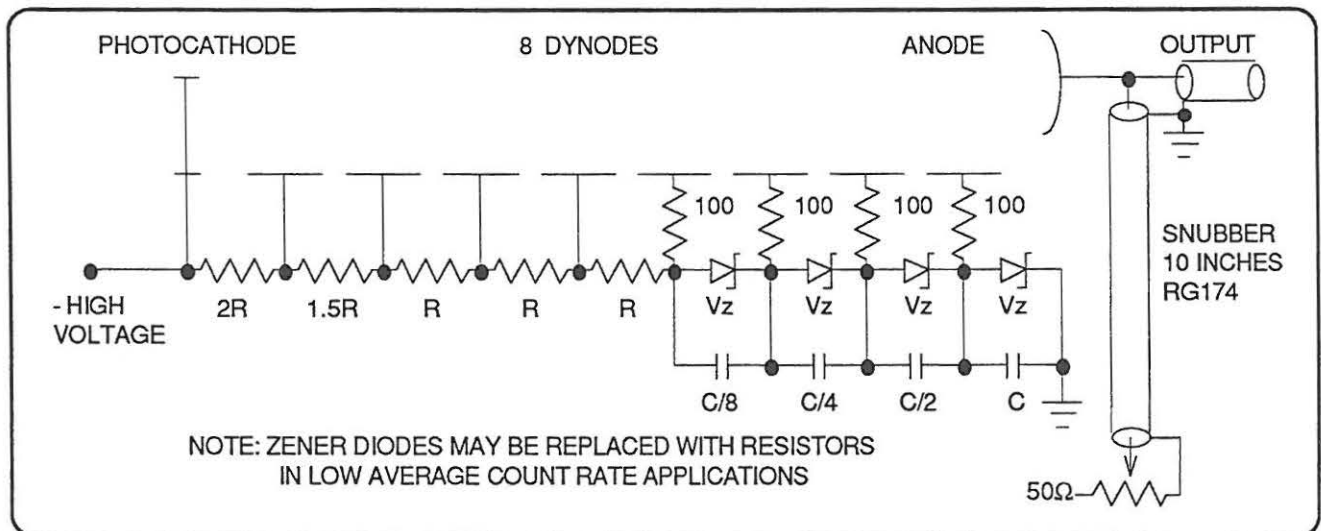


FIGURE B: PMT BASE FOR PHOTON COUNTING

signal cable to a few hundred volts. When the signal cable is connected to a preamplifier or to a photon counter, the stored charge on the cable may damage the front-end of the instrument. If you decide not to use a snubber network, please install a 100 kOhm resistor between the anode and ground to protect your instruments.

(2) The risetime of the output current pulse is often much faster than the falltime. A snubber network may be used to sharply reduce the falltime, greatly improving the pulse pair resolution of the PMT.

(3) Ringing (with a few nanosecond period) is very common on PMT outputs (especially if the final dynode stages are bypassed with capacitors). A snubber network may be used to cancel these rings which can cause multiple counts from a single photon.

(4) The snubber network will help to terminate reflections from the input to the preamplifier.

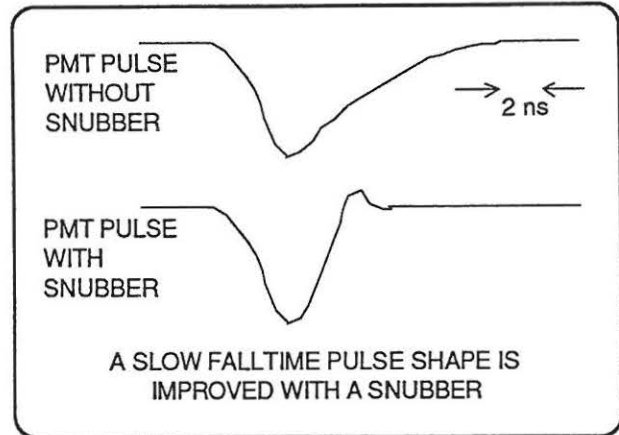
A good starting point for a snubber network is a 10 inch piece of RG174/U coax cable with a small 50 Ohm pot connected to the end so that the terminating impedance may be adjusted from 0 to 50 Ohms. (A 10 inch cable will have a round trip time of about 5 ns -- be sure your PMT has a risetime less than this.) The other end of this cable is connected to the anode of the PMT, together with the output signal cable.

Output current pulses will split, 50% going out the signal cable, and 50% going into the snubber. If the snubber pot is adjusted to 50 Ohms there will be no reflection -- the only affect the snubber has is to attenuate the signal by a factor of two.

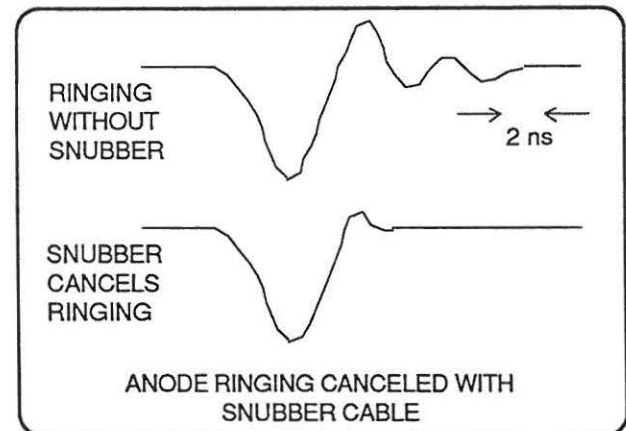
The reflection coefficient for a cable with a characteristic impedance R_0 , terminated into a resistance R_t , is given by:

$$\text{Reflection Coefficient} = (R_t - R_0) / (R_t + R_0)$$

If the pot is adjusted to a value below 50 Ohms, then some portion of the signal will be inverted and reflected back toward the anode. This reflected (and inverted) signal is delayed by the round trip time in the snubber cable and sent out the signal cable. The amount of the reflection is adjusted for the best pulse shape as shown in the figure below.



The round trip time in the snubber cable may be adjusted so that the reflected signal cancels anode signal ringing. This is done by using a cable length with a round trip time equal to the period of the anode ringing.



PMT BASE CONCLUSIONS:

- (1) Taper voltage divider for higher gain in first stages.
- (2) Bypass last few dynodes in pulsed applications.
- (3) Use a snubber circuit to shape the output pulse.

CATHODE SHIELDING

Head-on PMT's have a semitransparent photocathode which is operated at negative high voltage. Use care so that no objects near ground potential contact the PMT near the photocathode.

MAGNETIC SHIELDING

Electron trajectories inside the PMT will be affected by magnetic fields. A field strength of a few Gauss can dramatically reduce the gain of a PMT. A magnetic shield made of a high permeability material should be used to shield the PMT.

PREAMPLIFIERS

The output of a PMT is a current pulse. This current is converted to a voltage by a load resistor. One would like to use a large resistor to get a large voltage pulse, however in photon counting it is important to maintain a high bandwidth for the output signal. Since charge on the anode is removed by the load resistance, smaller load resistances increase the bandwidth. The bandwidth of a 10 pF anode with a 100 Ohm load is 300 MHz.

For convenience, 50 Ohm systems are usually used. The current pulse from the PMT travels down a 50 Ohm cable which is terminated by the 50 Ohm input impedance of a preamplifier. The attenuation of RG-58 coax cable at 300 MHz is about 1 dB/ 10 ft. and so it does not significantly degrade performance in this application.

To allow counting to 200 MHz, a preamplifier with a bandwidth which is somewhat larger than 200 MHz is required. The SR440 preamplifier has four gain of 5 amplifiers, each with 50 Ohm input impedance and a 300 MHz bandwidth. The amplifiers may be cascaded for gains of 5, 25, 125, or 625.

GAIN REQUIREMENT

The SR400 Photon Counter can detect pulses as low as 2 mV. To allow for some adjustment of the discriminator threshold and to provide better noise immunity, a more practical lower limit on pulse size is about 10 mV. The highest discriminator level which may be set is 300 mV. The preamplifier should have enough gain to amplify anode pulses to between 10 mV and 300 mV (100 mV is a good target value).

Using the result that pulse height (in mV) is about 4x the tube gain (in millions) divided by the risetime (in ns), a PMT with a gain of 4 million and a risetime of 2 ns will provide 8 mV output pulses. Half of the pulse amplitude will be lost in the anode snubber, so a gain of 25 is required to boost the output pulses to 100 mV amplitude.

PLATEAUING A PMT

Due to the statistical nature of the secondary emission process, there is a distribution of signal pulse heights coming from the PMT. There is another distribution of noise pulse heights. Noise which results from thermionic emission from the photocathode can not be distinguished from signal, however, noise pulses from dynode thermionic emission will have a lower mean pulse height. The PMT should be operated at sufficient high voltage that the mean signal pulse height is well above the pulse height of other noise sources such as preamp noise and EMI pickup.

There are two reasons for carefully selecting the input discriminator level. The first is to improve the signal-to-noise ratio by setting the discriminator level above most of the noise pulses, but below most of the signal pulses. The second reason is to reduce drift: if the discriminator threshold is set to the middle of the signal pulse height distribution, then small changes in the tube gain can cause a large change in the count rate.

There is no exact prescription for setting the discriminator threshold: the procedure used will depend somewhat on the nature of the measurement. If dark counts are a problem then the discriminator level should be set higher than when drift is a concern. If the PMT is cooled (reducing thermionic emission) then a lower discriminator level is probably okay. If the PMT has a ring on the anode signal then the discriminator level should be set high enough so that the rings are not counted.

THE 'CORRECT' WAY

The tube should be operated at the maximum high voltage recommended by the manufacturer. Use enough preamplifier gain so that the single photon pulse height is about 100 mV. Provide enough light to the PMT for a count rate of a few megahertz. Using a 300 MHz oscilloscope, adjust the snubber termination for minimum ringing on the anode signal. Take the pulse-height spectrum of the anode signal by scanning both input discriminators and plotting A-B. (See the SETUP section of this manual for exact details.) If the PMT dark count rate is a concern, then you will also need to take the pulse height spectrum of the dark count signal. It will take much longer to take the dark count spectrum because the count rate should be much lower. The object is to find a discriminator

level which is higher than the mean noise pulse height, and below the mean signal pulse height.

THE 'FAST AND PRETTY GOOD' WAY

This technique works very well and is particularly suited for those who do not want to make a career out of plateauing their PMT's. The PMT should be operated at (or a bit above) the recommended maximum high voltage. Provide enough illumina-

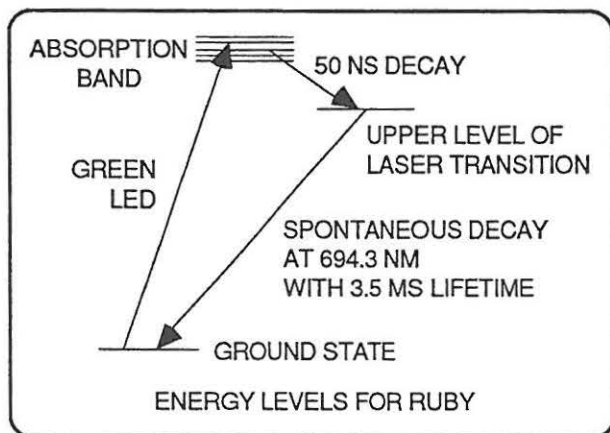
tion for a count rate of a few megahertz, and enough preamp gain to get pulse heights of about 100 mV. Using a 300 MHz oscilloscope, adjust the snubber termination impedance for the best pulse shape. Look carefully at the anode pulse shape and set the discriminator to a level which is above any ringing, but well below the mean pulse height. If there is lots of EMI or amplifier noise then increase the PMT's high voltage to increase the signal pulse height.

TYPICAL EXPERIMENT

This experiment is typical of many photon counting experiments. A pulsed light source is used to pump atoms to an excited state. Decays from the excited state are observed, allowing the lifetime of the state to be measured.

ENERGY LEVELS OF RUBY

The energy level diagram of Ruby is shown below. There are absorption bands around 400 nm and 550 nm. The Cr^{+++} ions which absorb light at these wavelengths decay in about 50 ns to the upper state of the well known laser transition. This state has a lifetime of about 3.5 ms, and decays to the ground state by emitting a photon at 694.3 nm.



The absorption band at 550 nm overlaps the emission line of a green LED. In this demonstration experiment, a pulsed green LED is used to quickly populate the excited state, and decays from the

excited state are seen through a bandpass interference filter centered on Ruby's 694.3 nm emission line.

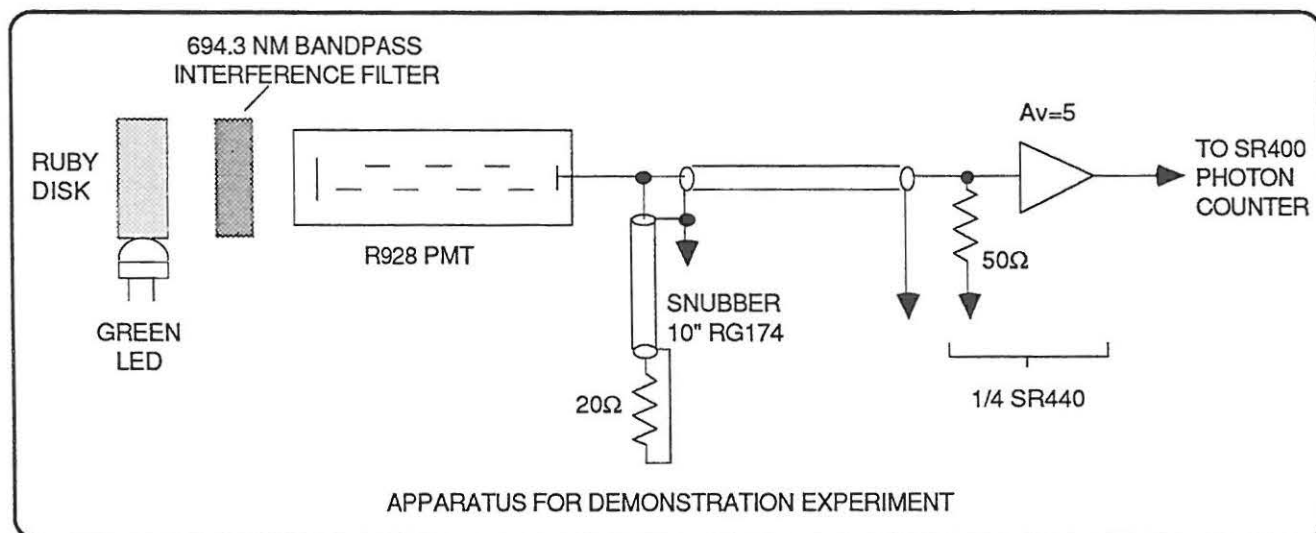
APPARATUS

The apparatus is shown below. The green LED is glued to the edge of a 1 cm diameter, 3 mm thick, ruby disk. The ruby disk is viewed through the bandpass interference filter by a Hamamatsu R928 PMT. This side-on PMT was selected for its high gain, fast rise time, and good red sensitivity.

The phototube base uses a tapered voltage divider, with about 3x the normal interstage voltage between the photocathode and the first dynode. This helps to narrow the pulse height spectrum for single photon events. The lower dynodes are bypassed, and 100 Ohm resistors are used between the dynodes and their bypass capacitors to reduce ringing in the anode signal. A snubber network consisting of a 10 inch piece of RG174 terminated into 20 Ohms is used to further reduce anode ringing and reduce the falltime of the output current pulse.

OPERATION

The PMT is operated at the maximum rated high voltage (1250 Vdc). The output pulses have a mean amplitude of 20 mV into 50 Ohms. To increase the pulse height to 100 mV, one amplifier in the SR440 preamp provides a gain of 5 with a 300 MHz bandwidth. The discriminator threshold is set to 20 mV. When viewed with a 300 MHz oscilloscope



Typical Experiment

cope, it is apparent that this threshold setting will count the majority of output pulses, but will not count anode rings or amplifier noise.

The green LED is flashed at a 60 Hz rate for about 500 μ s. During this time, population integrates in the upper level of the laser transition. Spontaneous decays from the upper level are counted by the photon counter during a 100 μ s gate. The intensity of the LED is adjusted for a count rate of about 10 MHz at the peak of the decay curve. This corresponds to counting about 1000 photons in each gate, or 60,000 photons per realtime second. The gate is scanned over 10 ms in 100 μ s steps in order to measure the lifetime of the state.

INSTRUMENT CONFIGURATION

The photon counter's gate generator is triggered by the same pulse which flashes the LED. The instrument is operated in the 'A FOR PRESET T' mode. The A-counter counts the photons which occur during the 100 μ s gate. The T-counter is preset to count 10 triggers, so data points will be sent to the computer six times per second at the 60 Hz trigger rate. After each data point is taken, the photon counter advances the A-gate by 100 μ s, resets the A-counter, and starts the next count interval.

Data may also be taken without a computer, using an X-Y chart recorder to record the data. The chart recorder is interfaced to the SR400 via two D/A outputs. The front panel D/A output has been set to output a voltage proportional to the value in the A-counter. (The D/A may also be set to output a value proportional to the Log of Counter A.) PORT1, a rear panel BNC output, has been set to take 0.1 V steps after each data point to provide a linear ramp to drive the X-axis of the chart recorder. Since the gate delay is also stepped after each data point, the chart recorder will plot the count rate vs. gate delay, showing the exponential decay of population from the excited state.

COMPUTER DATA ACQUISITION AND REDUCTION

The SR465, a data acquisition and control program for the SR400 photon counter, greatly expands the capabilities of the instrument. Intended for IBM PC's and compatible computers, the program sets the photon counter mode, reads, displays and stores scan data, and can fit curves to experimental results.

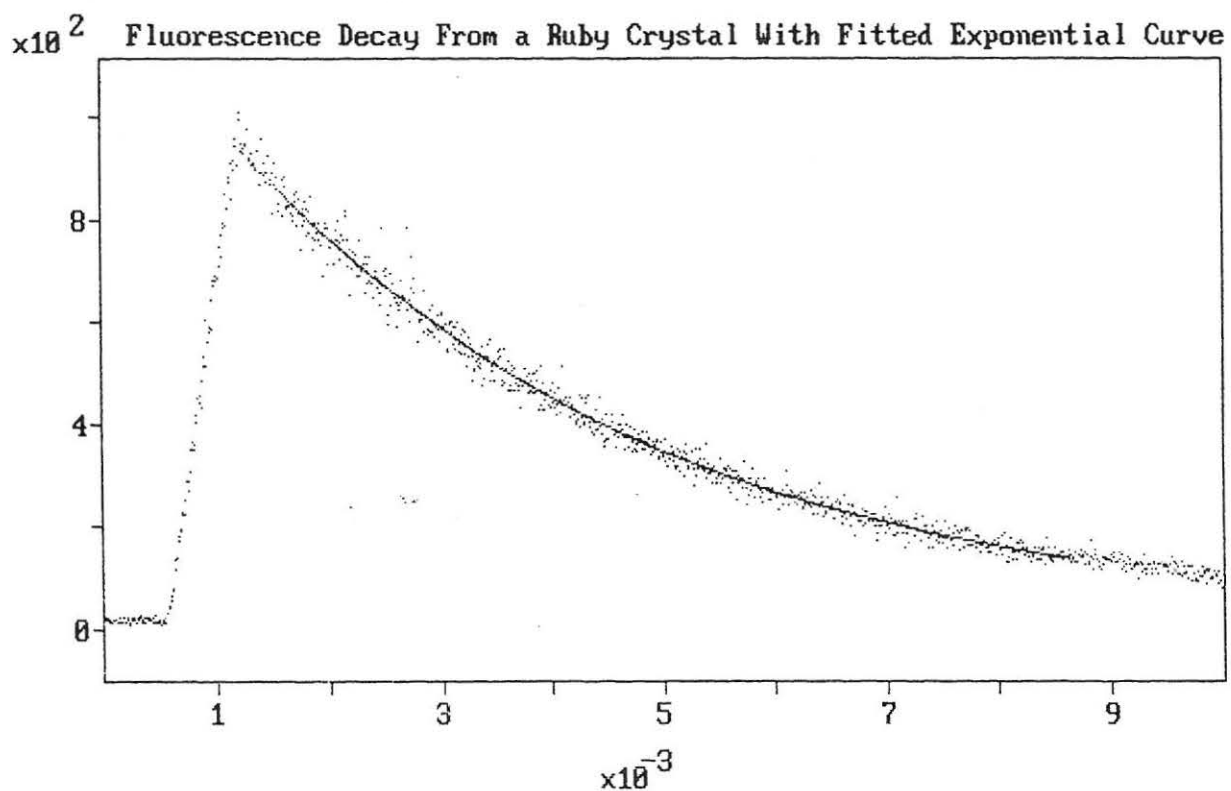
In this experiment, the program first sets the photon counter mode by recalling the instrument configuration from a disk file. Data is plotted in real time, with the results of several scans being aver-

INSTRUMENT CONFIGURATION

MODE:	A FOR PRESET T A = INPUT 1 T = TRIG T SET = 1E1 N PERIODS = 100 AT N=STOP DWELL=2E-3 D/A = A D/A RANGE = #####432# DISPLAY = HOLD	Counter A for preset interval Counter A to count input #1 Counter T to count triggers Counter T preset to 10 counts Data points in scan = 100 Single scan, 2 ms dwell D/A output shows Counter A D/A full scale is 9,999 counts Display shows final count value
A GATE:	A GATE=SCAN Δ =100.0 μ s A DELAY = 0.000 μ s A WIDTH = 100.0 μ s	Scan A's gate in 100 μ s steps First point's gate delay is 0 Gate width set to 100 μ s
LEVELS:	TRIG SLOPE = RISE TRIG LVL = +1.000 V A DISC SLOPE = FALL A DISC LVL = -20.0 mV PORT1 = SCAN Δ =+0.10 PORT1 LVL = + 0.00 V	Trigger on rising edge of LED Trigger threshold set to +1.00 A's disc on falling edge A's disc set to -20 mV Rear panel BNC takes 0.1 V steps Rear panel BNC starts at 0.0 V

aged to reduce statistical noise. When the data acquisition is complete, data reduction routines fit an

exponential to the decay curve to determine the lifetime of the excited state. Finally, the data curves are printed (see figure below), and stored to disk.



SR465 DATA ACQUISITION SOFTWARE

DEMO DISK

Included with every SR400 is a floppy disk containing a demo version of the SR465 data acquisition program for the IBM PC and compatibles. The demo SR465 is identical to the software described here except that data can not be saved to disk or printed on a printer. The complete version of the SR465 and source code may be ordered from Stanford Research Systems. To start the demo, run the program INTRO.

HARDWARE REQUIREMENTS

IBM PC/XT/AT or compatible with 512K RAM running DOS 2.1 or greater. 8087/80287 coprocessor is optional. RS-232 port or National Instruments PC2, PC2A, or PCIII card or Capital Equipment Corp. PC488 card required to communicate with the SR400.

SR465 PROGRAM DISK CONTENTS

SR465.EXE

SR465 program for systems without an 8087 numeric coprocessor

SR467.EXE

SR465 program for systems with an 8087 coprocessor

SR465.CFG

SR465 configuration file

SR4CON.EXE

Program for creating configuration files

RUBY.DAT

Sample data file

BINTOASC.EXE

Program to convert binary data files to ASCII format

SR465 GRAPHICS DISK CONTENTS

METAWNDO.EXE

Graphics Driver for SR465 program

PRTSCRN.EXE

Printer Driver for SR465 program

SYSTEM???.FNT

Font files for different graphics adapters

QUICK INTRODUCTION

This section provides a quick introduction to running the SR465 program. It is assumed that you have an IBM PC/XT/AT or compatible with 512K of RAM and a hard disk, running DOS 2.1 or greater. It is also assumed that you are familiar with the operation and features of the SR400 photon counter. If you encounter any problems with the procedure described below, refer to the User's Guide section of the manual for more detailed information.

1) Copy the contents of both the SR465 program disk and the SR465 graphics disk onto your hard disk.

2) Install the graphics driver by typing METAWNDO at the C> prompt followed by the Return or Enter key. (In the future, we will use the symbol <CR> to represent the Return or Enter key.)

3) Install the printer driver by typing PRTSCRN <CR> at the DOS prompt.

4) Run the SR465 program by typing SR465 <CR>. The program will respond with a description of the current graphics hardware environment. (Note that the program will always assume the existence of a mouse, even if none is present. Ignore the message concerning the mouse.) If the program's description of the graphics hardware is correct, type <CR> to run the program.

At this point the SR465 logon message should be displayed, giving the version number of the program. If not, consult the User's Guide section of the manual. Assuming it is, let us investigate a few of the features of the SR465.

Notice the menu bar at the top of the program screen. Options are selected from the menu bar by pressing the function key associated with the appropriate selection. The menu you see now is the "main menu." To return towards the main menu from any sub-menu use the <Esc> key. To retrieve

the sample data file, first press the F4 key to access the disk menu. Now get a directory of the data files by pressing the F3 key. You'll notice only the file RUBY.400 is displayed. To recall the file, simply press <CR>. Note that when the file is recalled a line of comments concerning the file is displayed in the menu area. To remove the comments press any key. Type <Esc> now to return to the main menu.

The exponential decay curve you should see was taken by observing the fluorescence decay of a ruby crystal pumped with a 500 microsecond pulse from a light emitting diode. (This experiment is described in the TYPICAL EXPERIMENT section of this manual.) The time profile of the decay was recorded by scanning the "A gate" of the SR400. Notice that the X-axis on the screen currently indicates sample number, not time. To make the X-axis indicate the "A gate" position, type F2 to enter the setup menu, and type F2 again to setup the display. Type F3 to choose the X-axis display. A vertical menu will appear on the screen indicating the quantities that can be plotted on the X-axis. Use the Cursor keys to select "A Gate" and type <CR> to confirm the selection. The X-axis will now display a range of 0-10 msec. Hit <Esc> a few times to return to the main menu.

One of the first things one might want to do with data of this sort is smooth it. Type F8 to enter the alternate menu, followed by an F1 to enter the smoothing menu. Select an 11 point smooth and type <CR>. The smoothed data will appear in a few seconds. We can now fit an exponential decay curve to the smoothed data. Type <Esc> to return to the alternate menu followed by an F3. Choose an exponential fit, and move the cursors using the left and right arrow keys to bracket the exponentially decaying region of the curve. The program will begin fitting the data and the χ^2 indicator in the menu area will report the current progress of the fit. After 10 or 20 seconds, the fit will appear on the screen and the parameters of the fit will be displayed in the menu area. Notice that the lifetime reported by the program will agree well with the accepted 3.5 msec lifetime of ruby.

This brief introduction should be sufficient to acquaint the user with the basic structure of the SR465. At this point, take some time to experiment with the program's features. If any problems arise, consult the User's Guide section of the manual.

USER'S GUIDE

HARDWARE REQUIREMENTS

The SR465 program is designed to work on an IBM PC/XT/AT or compatible computer equipped with at least 512K of RAM. An 8087/80287 numeric coprocessor is not required but will greatly improve the program's performance. To connect the computer to the SR400 you will need one of the following:

RS-232 port (COM1: or COM2:)

National Instruments GPIB card (PC2, PC2A, or PCIII)

Capital Equipment Corp. GPIB card (NEC7210 or TMS9914A version)

A wide variety of graphics hardware is supported by the SR465 program. Among the many graphics cards supported are:

AT&T DCB
AT&T DEB
IBM CGA
IBM EGA
Sigma Designs Color-400 card
Hercules Graphics Card
STB Graphics Plus-II Card
Tecmar Graphics Master
Vega Video-7 Deluxe
Everex Edge
Wyse-700
IBM PC-3270

Of course, the resolution of the screen display and the program's use of color will depend on the particular graphics hardware in use.

GRAPHICS CONFIGURATION

Before running the SR465 program it is necessary to install both the graphics driver and the printer driver. These two programs are located on the "SR465 Graphics Disk." To load the printer driver, type PRTSCRN<CR>. Once the printer driver is loaded, you can obtain a hard copy of the current screen display by pressing <Shift-PrtSc>. The following options may be specified when loading the printer driver:

PRTSCRN /1 Use the LPT1: printer
 PRTSCRN /2 Use the LPT2: printer
 PRTSCRN /n normal video
 PRTSCRN /i inverse video
 PRTSCRN /l "landscape" mode
 PRTSCRN /p "portrait" mode
 PRTSCRN /s show options in effect

Once PRTSCRN has been loaded, you may re-execute the program in order to change the options in effect.

The graphics driver provides a hardware independent interface between the program and the various graphics cards that may be present in the system. To install the graphics driver type METAWNDO<CR>. A message will be displayed confirming that the driver was successfully loaded. Because the driver occupies about 100K of RAM, you may wish to remove the driver at some point after it has been loaded. To do this, type METAWNDO /k <CR>.

The graphics driver makes use of the font files (SYSTEM??FNT) found on your SR465 graphics disk. Make sure that these font files have been copied to the drive and directory from which you will be running the SR465 program.

To run a program that makes use of the memory resident graphics driver, simply type the program name, as you would for any other program. The program responds by giving a brief description of the current graphics environment. For example, the program might display the following message on a system with an IBM EGA graphics card:

C>SR465

**IBM Enhanced Graphics Adaptor,
 640X350 16-color
 Microsoft Mouse driver in use.**

Press Return to continue . . .

The graphics driver always attempts to locate a mouse driver, even if none is in use. Since the SR465 programs do not use a mouse, the message concerning the mouse may always be ignored.

If the message accurately describes the graphics hardware, type <CR> to continue with the program. If not, type Control-C to abort the program

and restart it using one of the following switches to specify a particular hardware environment.

SR465 /A
 AT&T Display 640x400 monochrome.

SR465 /X
 IBM PC-3270 720x350 monochrome

COMMUNICATION CONFIGURATION

The SR465 program reads communication configuration information from the SR465.CFG file. When shipped, this program is configured to operate without an SR400. In this mode you will be able to recall data from disk files and use the data analysis routines, but of course no data can be taken. To configure the SR465 for a given set of communications hardware, run the program SR4CON.

SR4CON creates a new SR465.CFG file which reflects the communications hardware specified by the user. It is not necessary to run SR4CON every time you run SR465, only when specifying new communications settings.

USING THE NATIONAL INSTRUMENTS GPIB CARD

When using the National Instruments PC2, PC2A, or PCIII GPIB card, or the IBM GPIB card (which is functionally identical to the PC2A card) you must specify certain communication parameters using the program IBCONF supplied by National Instruments.

Consult your GPIB card manual for detailed information on running IBCONF. The required parameters are:

Device Name	Set to 'SR400'
EOS Byte	Set to 0AH (LF)
Terminate Read on EOS	YES
Set EOI with EOS on write	NO
Set EOI w/ last byte of write	NO

Note that the SR400's GPIB address may be set to any value as long as the value set on the instrument agrees with the value entered with IBCONF.

USING THE CAPITAL EQUIPMENT CORP GPIB CARD

The SR465 may be used with both the old (TMS9914A chip) and new (NEC 7210A chip) versions of the CEC GPIB card. However, Rev N of the CEC firmware contains an error which makes it incompatible with the SR400 and the SR465 program. To determine whether you have Rev N of the CEC firmware, run the program TEST488 supplied on the disk you received from CEC. If you do have Rev N, contact CEC for a firmware update.

The SR465 program assumes a starting address of 0C000:0000 (hex) for the CEC firmware. If you have changed your firmware starting address from this value (if you have an EGA card, for instance, this will have been necessary) you will have to edit SR465.CFG to reflect this change. Using any word processor, edit the value shown below in boldface to the decimal equivalent of the starting firmware segment.

```
4 0 0 0C
49152 696 0 0
0 15 10 11 14 12 8 4 15
```

For instance if your CEC firmware is located at 0D000:0000 (hex), change the value shown to 53248 (53248 = 13*4096 = D000 (hex)).

USING RS232

The SR465 can communicate with the SR400 on RS-232 via either the COM1: or COM2: port. (selected in SR4CON) In the RS-232 mode, the SR465 communicates at 19.2 Kbaud. It is not necessary to configure your computer's COM port with a MODE command. The SR465 takes care of all necessary configuration. Although DOS only exploits communication rates up to 9.6 kbaud, the hardware in IBM and compatible computers works perfectly well at 19.2 kbaud.

The cable used with the SR465 program should be a 'straight through' RS-232 cable, not a null-modem cable. This is because the SR400 is configured as a DCE while most computer COM ports are DTE. Only the standard 7 wire RS-232 connection is necessary.

Make sure that the RS-232 parameters of the SR400 are set correctly. The COM menu should be set to:

```
RS232 BAUD=19200
RS232 BITS=8
RS232 PARITY=NONE
RS232 ECHO=OFF
```

PROGRAM OPERATION AND ORGANIZATION

Since the SR465 uses a modern, menu-driven, interface, many facets of its operation are self-explanatory. Menu items are selected from the menu bar at the top of the screen, or by using the cursor keys to highlight the appropriate item on a "pop-up" menu. In the popup menus, the cursor keys are used to move the selection up and down, while the <Home> key is used to select the first selection on the list. To leave a menu and return to the previous menu, it is usually necessary only to press the <Esc> key.

The structure of the SR465 program follows the organization of the SR400 photon counter closely. The program maintains two buffers corresponding to the A and B counters of the SR400. These counter buffers are exact copies of the counter buffers in the SR400's memory. When a scan is started the program begins transferring the contents of the SR400's counter buffers into its own counter buffers. The SR465 program also maintains separate display arrays to store data displayed on the screen. This distinction means that it is always possible to recall the original, unadulterated, counter data no matter what operations have been performed on the data on the screen. This is more fully explained below in the section on the Display Setup menu.

MENU CHOICES: MAIN MENU

F1 Scan/Pause/Resume

The F1 key initiates data transfer from the SR400 to the computer. Any values that have been programmed to scan (discriminator levels, gate delays, etc.) will begin to scan in accordance with the values programmed in the F2 (Setup) menu. As data is taken it will be plotted on the screen. Note that, although at slow scan rates or large dwell times data appears to be plotted synchronously with the SR400, at faster scan rates, plotting of data will lag behind the SR400. Because of the SR400's buffering capabilities, however, in no case will any data be lost.

When the F1 key is first pressed, the program responds by asking how many scans are to be performed. In the multiple scan mode, the program sums data from individual scans into the computer's memory. It is often more desirable to take several scans with short count intervals and sum them rather than a single scan with long counting intervals, since in the former case, the effects of any long term drifts are distributed equally over all parts of the scan. Of course at the end of each individual scan any gate delays or discriminator levels that are in the 'scanning' mode are reset to their initial levels.

Once a scan has been initiated, the F1 menu selection changes to 'Pause.' Pressing F1 in this mode causes the SR400 to temporarily stop data acquisition, although the counter buffers maintain their values. Pressing F1 to pause is equivalent to pressing the 'Stop' key on the SR400 once. Once paused, the F1 menu selection changes to 'Resume.' Data acquisition is resumed by pressing the key in this mode.

When the scan finishes, the F1 selection reverts to 'Scan.' Notice, however, that a new scan may not be initiated until the F6 key has been used to reset the counter buffers. This is to prevent accidental loss of data.

F2 Setup

The Setup menu allows the user to change the configuration of the SR400, the screen display, and store and recall files containing complete SR400 setups. Each submenu will be individually described.

F1 Setup SR400

This selection lets the user change the current settings of the SR400 photon counter. A submenu is displayed allowing selection of three general categories, 'Mode,' 'Levels,' and 'Gates.' These categories correspond to the menu selection buttons on the front panel of the SR400. Once a category is selected, a popup menu appears listing the items in that category that may be programmed. The current value of the highlighted item is displayed at the top of the screen. To highlight a different item simply use the cursor keys. To change the value of the highlighted item press <CR>. A new popup menu will appear listing the possible values of the highlighted item. Move the

cursor to the desired value and press <CR>. To escape without changing the value of the item, use the <Esc> key. When the selected item requires numeric entry (for instance a gate delay), the program will beep and a blinking cursor will be displayed at the top of the screen. Type in the desired numeric value and press <CR>.

F2 Setup Display

This selection is used to change the SR465 screen display. The program allows simultaneous display of two quantities, corresponding to the labeled axes on the left and right side of the screen. (On Color systems, the data and axes are color coded) In addition, the program allows the user to vary the quantity displayed on the X-axis. The first submenu allows the user to select the appropriate axis using the function keys. Once an axis has been chosen, a popup menu appears listing the various quantities which may be displayed on that axis. These are:

- | | |
|--------|---|
| A | Plot the contents of the Counter A buffer |
| B | Plot the contents of the Counter B buffer |
| A+B | Plot the sum of the A and B buffers |
| A-B | Plot the difference of the A and B buffers |
| A/B | Plot the ratio of the A and B buffers. If B=0 then A/B is set equal to the value in the A counter. |
| Disc A | Plot the corresponding SR400 parameter. Note that |
| Disc B | unless the chosen parameter is in its |
| Disc T | 'scanning' mode, the displayed value will be a constant. In particular, |
| Gate A | choosing a constant value for the X-axis display |
| Gate B | causes the entire graph to be compressed into a |
| Port 1 | vertical line. This is not recommended. |
| Port 2 | |
| CAL | Only allowed on the X-Axis. Allows the user to display an arbitrarily calibrated X-axis. The X value for two points must be entered using the 'Cursor' menu before this display option may be chosen. |
| Null | Plot nothing on this axis. This should not be chosen for the X-axis |

Index The ordinal index of the scan, e.g. the 43rd scan point has index 43. This is the default choice for the X-axis.

When a selection has been made for a given axis, the program copies data from the counter buffers into the appropriate display array. Since the counter buffers are unaffected by smoothing, scaling, and other display functions, the setup display menu provides a means of recovering the original counter data after several screen operations have been performed.

F3 DIR Setup Files

A complete SR400 instrument setup may be stored to disk in the form of a 'setup file'. These files have the extension '.465'. Choosing DIR in the setup menu causes the program to display a directory of all the setup files in the current disk and directory. (The current drive and directory may be changed in the F4 'Disk' menu; this is described later on.) To recall a setup file move the cursor to the appropriate file and press <CR>.

Note that no data is stored in the setup files, only instrument settings. (Any Macros currently in effect are also stored in the setup files. A description of Macros is given later on in this manual)

F4 Save Setup File

This selection saves the current SR400 settings in a setup file. The program prompts for the file name, and the setup file is saved in the current drive and directory. Entering the filename preceded by a drive:path allows the file to be saved in an arbitrary directory. It is not necessary to enter an extension with the filename, the program supplies the '.465' extension.

F5 Get Setup File

Allows the user to recall a setup file with a specific name. Once again, a drive:path may be specified, and the extension '.465' need not be entered by the user.

F6 Summary

Provides a full screen summary of the current SR400 settings. Pressing any key returns to the normal display. The summary screen may be ac-

cessed in two additional ways. On graphics cards with two graphics pages (e.g. Hercules card or EGA w/256K of RAM) the summary screen can be displayed at any point in the program by pressing <Ctrl-S>. On systems with only a single graphics page, the summary can be displayed by pressing <Ctrl-S> only when the program is in the main menu or the 'Other' menu.

F3 Scale

The Scale menu allows the user to scale the graphs displayed on the screen. After selecting the axis to be scaled the program offers three scaling options.

Enter Values

Allows the user to explicitly enter minimum and maximum values for the axis.

Zoom

Displays a cursor on the screen which may be moved using the cursor keys. The F1 and F2 keys provide a factor of 2X zoom in and zoom out respectively around the cursor position.

Autoscale

Provides a 'best guess' scaling based on the data being plotted, current scan progress, etc.

F4 Disk

This menu is concerned with saving and recalling data files. The data files have the extension '.400' to differentiate them from setup files (which have the extension '.465'). Data files contain the current screen display, the current values of the counter buffers, and the current scaling of the screen displays. The submenu choices are as follows:

F1 Save

Saves the currently displayed data in a data file. The user is prompted for a filename, which is automatically supplied with the extension '.400'. The file is saved in the current default drive and directory, unless a specific drive:path is explicitly included in the filename.

F2 Get

Recalls a data file from the disk. The user is prompted for a filename, which is automatically supplied with the extension '.400'. The current default drive

and directory is searched for the given file, unless a specific drive:path is explicitly included with the filename.

F3 DIR

Displays a list of data files in the default drive and directory. A file may be recalled from this list by moving the cursor to the correct filename and pressing <CR>.

F4 Set Defaults

Allows the user to enter the default disk drive and directory for both data and setup files. If no default is explicitly specified the program looks for files in the current directory, i.e. the directory from which the program was run.

F5 Cursor

The cursor has a variety of functions. Basically, it displays a cursor which may be moved with the cursor keys, and displays the data values corresponding to the cursor position at the top of the screen. Note that the <Home> key always moves the cursor to the far left of its allowed field, while the <PgUp> key moves the cursor to the far right of its field. The cursor movement employs an acceleration routine which causes the cursor to move faster when the cursor key is pressed more rapidly. In addition, the cursor may be used to change data and to calibrate the X-axis.

Pressing the F1 key in the cursor mode allows the user to change the data value for either the left or right axis.

Pressing the F2 key allows the user to enter a 'calibrated' X-value for the current cursor position. Once two such calibration points are entered, the user may choose CAL from the X-axis Setup Display menu to display a calibrated X-axis.

F6 Reset

The F6 key clears both the display and the counter buffers and readies the program for a new scan. Note that it is necessary to press F6 twice to activate the reset function. This is done in order to prevent loss of data when hitting F6 accidentally.

F7 Quit

The F7 key allows the user to quit the SR465 program and return to DOS.

F8 Other

The F8 key causes the program to display the 'Other' menu.

OTHER MENU FUNCTIONS

The 'Other' menu is primarily concerned with data manipulation routines. The 'Other' choices are:

F1 Smooth

Allows the user to smooth the data displayed on either the left or right axis. The program employs a Savitsky-Golay smoothing algorithm using 5, 11, 17, or 25 points. For details on the Savitsky-Golay smoothing technique see Abraham Savitsky and Marcel J.E. Golay, "Smoothing and Differentiation of Data by Simplified Least Squares Procedures," Analytical Chemistry, (36) 8, July 1964.

F2 Stats

Displays a variety of statistical information about a display region selected by the user. For the region between the cursors, the program calculates: the mean, the variance (the sum of the squared deviations from the mean), the area under the data curve, the area under the baseline defined by the two endpoints, and the net area (total area minus baseline area). In addition, the program displays the curves $Y \pm \sqrt{Y}$, the region bounded by a single standard deviation of the Poisson distribution. Since the photon rate in low light level situations is governed by the Poisson distribution, the standard deviation can provide clues as to the statistical significance of a given data feature.

F3 Fit

The SR465 program allows the user to fit one of three curves to displayed data: lines, exponentials, and Gaussians. To use the fitting feature, select the desired type of fit from the popup menu, and select a region to be fit using the cursor keys. The SR465 uses the Marquardt 'Gradient-expansion' algorithm to minimize χ^2 . As the fit progresses, the current value of χ^2 will be displayed at the top

of the screen. The program is designed to terminate the fit when the change in chi squared falls below .0001%, although the fit may be terminated at any time by pressing the F1 key. When the fit terminates, the fitted curve is plotted and the parameters of the fit are displayed at the top of the screen. Once a fit has been performed, the fitted curve will remain on the screen until 'Deactivate current fit' is chosen in the Fit submenu.

F4 Calc

Calc, short for calculate, allows the user to perform simple arithmetic calculations with the data displayed on the screen. This menu selection, first prompts you for an axis to operate on, then an operation, and finally, an operand. The operand can be either a constant, or the data corresponding to the left or right axis. Only the display of the first selected axis is changed, the operand axis remains unchanged.

F5 Cursor

This selection performs the same function as in the main menu.

F6 Label

"Label Mode" removes the menu bar at the top of the screen to allow the user to create hardcopies without a menu. In addition, a blinking cursor lets the user type text onto the display screen. Once again, to dump the screen display to a printer, use the <PrtSc> key. To remove the cursor from the display, simply move it off the screen with one of the cursor keys.

F7 Quit

This selection performs the same function as in the main menu.

F8 Other

Returns the display to the "Main Menu"

MACROS

A powerful feature of the SR465 program is the ability to record sequences of keystrokes in a 'Macro,' and play back the sequence with a single keystroke. The SR465 can store up to 5 macros, each containing 200 keystrokes. To record macro #n, press Control-Fn, where Fn is the function key F1-F5. The message "Recording Macro" will be briefly displayed at the top of the screen. Then, enter the keystrokes normally. To finish the macro, type Control-Fn again. The message "Terminating Macro Recording" will be briefly displayed. To play back the macro, type Shift-Fn.

A number of cautions must be observed when using the macro facility. First, remember that a macro must normally be played back starting from the same point in the program that it was recorded from. If not, the sequence of recorded keystrokes will most likely make no sense. For this reason, it is often useful to include a series of several 'Escapes' at the beginning of each macro to return the program to the main menu. Secondly, when choosing items from a popup menu, remember to use the 'Home' key to move the cursor to the top position before selecting the desired item. That way, no matter what item is currently active when the macro is played back, the recorded sequence will still make the same menu selection.

The macros are recorded in the '.465' setup files, so that a frequently used set of macros may be conveniently stored and recalled.

SR465 DATA FILES

To keep the length of the data files down, the SR465 uses a binary file format to store data files. Unfortunately, this is not the most convenient format for other programs to access the data stored in these files. We have therefore provided a program BINTOASC.EXE to convert SR465 data files into an ascii format.

BINTOASC prompts for an input file name; be sure to include the '.400' extension. The output is an ascii file with the same name and extension '.ASC'. For example, here is a section of the file RUBY.ASC created by using BINTOASC on the sample data file RUBY.400:

ruby fluorescence decay

```

19 0 1.0000E+00 1.9000E+01 0.0000E+00
15 0 2.0000E+00 1.5000E+01 0.0000E+00
17 0 3.0000E+00 1.7000E+01 0.0000E+00
16 0 4.0000E+00 1.6000E+01 0.0000E+00
16 0 5.0000E+00 1.6000E+01 0.0000E+00
11 0 6.0000E+00 1.1000E+01 0.0000E+00
10 0 7.0000E+00 1.0000E+01 0.0000E+00
11 0 8.0000E+00 1.1000E+01 0.0000E+00
15 0 9.0000E+00 1.5000E+01 0.0000E+00

```

The first line of the ascii file is any comment stored with the file. Each of the next lines corresponds to a single data point.

The first value gives the A counter contents at the point, followed by the B counter contents, the X-axis value, the lefthand Y-axis, and the righthand Y-axis. Data in this format may be easily read using virtually any computer and language.

For those users who must access the binary files directly, the format of the file is given below:

BINARY FILE FORMAT

<u>Variable</u>	<u>Type</u>	<u>Explanation</u>
comments	character*60	Comment line for file
nBin	integer*2	Next data point in scan
nOfBins	integer*2	Total number of data points in scan
Scanning	logical*2	Indicates scan in progress
func(3)	integer*2	Function code for left axis, right-axis and X-axis (see below)
nScn	integer*2	current scan number
nleft	integer*2	index of leftmost displayed data point
nright	integer*2	index of rightmost displayed data point
NofScans	integer*2	total number of scans in set
ymin(3)	real*4	minimum values for left axis, right-axis and X-axis
notused	real*4	not used
deltay(3)	real*4	max-min values for left-axis, right-axis, and X-axis
notused	real*4	not used
Paused	logical*2	indicates scan in paused state
qADc	real*4	initial value for A Discriminator
qADcInc	real*4	increment for A Discriminator
qBDc	real*4	initial value for B Discriminator
qBDcInc	real*4	increment for B Discriminator
qTDc	real*4	initial value for T Discriminator
qTDcInc	real*4	increment for T Discriminator
qAGtDly	real*4	Gate A initial delay
qAGtWth	real*4	Gate A width
qAGtInc	real*4	Gate A delay increment
qBGtDly	real*4	Gate B initial delay
qBGtWth	real*4	Gate B width
qBGtInc	real*4	Gate B delay increment
qPrt1	real*4	Port 1 initial value
qPrt1Inc	real*4	Port 1 increment
qPrt2	real*4	Port 2 initial value
qPrt2Inc	real*4	Port 2 increment
ActCal	logical*2	indicates an active x-axis calibration

<u>Variable</u>	<u>Type</u>	<u>Explanation</u>
notused	integer*2	not used
nCal(2)	integer*2	indices of first and second cal points
xCal(2)	real*4	X values of first and second cal points
CntrA(1)	integer*4	Counter A Buffer
.		
.		
CntrA(nOfBins)		
CntrB(1)	integer*4	Counter B Buffer
.		
.		
CntrB(nOfBins)		
Xchn(1)	real*4	X-axis data
.		
.		
Xchn(nOfBins)		
LftChn(1)	real*4	Lefthand Y-axis data
.		
.		
LftChn(nOfBins)		
RtChn(1)	real*4	Righthand Y-axis data
.		
.		
RtCHn(nOfBins)		

SR440 PREAMPLIFIER

The Model SR440 Fast Preamp contains four wide bandwidth, dc coupled amplifiers designed to be used independently or cascaded to provide gains of 5, 25, 125 and 625.

The fast rise time, low noise and dc accuracy of the SR440 make it the ideal instrument for amplifying outputs of fast photomultiplier tubes and photodiodes.

SPECIFICATIONS

Inputs	50 Ω impedance, dc coupled, BNC connectors.
Outputs	dc coupled, BNC connectors (terminate into 50 Ω .)
Voltage Gain	5.0 per channel, $\pm 2\%$, cascadable to 625.
Bandwidth	dc to 300 MHz.
Input Range	± 250 mV.
Linearity	2%
Noise	<25 μ V rms referenced to the input.
Stability	10 μ V/ $^{\circ}$ C referenced to the inputs (0 - 50 $^{\circ}$ C.)
Offset	input ± 50 μ V adjustable.
Rise/Fall Time	1.2 ns
Propagation Delay	2 ns per channel.
Recovery Time	< 4 ns for a times 20 overload.
Protection	± 3.5 V dc, ± 50 V transient.
Mechanical	7.7" x 6.7" x 2".
Power	16W, 100/120/ 220/240 V, 50/ 60 Hz

OPERATION

The SR440 Fast Preamplifier is useful for amplifying small signals to levels that allow processing by other instruments such as the SR400 Gated Photon Counter. The four channels may be used independently or cascaded as desired. The following table gives the maximum input voltage for linear operation as a function of gain for 1 to 4 channels,

<u>Channels</u>	<u>Gain</u>	<u>Maximum Input</u>
1	5	250 mV
2	25	50 mV
3	125	10 mV
4	625	2 mV

TROUBLESHOOTING

WARNING: Dangerous voltages are present on the printed circuit board. Always turn the power off and disconnect the line cord before removing the cover or changing components. To calibrate the unit while it is operating, first disconnect the line cord, remove the 4 screws from the feet and slide the chassis forward about three inches. In this manner, no dangerous voltages will be exposed when power is reconnected.

The diodes in the overload protection circuit can be damaged by excessive voltages at the input of any stage. To check for damage measure for 0.6 Volts across both diodes in D1. D2 and D3 can be tested by simply removing them from the circuit and then checking for degradation of the amplifiers pulse response as each diode is reconnected. Sockets have been provided for these parts and replacement diodes are available from the factory.

CALIBRATION

To calibrate the input offset, connect a voltmeter to the input and adjust P1 for a reading of 0 ± 10 μ V. To calibrate the output offset, connect a voltmeter to the output and adjust P2 for a reading of 0 ± 50 μ V.

SR400 CIRCUIT DESCRIPTION

The SR400 has five printed circuit boards. Almost all of the components are located on the main circuit board, including the microprocessor and interfaces, fast counters and discriminators, gate generators, and digital to analog conversion circuits. There is one circuit board in the power supply enclosure. The other three circuit boards are much smaller and are mounted to the front panel. They hold the 2 line by 24 character LCD display, the key pad, and the status LED indicators.

MICROPROCESSOR CONTROL

MICROPROCESSOR SYSTEM (sheet 12)

The microprocessor, U1206, is a Z8800 (Super 8) microcontroller which integrates a fast processor, UART, counter-timers, interrupt controller, DMA controller, and RAM into one VLSI component. This high degree of integration is essential to the performance of the SR400.

The Super 8 is clocked by the 20 MHz crystal time base oscillator. The data bus is multiplexed with the lower 8 bits of the address bus. U1207 latches the address bits at the start of every memory cycle. The firmware and calibration bytes are stored in the 32K x 8 UVEPROM, U1208. U1209 is a 32K x 8 CMOS static RAM. The large RAM allows large scans (up to 2000 points) to be internally buffered. The RAM is battery backed up allowing instrument settings to be saved. Q1201-3 perform power-up and power-down reset and RAM protection.

I/O port strobes are generated by U1201 and U1211, RD and WR are decoded by U1203.

RS232 INTERFACE (sheet 12)

The Super 8 has an on-chip UART and baud rate generator. U1205 level shifts and buffers the signals to the external connector. The connector is a data communications equipment (DCE) type. The baud rate is derived from the processor clock. Any standard baud rate from 300 to 19.2K baud may be programmed.

GPIB INTERFACE (sheet 13)

The GPIB (IEEE-488) interface is provided by the TMS9914A controller, U1311. U1309 and U1310 buffer data I/O to the GPIB connector. U1311 is programmed to interrupt to the processor whenever

there is bus activity addressed to the SR400.

INPUT PORTS (sheet 13)

U1301 and U1302 read the prescale bits from counters A and B. They are latched by CTR_1 to provide unambiguous data while the counters are counting. U1303 is the keyboard input buffer. The upper bits of U1301-3 are used to read various status signals.

SLOW COUNTERS (sheet 13)

U1304-8 are 5 MHz 8253 VLSI counters. Each chip has 3 multi-mode 16 bit counters. These counters count the middle bits in counters A,B, and T, the gate delay and width oscillators, and the crystal time base for the dwell time.

OUTPUT PORTS (sheet 14)

U1401-1406 are 8 bit data latches. U1401, U1402, and U1406 provide control bits for the discriminators, counters, and gates. U1403 holds the preset value for the T counter. U1404 and U1405 drive the front panel LED indicators and keyboard strobes.

FRONT PANEL (sheets 12,13,14,16)

The front panel is connected to the main board through 5 cables.

The knob is an optical encoder buffered by U1202. Each transition of its outputs interrupts the processor which keeps track of its direction and speed.

The LCD display connects directly to the processor data bus. Besides the 8 data lines, there are 2 address lines, chip select, power and ground, and display contrast control.

The electroluminescent backlight requires a 100 Vac, 100 Hz signal. This power is derived from the dc power supplies and is isolated from the ac main.

The status LED indicators are driven by U1405.

The key pad is strobed by U1405 and read by U1303.

DIGITAL TO ANALOG CONVERTER (sheet 15)

U1502 is a 12-bit D/A converter which is loaded 4 bits at a time by the processor. The voltage output from 4/4 U1504 is 0 to -10.24 Vdc. The rest of U1504 shifts and attenuates this output to provide control voltages. These analog outputs are multiplexed onto 20 sample and hold amplifiers. Each S/H is refreshed every 2 ms.

U1503 multiplexes the gate generator calibration voltages. U1506 multiplexes the gate adjustment voltages as well as the analog outputs and the LCD contrast. U1501 provides the discriminator thresholds.

FAST COUNTERS

SIGNAL INPUTS (sheet 1)

Signal inputs INPUT 1 and INPUT 2 pass through amplifiers AMP 1 and AMP 2. These are dc to 300 MHz, gain of 5 amplifiers. They have a 1.2 ns rise/fall time, ± 300 mV input range, and a 4 ns recovery time from a 20X overload. The inputs are protected to ± 5 Vdc and ± 50 V transients.

The following is a brief description of amplifier AMP 1. AMP 2 is identical. R102, R103, R104, P101, D101, D102, and D103 comprise an overload protection circuit which clamps the inputs to the amplifiers at $\pm .35$ V. Ac gain is provided by Q101 and Q102 and is compensated by C107. U101 sets the amplifier's dc gain. Q102 sums the ac and dc signals and can drive its 100 Ω output to ± 2 V. U102, D130, and Q130 provide a temperature compensated bias voltage for the overload protection circuits which is adjusted by P101 to null the dc offset at the input.

DISCRIMINATORS (sheet 2)

Comparators U202, U203, U204, U209 discriminate the AMP 1 and AMP 2 outputs into ECL levels. U202 is the A discriminator when the input to counter A is INPUT 1. The threshold is provided by analog voltage A_DISC and buffered by 1/4 U208. The output transistor, 1/4 U207, provides a high frequency, low impedance output. The comparator is operated in the Schmitt trigger configuration with about 20 mV of hysteresis. Since the input signal has been amplified by 5, this represents about 4 mV of hysteresis at the input. Control signal

A_POL inverts the comparator output in 1/4 U1206 if the discriminator slope is set to FALL.

U203 is the B discriminator when the input to counter B is INPUT 1. If the input to counter B is INPUT 2, U209 is the B discriminator. Operation is identical to the A discriminator except that the outputs of U203 and U209 are multiplexed through U210. The B_SEL line selects the appropriate output and B_POL programs the slope.

U204 is the T discriminator when the input to counter T is INPUT 2.

The comparators may be inhibited by the DISC_INH signal which is derived from the INHIBIT input on the rear panel. When DISC_INH is asserted, the comparators will be inactive and their outputs frozen.

U201 is the gate trigger discriminator. It operates like the input discriminators except that it has about 75 mV of hysteresis.

A AND B COUNTERS (sheets 3 and 5)

There are 2 gated counting channels capable of 220 MHz operation and essentially infinite count capacity. Counter A will be described below.

The input to counter A is selected by 2/4 and 3/4 of U301. The signal APULSE_INH disables input from INPUT 1 and the signal A10MHZ_INH disables input from the 10 MHz timebase. The selected input is buffered by 4/4 of U301 to drive the A_DISC output drivers, Q301 and Q302. When both inputs to 1/4 of U301, !A_GATE and !TIMER are both low, the counter is enabled. (!SIGNAL means NOT_SIGNAL or SIGNAL_BAR) !TIMER is low the duration of the count period as determined by the preset condition. !A_GATE is low when the A gate is active. !A_GATE is always low if A gate is CW. Thus only those gates which fall within the timer count period enable counting. 1/2 U302 is the gate multiplier. If a rising edge from the input source clocks 1/2 U302 while !A_GATE and !TIMER are both low, then a 2 ns positive pulse will appear at the Q output. This ensures that only leading edges of the input that occur during the gate are counted. 2/2 U302 and 1/2 U303 are divide by 2 flip flops and are the low 2 bits of the counter. 1/4 and 2/4 of U307 translate their outputs to TTL levels. U304 is a FAST TTL divide by 16 counter which counts the output of the ECL flip

flops. 1/4 and 2/4 of U307 and U304 comprise the low 6 bits (/64) of the counter. Their outputs can be read by the processor via U1301. The last bit of U304 then clocks 1/3 U1305, an 8253 counter. This counter is programmed to divide by 6250. Its output then interrupts the processor and the highest bits of the count are updated within the processor. The signal !RST_TTL clears all the counters to zero. LOAD preclocks the 8253 after !RST_TTL to preload the counter. The counter is reset at the end of each count period.

The input to counter B is always B_PULSE and is buffered by 2/3 U305 to drive the B_DISC output driver. The counter is enabled when !B_GATE and !TIMER are both low. !TIMER may be forced low by B_SET in the case when B counter is preset. In this case, the B counter is still gated by !B_GATE. The rest of B counter is identical to A counter.

T COUNTER (sheets 4 and 5)

The input to counter T is selected by U401 and can be the 10 MHz timebase, the gate trigger, the output of the T discriminator, or B_PULSE X B_GATE (when B is preset). Note that only counter T is presettable and when B is selected to be preset, counter B's input is routed to counter T. 3/3 U305 buffers the input and drives the T_DISC output. TDISC_INH inhibits the T DISC output when B is preset. Counter T is in a timer configuration. 1/2 U402, U403, FAST TTL counter U404 and the 3 counters in U1304 comprise the T counter. The low 6 bits are preset by CT0-CT6. U1304 is preset by the processor loading the counters. Counter T is preset to the programmed TSET or BSET number at the end of each count period. When !COUNT goes low, the next rising edge from the input will clock 2/2 U402 starting the TIMER or count period. The ECL counter 1/2 U402 is then enabled to count. Counting continues until the preset number of counts have occurred and all inputs to 2/2 U406 are low. When this happens, 1/2 U405 latches this condition and terminates the TIMER pulse from 2/2 U402. The TIMER pulse thus lasts for the preset number of pulse periods from the input. Note that counting does not begin until the first pulse after !COUNT goes low. The end of the TIMER pulse generates TIMEOUT which signals that counting has halted and the counters may be read. The processor can use the !STOP to halt counting at any time.

START/STOP AND DWELL (sheet 6)

U605 is a 20 MHz crystal oscillator and is the timebase for the entire instrument. The output of U605 directly clocks the processor. U602 provides 10 MHz to the counters and 5 MHz to the dwell timer and GPIB controller.

2/3 and 3/3 of U1308 make up the dwell timer. Assuming that the counters have been reset, the Q output of 2/3 U1308 will be low and the !Q output of the start latch, 1/2 U604, will be high. When the START key is pressed or a CS command is received, the processor asserts !CPU_START which sets !Q of the start latch low. An EXTERNAL START INPUT does the same thing. !Q going low clocks the dwell latch, 2/2 U603, so that its Q output goes high. This in turn sets !COUNT low and the count period begins. When the preset condition is met and the TIMER pulse terminates, the TIMEOUT signal clocks the timeout latch, 1/2 U501. The timeout latch interrupts the processor indicating that counters A and B have valid data. The timeout latch also clears the dwell latch, 2/2 U603, raising !COUNT high again. This also resets and enables the dwell counters in U1308 to count down the programmed dwell time. During the first 1 ms of the dwell time, the processor is reading and resetting the counters, reloading the T counter, and updating any scanned parameters. When the processor is finished, it strobes !T_REARM low which resets the timeout latch. When the dwell timer times out, the Q output of 2/3 U1308 goes low which clocks the dwell latch and counting begins again.

If the dwell mode is external, the output of the dwell timer is always low and the EXT START INPUT clocks the start latch to start the next count period. 1/4 U601 disables the start latch while counting is in progress and is not re-enabled until the processor is finished servicing the counters. The EXT STOP INPUT clocks the stop latch, 2/2 U603 if the counters have previously been started. The stop latch asserts !STOP which terminates the TIMER pulse and generates a TIMEOUT signal.

The DWELL OUTPUT is driven from the dwell latch via Q601 and Q602. This output is high between count periods. The minimum output low pulse width is 100 ns even if the preset condition is shorter.

GATE GENERATORS

The gates are generated using four programmable delay generators, one for each delay and one for each width. Each delay can range from 10 ns to 1 s. Each delay is made up of a 4 MHz counter and an analog delay which covers 250 ns. To program a delay, the appropriate number of oscillator clocks (250 ns) are loaded into the counter and the remainder (<250 ns) is programmed into the analog delay. All counters and analog delays are hardware reset when they terminate to give a 1 MHz retrigger rate.

The following discussion focuses on the A GATE delay. B GATE delay and both gate widths are similar.

DIGITAL DELAY (sheet 8)

The discriminated trigger input clocks the trigger latch, 2/2 U405. The output of the trigger latch is converted to the TTL signal, DLY_ENA. DLY_ENA starts the 4 MHz delay oscillator. The output of this oscillator is phase coherent with the gate trigger.

The delay oscillator is a TTL oscillator comprising U803 and U801. U803 is a fast TTL comparator with U801 as an output buffer. The RC feedback network of R810 and C806 sets the frequency. U801 is used as the buffer since HC gates have output swings which are not temperature dependent but follow their power supply voltages. The +5V power for this oscillator is derived from the precision +10.000V reference, U1505. 1/2 U802 and Q801 form a precision, low drift +5V regulator. This results in a temperature stable frequency.

The frequency may be adjusted slightly by changing the analog voltage DLY_FREQ. 2/2 U802 and Q802 form a current source to differential pair Q803 and Q804. Q803 and Q804 switch this current from ground to C806. When U801 is discharging C806, Q803 will be charging C806. Thus the half cycle when C806 is high can be lengthened by changing the DLY_FREQ voltage and the frequency adjusted by the processor. The proper DLY_FREQ voltage is determined at the factory and stored in the ROM. This provides automated calibration of the delay clock.

The 4 MHz signal is counted by 1/3 and 2/3 of U1307. When the programmed number of cycles have been counted, the output of 2/4 U805 goes

low. The next clock pulse clocks the latch 1/2 U804 to generate !DLYA_TIMEOUT which signals the end of the digital portion of the delay. The line !DLYA_COUNT0 is asserted low when no digital delay is required (short delays).

ANALOG DELAY (sheet 7)

The analog delay is determined by charging the capacitor C706 with the constant current source Q701. The constant current source is programmed by D/A voltage DLYA_SLOPE and is calibrated by the processor. When the digital portion of the delay is complete, !DLYA_TIMEOUT goes low which causes the differential pair, Q702 and Q703, to switch the current away from the operational transconductance amplifier (OTA) 1/2 U704 and to C706. The voltage on C706 then ramps up until it reaches -5 V at which point the differential comparator, Q704 and Q705, switches state and generates the EOD_A (end of delay) signals. The OTA is used to precharge the capacitor C706 to a voltage programmed by D/A output DLYA_ADJ. Changing this voltage will change the duration of the analog delay. The current source is calibrated to charge C706 at a rate of 2 mV/ns.

DELAY RESET (sheet 8)

When both gate delays have finished, EOD_A and EOD_B will both be high. This triggers the one-shot 1/2 U806 to provide an 800 ns reset pulse. This pulse resets the trigger latch and switches the analog charging current from C706 back to the OTA. The OTA then discharges C706 back to the initial programmed voltage. Q801 provides an additional current kick to speed up this recharging process. The reset pulse also reloads the digital delay counters. Any trigger received during a delay or a delay reset is flagged as a rate error.

GATES (sheet 9)

The EOD_A signal clocks the width latch 1/2 U902. The width latch in turn clocks the gate latch 2/2 U902 thereby starting the A GATE. The width latch also starts the width delay. The width delay is similar to the gate delay described above. The A width delay has its own phase coherent oscillator which is started by the start of the respective gate. This allows both the delay and width to be active at the same time and allows a gate delay and width to cover nearly all of the trigger period simultaneously.

The EOW_A (end of width) signal resets the gate latch 2/2 U902 and terminates the A gate. The gate latch drives the A GATE output and enables counter A.

POWER SUPPLY

The power supply is a switchmode power supply operating directly from the rectified line voltage.

POTENTIALLY LETHAL VOLTAGES ARE PRESENT IN THIS CIRCUIT. DO NOT ATTEMPT TO SERVICE THE POWER SUPPLY. ALWAYS LEAVE THE COVERS ON WHEN OPERATING.

The power supply is over-Voltage, under-Voltage, and short circuit protected. and switching frequency is approximately 50 KHz. The fan control circuit and the LCD backlight power are also in the power supply enclosure.

Parts List

REF #	VALUE	DESCRIPTION	REF #	VALUE	DESCRIPTION
C 1652	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 701	47X3	Resistor Network, SIP, 1/4W,2% (SEP)
C 1653	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 702	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
C 1654	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 704	5X220	Resistor Network SIP 1/4W 2% Common
C 1655	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 705	5X100	Resistor Network SIP 1/4W 2% Common
C 1656	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 706	5X2.0K	Resistor Network SIP 1/4W 2% Common
C 1657	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 711	47X3	Resistor Network, SIP, 1/4W,2% (SEP)
C 1658	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 712	100X4	Resistor Network, SIP, 1/4W,2% (SEP)
C 1659	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 713	7X820	Resistor Network SIP 1/4W 2% Common
C 1660	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U	N 801	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
C 1670	2.2U	Capacitor, Tantalum, 35V, 20%, Rad	N 802	2.0KX5	Resistor Network, SIP, 1/4W,2% (SEP)
C 1671	2.2U	Capacitor, Tantalum, 35V, 20%, Rad	N 811	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
C 1672	2.2U	Capacitor, Tantalum, 35V, 20%, Rad	N 901	47X3	Resistor Network, SIP, 1/4W,2% (SEP)
C 1673	2.2U	Capacitor, Tantalum, 35V, 20%, Rad	N 902	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
C 1674	2.2U	Capacitor, Tantalum, 35V, 20%, Rad	N 904	5X220	Resistor Network SIP 1/4W 2% Common
D 101	MSD6100	Diode	N 905	7X100	Resistor Network SIP 1/4W 2% Common
D 102	5082-2800	Diode	N 906	7X100	Resistor Network SIP 1/4W 2% Common
D 103	5082-2800	Diode	N 907	9X100	Resistor Network SIP 1/4W 2% Common
D 104	1N5240	Diode	N 908	120X3	Resistor Network, SIP, 1/4W,2% (SEP)
D 121	MSD6100	Diode	N 911	47X3	Resistor Network, SIP, 1/4W,2% (SEP)
D 122	5082-2800	Diode	N 912	7X100	Resistor Network SIP 1/4W 2% Common
D 123	5082-2800	Diode	N 1001	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
D 124	1N5240	Diode	N 1002	2.0KX5	Resistor Network, SIP, 1/4W,2% (SEP)
D 130	MSD6100	Diode	N 1021	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
D 201	1N4148	Diode	N 1022	2.0KX5	Resistor Network, SIP, 1/4W,2% (SEP)
D 202	1N4148	Diode	N 1101	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)
D 701	1N4148	Diode	N 1201	5X20K	Resistor Network SIP 1/4W 2% Common
D 711	1N4148	Diode	N 1202	5X10K	Resistor Network SIP 1/4W 2% Common
D 801	1N4148	Diode	N 1301	7X22K	Resistor Network SIP 1/4W 2% Common
D 811	1N4148	Diode	N 1401	9X820	Resistor Network SIP 1/4W 2% Common
D 901	1N4148	Diode	N 1402	8X1.0K	Resistor Network, DIP, 1/4W,2%,8 ind
D 911	1N4148	Diode	N 1403	9X1.5K	Resistor Network SIP 1/4W 2% Common
D 1001	1N4148	Diode	N 1404	9X820	Resistor Network SIP 1/4W 2% Common
D 1021	1N4148	Diode	N 1405	8X1.0K	Resistor Network, DIP, 1/4W,2%,8 ind
D 1101	1N4148	Diode	N 1406	9X1.5K	Resistor Network SIP 1/4W 2% Common
D 1201	1N4148	Diode	N 1407	7X150	Resistor Network, DIP, 1/4W,2%,8 ind
D 1202	1N4148	Diode	N 1408	7X150	Resistor Network SIP 1/4W 2% Common
D 1203	1N4148	Diode	N 1501	10KX5	Resistor Network, SIP, 1/4W,2% (SEP)
D 1301	1N5231B	Diode	N 1502	20KX5	Resistor Network, SIP, 1/4W,2% (SEP)
D 1401	1N4148	Diode	N 1503	10KX4	Resistor Network, SIP, 1/4W,2% (SEP)
D 1402	1N4148	Diode	N 1504	10KX4	Resistor Network, SIP, 1/4W,2% (SEP)
DS1401	HLMP-0504	LED, Rectangular	P 101	200	Pot, Multi-Turn Trim
DS1402	HLMP-0504	LED, Rectangular	P 121	200	Pot, Multi-Turn Trim
DS1403	HLMP-0504	LED, Rectangular	P 1501	500	Pot, Multi-Turn Trim
DS1404	HLMP-0504	LED, Rectangular	PC1	SR400-29	Printed Circuit Board
DS1405	HLMP-0504	LED, Rectangular	PC2	SR400-41	Printed Circuit Board
DS1406	HLMP-0401	LED, Rectangular	Q 101	MRF904	Transistor, TO-72 Package
DS1407	HLMP-0504	LED, Rectangular	Q 102	MRF965	Transistor, TO-46 Package
DS1408	HLMP-0504	LED, Rectangular	Q 121	MRF904	Transistor, TO-72 Package
DS1409	HLMP-0504	LED, Rectangular	Q 122	MRF965	Transistor, TO-46 Package
DS1410	HLMP-0401	LED, Rectangular	Q 130	2N3906	Transistor, TO-92 Package
DS1411	HLMP-0504	LED, Rectangular	Q 301	2N5770	Transistor, TO-92 Package
J 1	RS232 F PC	Connector, D-Sub, Right Angle PC, Fema	Q 302	2N5770	Transistor, TO-92 Package
J 2	IEEE488	IEEE488, Reverse, PC Mount	Q 303	2N5770	Transistor, TO-92 Package
L 301	FB43-301	Ferrite Beads	Q 304	2N5770	Transistor, TO-92 Package
L 302	FB43-301	Ferrite Beads	Q 305	2N5770	Transistor, TO-92 Package
L 303	FB43-301	Ferrite Beads	Q 306	2N5770	Transistor, TO-92 Package
L 304	FB43-301	Ferrite Beads	Q 307	2N5770	Transistor, TO-92 Package
L 305	FB43-301	Ferrite Beads	Q 308	2N5770	Transistor, TO-92 Package
N 201	9X100	Resistor Network SIP 1/4W 2% Common	Q 309	2N5770	Transistor, TO-92 Package
N 202	10KX4	Resistor Network, SIP, 1/4W,2% (SEP)	Q 310	2N5770	Transistor, TO-92 Package
N 203	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)	Q 601	2N2222	Transistor, TO-18 Package
N 204	5X3.0K	Resistor Network SIP 1/4W 2% Common	Q 602	2N2222	Transistor, TO-18 Package
N 205	1.0KX4	Resistor Network, SIP, 1/4W,2% (SEP)	Q 701	2N3906	Transistor, TO-92 Package
N 301	7X100	Resistor Network SIP 1/4W 2% Common	Q 702	2N3906	Transistor, TO-92 Package
N 302	7X100	Resistor Network SIP 1/4W 2% Common	Q 703	2N3906	Transistor, TO-92 Package
N 303	7X100	Resistor Network SIP 1/4W 2% Common	Q 704	2N5770	Transistor, TO-92 Package
N 304	7X100	Resistor Network SIP 1/4W 2% Common	Q 705	2N5770	Transistor, TO-92 Package
N 305	5X2.0K	Resistor Network SIP 1/4W 2% Common	Q 711	2N3906	Transistor, TO-92 Package
N 306	5X100	Resistor Network SIP 1/4W 2% Common	Q 712	2N3906	Transistor, TO-92 Package
N 307	5X270	Resistor Network SIP 1/4W 2% Common	Q 713	2N3906	Transistor, TO-92 Package
N 308	5X100	Resistor Network SIP 1/4W 2% Common	Q 714	2N5770	Transistor, TO-92 Package
N 309	47X5	Resistor Network, SIP, 1/4W,2% (SEP)	Q 715	2N5770	Transistor, TO-92 Package
N 310	47X5	Resistor Network, SIP, 1/4W,2% (SEP)	Q 801	2N3904	Transistor, TO-92 Package
N 401	7X100	Resistor Network SIP 1/4W 2% Common	Q 802	2N3906	Transistor, TO-92 Package
N 402	5X100	Resistor Network SIP 1/4W 2% Common	Q 803	2N3906	Transistor, TO-92 Package
N 403	7X100	Resistor Network SIP 1/4W 2% Common	Q 804	2N3906	Transistor, TO-92 Package
N 601	10KX4	Resistor Network, SIP, 1/4W,2% (SEP)	Q 811	2N3906	Transistor, TO-92 Package
N 602	1.0KX3	Resistor Network, SIP, 1/4W,2% (SEP)	Q 901	2N3906	Transistor, TO-92 Package

REF#	VALUE	DESCRIPTION	REF#	VALUE	DESCRIPTION
Q 902	2N3906	Transistor, TO-92 Package	R 220	47	Resistor, Carbon Film, 1/4W, 5%
Q 903	2N3906	Transistor, TO-92 Package	R 221	10	Resistor, Carbon Film, 1/4W, 5%
Q 904	2N5770	Transistor, TO-92 Package	R 222	470	Resistor, Carbon Film, 1/4W, 5%
Q 905	2N5770	Transistor, TO-92 Package	R 223	47	Resistor, Carbon Film, 1/4W, 5%
Q 911	2N3906	Transistor, TO-92 Package	R 224	100	Resistor, Carbon Film, 1/4W, 5%
Q 912	2N3906	Transistor, TO-92 Package	R 225	100	Resistor, Carbon Film, 1/4W, 5%
Q 913	2N3906	Transistor, TO-92 Package	R 226	47	Resistor, Carbon Film, 1/4W, 5%
Q 914	2N5770	Transistor, TO-92 Package	R 227	10	Resistor, Carbon Film, 1/4W, 5%
Q 915	2N5770	Transistor, TO-92 Package	R 228	10	Resistor, Carbon Film, 1/4W, 5%
Q 1001	2N3904	Transistor, TO-92 Package	R 229	2.0K	Resistor, Carbon Film, 1/4W, 5%
Q 1002	2N3906	Transistor, TO-92 Package	R 230	47	Resistor, Carbon Film, 1/4W, 5%
Q 1003	2N3906	Transistor, TO-92 Package	R 231	100	Resistor, Carbon Film, 1/4W, 5%
Q 1004	2N3906	Transistor, TO-92 Package	R 232	10	Resistor, Carbon Film, 1/4W, 5%
Q 1021	2N3904	Transistor, TO-92 Package	R 233	10	Resistor, Carbon Film, 1/4W, 5%
Q 1022	2N3906	Transistor, TO-92 Package	R 234	10	Resistor, Carbon Film, 1/4W, 5%
Q 1023	2N3906	Transistor, TO-92 Package	R 235	10	Resistor, Carbon Film, 1/4W, 5%
Q 1024	2N3906	Transistor, TO-92 Package	R 401	2.0K	Resistor, Carbon Film, 1/4W, 5%
Q 1101	2N3906	Transistor, TO-92 Package	R 402	2.0K	Resistor, Carbon Film, 1/4W, 5%
Q 1201	2N5210	Transistor, TO-92 Package	R 403	820	Resistor, Carbon Film, 1/4W, 5%
Q 1202	2N5210	Transistor, TO-92 Package	R 404	470	Resistor, Carbon Film, 1/4W, 5%
Q 1203	2N5210	Transistor, TO-92 Package	R 405	390	Resistor, Carbon Film, 1/4W, 5%
R 101	51.1	Resistor, Metal Film, 1/8W, 1%	R 406	820	Resistor, Carbon Film, 1/4W, 5%
R 102	7.50K	Resistor, Metal Film, 1/8W, 1%	R 407	470	Resistor, Carbon Film, 1/4W, 5%
R 103	7.50K	Resistor, Metal Film, 1/8W, 1%	R 408	390	Resistor, Carbon Film, 1/4W, 5%
R 104	2.21K	Resistor, Metal Film, 1/8W, 1%	R 409	51	Resistor, Carbon Film, 1/4W, 5%
R 105	51K	Resistor, Carbon Film, 1/4W, 5%	R 501	10K	Resistor, Carbon Film, 1/4W, 5%
R 106	4.7K	Resistor, Carbon Film, 1/4W, 5%	R 502	1.0K	Resistor, Carbon Film, 1/4W, 5%
R 107	316	Resistor, Metal Film, 1/8W, 1%	R 601	51	Resistor, Carbon Comp, 1/2W, 5%
R 108	47.5	Resistor, Metal Film, 1/8W, 1%	R 602	51	Resistor, Carbon Comp, 1/2W, 5%
R 109	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 701	20.0K	Resistor, Metal Film, 1/8W, 1%
R 110	316	Resistor, Metal Film, 1/8W, 1%	R 702	4.99K	Resistor, Metal Film, 1/8W, 1%
R 111	510	Resistor, Carbon Comp, 1/2W, 5%	R 703	3.9K	Resistor, Carbon Film, 1/4W, 5%
R 112	10.0	Resistor, Metal Film, 1/8W, 1%	R 704	2.0K	Resistor, Carbon Film, 1/4W, 5%
R 113	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 705	100	Resistor, Carbon Film, 1/4W, 5%
R 114	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 706	47	Resistor, Carbon Film, 1/4W, 5%
R 115	4.99K	Resistor, Metal Film, 1/8W, 1%	R 707	402	Resistor, Metal Film, 1/8W, 1%
R 116	1.24K	Resistor, Metal Film, 1/8W, 1%	R 708	2.49K	Resistor, Metal Film, 1/8W, 1%
R 121	51.1	Resistor, Metal Film, 1/8W, 1%	R 709	100	Resistor, Carbon Film, 1/4W, 5%
R 122	7.50K	Resistor, Metal Film, 1/8W, 1%	R 710	47	Resistor, Carbon Film, 1/4W, 5%
R 123	7.50K	Resistor, Metal Film, 1/8W, 1%	R 711	20.0K	Resistor, Metal Film, 1/8W, 1%
R 124	2.21K	Resistor, Metal Film, 1/8W, 1%	R 712	4.99K	Resistor, Metal Film, 1/8W, 1%
R 125	51K	Resistor, Carbon Film, 1/4W, 5%	R 713	3.9K	Resistor, Carbon Film, 1/4W, 5%
R 126	4.7K	Resistor, Carbon Film, 1/4W, 5%	R 714	2.0K	Resistor, Carbon Film, 1/4W, 5%
R 127	316	Resistor, Metal Film, 1/8W, 1%	R 715	100	Resistor, Carbon Film, 1/4W, 5%
R 128	47.5	Resistor, Metal Film, 1/8W, 1%	R 716	47	Resistor, Carbon Film, 1/4W, 5%
R 129	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 717	402	Resistor, Metal Film, 1/8W, 1%
R 130	316	Resistor, Metal Film, 1/8W, 1%	R 718	2.49K	Resistor, Metal Film, 1/8W, 1%
R 131	510	Resistor, Carbon Comp, 1/2W, 5%	R 719	100	Resistor, Carbon Film, 1/4W, 5%
R 132	10.0	Resistor, Metal Film, 1/8W, 1%	R 720	47	Resistor, Carbon Film, 1/4W, 5%
R 133	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 721	390	Resistor, Carbon Film, 1/4W, 5%
R 134	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 722	470	Resistor, Carbon Film, 1/4W, 5%
R 135	4.99K	Resistor, Metal Film, 1/8W, 1%	R 724	390	Resistor, Carbon Film, 1/4W, 5%
R 136	1.24K	Resistor, Metal Film, 1/8W, 1%	R 725	470	Resistor, Carbon Film, 1/4W, 5%
R 140	3.74K	Resistor, Metal Film, 1/8W, 1%	R 727	1.6K	Resistor, Carbon Film, 1/4W, 5%
R 141	2.32K	Resistor, Metal Film, 1/8W, 1%	R 728	1.1K	Resistor, Carbon Film, 1/4W, 5%
R 142	2.4K	Resistor, Carbon Film, 1/4W, 5%	R 729	820	Resistor, Carbon Film, 1/4W, 5%
R 143	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 730	1.0K	Resistor, Carbon Film, 1/4W, 5%
R 144	10	Resistor, Carbon Film, 1/4W, 5%	R 731	1.0K	Resistor, Carbon Film, 1/4W, 5%
R 145	10	Resistor, Carbon Film, 1/4W, 5%	R 733	200	Resistor, Carbon Film, 1/4W, 5%
R 201	100	Resistor, Carbon Film, 1/4W, 5%	R 734	200	Resistor, Carbon Film, 1/4W, 5%
R 202	47	Resistor, Carbon Film, 1/4W, 5%	R 801	1.0K	Resistor, Carbon Film, 1/4W, 5%
R 203	10	Resistor, Carbon Film, 1/4W, 5%	R 802	120	Resistor, Carbon Film, 1/4W, 5%
R 204	2.0K	Resistor, Carbon Film, 1/4W, 5%	R 803	7.50K	Resistor, Metal Film, 1/8W, 1%
R 205	47	Resistor, Carbon Film, 1/4W, 5%	R 804	47	Resistor, Carbon Film, 1/4W, 5%
R 206	100	Resistor, Carbon Film, 1/4W, 5%	R 805	100	Resistor, Carbon Film, 1/4W, 5%
R 207	47	Resistor, Carbon Film, 1/4W, 5%	R 806	1.0K	Resistor, Carbon Film, 1/4W, 5%
R 208	10	Resistor, Carbon Film, 1/4W, 5%	R 807	4.3K	Resistor, Carbon Film, 1/4W, 5%
R 209	2.0K	Resistor, Carbon Film, 1/4W, 5%	R 808	1.0K	Resistor, Carbon Film, 1/4W, 5%
R 210	47	Resistor, Carbon Film, 1/4W, 5%	R 809	4.3K	Resistor, Carbon Film, 1/4W, 5%
R 211	100	Resistor, Carbon Film, 1/4W, 5%	R 810	909	Resistor, Metal Film, 1/8W, 1%
R 212	100	Resistor, Carbon Film, 1/4W, 5%	R 811	10	Resistor, Carbon Film, 1/4W, 5%
R 213	47	Resistor, Carbon Film, 1/4W, 5%	R 812	100	Resistor, Carbon Film, 1/4W, 5%
R 214	10	Resistor, Carbon Film, 1/4W, 5%	R 901	20.0K	Resistor, Metal Film, 1/8W, 1%
R 215	2.0K	Resistor, Carbon Film, 1/4W, 5%	R 902	4.99K	Resistor, Metal Film, 1/8W, 1%
R 216	47	Resistor, Carbon Film, 1/4W, 5%	R 903	3.9K	Resistor, Carbon Film, 1/4W, 5%
R 217	100	Resistor, Carbon Film, 1/4W, 5%	R 904	2.0K	Resistor, Carbon Film, 1/4W, 5%
R 218	10K	Resistor, Carbon Film, 1/4W, 5%	R 905	100	Resistor, Carbon Film, 1/4W, 5%
R 219	1.0K	Resistor, Carbon Film, 1/4W, 5%	R 906	47	Resistor, Carbon Film, 1/4W, 5%

Parts List

REF #	VALUE	DESCRIPTION	REF #	VALUE	DESCRIPTION
R 907	402	Resistor, Metal Film, 1/8W, 1%	U 204	AD9685BH	Integrated Circuit
R 908	2.49K	Resistor, Metal Film, 1/8W, 1%	U 206	MC10H113	Integrated Circuit
R 909	100	Resistor, Carbon Film, 1/4W, 5%	U 207	MPQ2222	Integrated Circuit
R 910	47	Resistor, Carbon Film, 1/4W, 5%	U 208	LF347	Integrated Circuit
R 911	20.0K	Resistor, Metal Film, 1/8W, 1%	U 209	AD9685BH	Integrated Circuit
R 912	4.99K	Resistor, Metal Film, 1/8W, 1%	U 210	MC10H102	Integrated Circuit
R 913	3.9K	Resistor, Carbon Film, 1/4W, 5%	U 301	MC10H102	Integrated Circuit
R 914	2.0K	Resistor, Carbon Film, 1/4W, 5%	U 302	MC10H131	Integrated Circuit
R 915	100	Resistor, Carbon Film, 1/4W, 5%	U 303	MC10131	Integrated Circuit
R 916	47	Resistor, Carbon Film, 1/4W, 5%	U 304	74F191	Integrated Circuit
R 917	402	Resistor, Metal Film, 1/8W, 1%	U 305	MC10H105	Integrated Circuit
R 918	2.49K	Resistor, Metal Film, 1/8W, 1%	U 306	MC10H131	Integrated Circuit
R 919	100	Resistor, Carbon Film, 1/4W, 5%	U 307	MC10125	Integrated Circuit
R 920	47	Resistor, Carbon Film, 1/4W, 5%	U 308	74F191	Integrated Circuit
R 921	390	Resistor, Carbon Film, 1/4W, 5%	U 401	MC10H102	Integrated Circuit
R 922	470	Resistor, Carbon Film, 1/4W, 5%	U 402	MC10H131	Integrated Circuit
R 924	390	Resistor, Carbon Film, 1/4W, 5%	U 403	MC10131	Integrated Circuit
R 925	470	Resistor, Carbon Film, 1/4W, 5%	U 404	74F191	Integrated Circuit
R 927	390	Resistor, Carbon Film, 1/4W, 5%	U 405	MC10H131	Integrated Circuit
R 928	430	Resistor, Carbon Film, 1/4W, 5%	U 406	MC10H109	Integrated Circuit
R 930	390	Resistor, Carbon Film, 1/4W, 5%	U 407	MC10124	Integrated Circuit
R 931	430	Resistor, Carbon Film, 1/4W, 5%	U 408	MC10125	Integrated Circuit
R 940	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 501	74HC74	Integrated Circuit
R 941	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 502	74HC32	Integrated Circuit
R 942	200	Resistor, Carbon Film, 1/4W, 5%	U 503	74HC4075	Integrated Circuit
R 943	200	Resistor, Carbon Film, 1/4W, 5%	U 504	74HC14	Integrated Circuit
R 1001	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 601	74HC02	Integrated Circuit
R 1002	120	Resistor, Carbon Film, 1/4W, 5%	U 602	74HC74	Integrated Circuit
R 1003	7.50K	Resistor, Metal Film, 1/8W, 1%	U 603	74HC74	Integrated Circuit
R 1004	47	Resistor, Carbon Film, 1/4W, 5%	U 604	74HC74	Integrated Circuit
R 1005	100	Resistor, Carbon Film, 1/4W, 5%	U 605	20.000MHZ	Crystal
R 1006	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 701	MC10H105	Integrated Circuit
R 1007	4.3K	Resistor, Carbon Film, 1/4W, 5%	U 702	MC10H115	Integrated Circuit
R 1008	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 703	MC10125	Integrated Circuit
R 1009	4.3K	Resistor, Carbon Film, 1/4W, 5%	U 704	LM13600	Integrated Circuit
R 1010	909	Resistor, Metal Film, 1/8W, 1%	U 705	LF347	Integrated Circuit
R 1011	10	Resistor, Carbon Film, 1/4W, 5%	U 801	74HC04	Integrated Circuit
R 1012	100	Resistor, Carbon Film, 1/4W, 5%	U 802	LF353	Integrated Circuit
R 1021	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 803	LT1016	Integrated Circuit
R 1022	120	Resistor, Carbon Film, 1/4W, 5%	U 804	74HC74	Integrated Circuit
R 1023	7.50K	Resistor, Metal Film, 1/8W, 1%	U 805	74HC32	Integrated Circuit
R 1024	47	Resistor, Carbon Film, 1/4W, 5%	U 806	74HC4538	Integrated Circuit
R 1025	100	Resistor, Carbon Film, 1/4W, 5%	U 807	74HC00	Integrated Circuit
R 1026	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 901	MC10124	Integrated Circuit
R 1027	4.3K	Resistor, Carbon Film, 1/4W, 5%	U 902	MC10H131	Integrated Circuit
R 1028	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 903	MC10H131	Integrated Circuit
R 1029	4.3K	Resistor, Carbon Film, 1/4W, 5%	U 904	MC10H105	Integrated Circuit
R 1030	909	Resistor, Metal Film, 1/8W, 1%	U 905	MC10125	Integrated Circuit
R 1031	10	Resistor, Carbon Film, 1/4W, 5%	U 906	LM13600	Integrated Circuit
R 1032	100	Resistor, Carbon Film, 1/4W, 5%	U 907	LF347	Integrated Circuit
R 1105	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 908	MC10101	Integrated Circuit
R 1106	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 1001	74HC04	Integrated Circuit
R 1201	10K	Resistor, Carbon Film, 1/4W, 5%	U 1002	LF353	Integrated Circuit
R 1202	100K	Resistor, Carbon Film, 1/4W, 5%	U 1003	LT1016	Integrated Circuit
R 1203	10K	Resistor, Carbon Film, 1/4W, 5%	U 1004	74HC04	Integrated Circuit
R 1204	200K	Resistor, Carbon Film, 1/4W, 5%	U 1005	LF353	Integrated Circuit
R 1205	3.3K	Resistor, Carbon Film, 1/4W, 5%	U 1006	LT1016	Integrated Circuit
R 1206	3.3K	Resistor, Carbon Film, 1/4W, 5%	U 1101	74F74	Integrated Circuit
R 1207	4.7K	Resistor, Carbon Film, 1/4W, 5%	U 1102	74HC74	Integrated Circuit
R 1208	1.5K	Resistor, Carbon Film, 1/4W, 5%	U 1103	74HC74	Integrated Circuit
R 1301	2.2K	Resistor, Carbon Film, 1/4W, 5%	U 1104	74HC32	Integrated Circuit
R 1302	1.5K	Resistor, Carbon Film, 1/4W, 5%	U 1105	74HC00	Integrated Circuit
R 1401	1.0K	Resistor, Carbon Film, 1/4W, 5%	U 1106	74HC04	Integrated Circuit
R 1501	19.6K	Resistor, Metal Film, 1/8W, 1%	U 1201	74HC154	Integrated Circuit
R 1502	4.02K	Resistor, Metal Film, 1/8W, 1%	U 1202	74HC14	Integrated Circuit
R 1503	4.99K	Resistor, Metal Film, 1/8W, 1%	U 1203	74HC32	Integrated Circuit
R 1504	8.06K	Resistor, Metal Film, 1/8W, 1%	U 1204	74HC04	Integrated Circuit
R 1505	200	Resistor, Carbon Film, 1/4W, 5%	U 1205	MAX232	Integrated Circuit
R 1506	200	Resistor, Carbon Film, 1/4W, 5%	U 1206	Z8800	Integrated Circuit
R 1507	10.0K	Resistor, Metal Film, 1/8W, 1%	U 1207	74HC374	Integrated Circuit
R 1508	10.0K	Resistor, Metal Film, 1/8W, 1%	U 1208	27256-200	Integrated Circuit
R 1509	200	Resistor, Carbon Film, 1/4W, 5%	U 1209	62256LP-15	Integrated Circuit
U 101	LT1007	Integrated Circuit	U 1211	74HC138	Integrated Circuit
U 102	LT1007	Integrated Circuit	U 1301	74HC374	Integrated Circuit
U 103	LT1007	Integrated Circuit	U 1302	74HC374	Integrated Circuit
U 201	AD9685BH	Integrated Circuit	U 1303	74HC244	Integrated Circuit
U 202	AD9685BH	Integrated Circuit	U 1304	UPD8253-5	Integrated Circuit
U 203	AD9685BH	Integrated Circuit	U 1305	UPD8253-5	Integrated Circuit

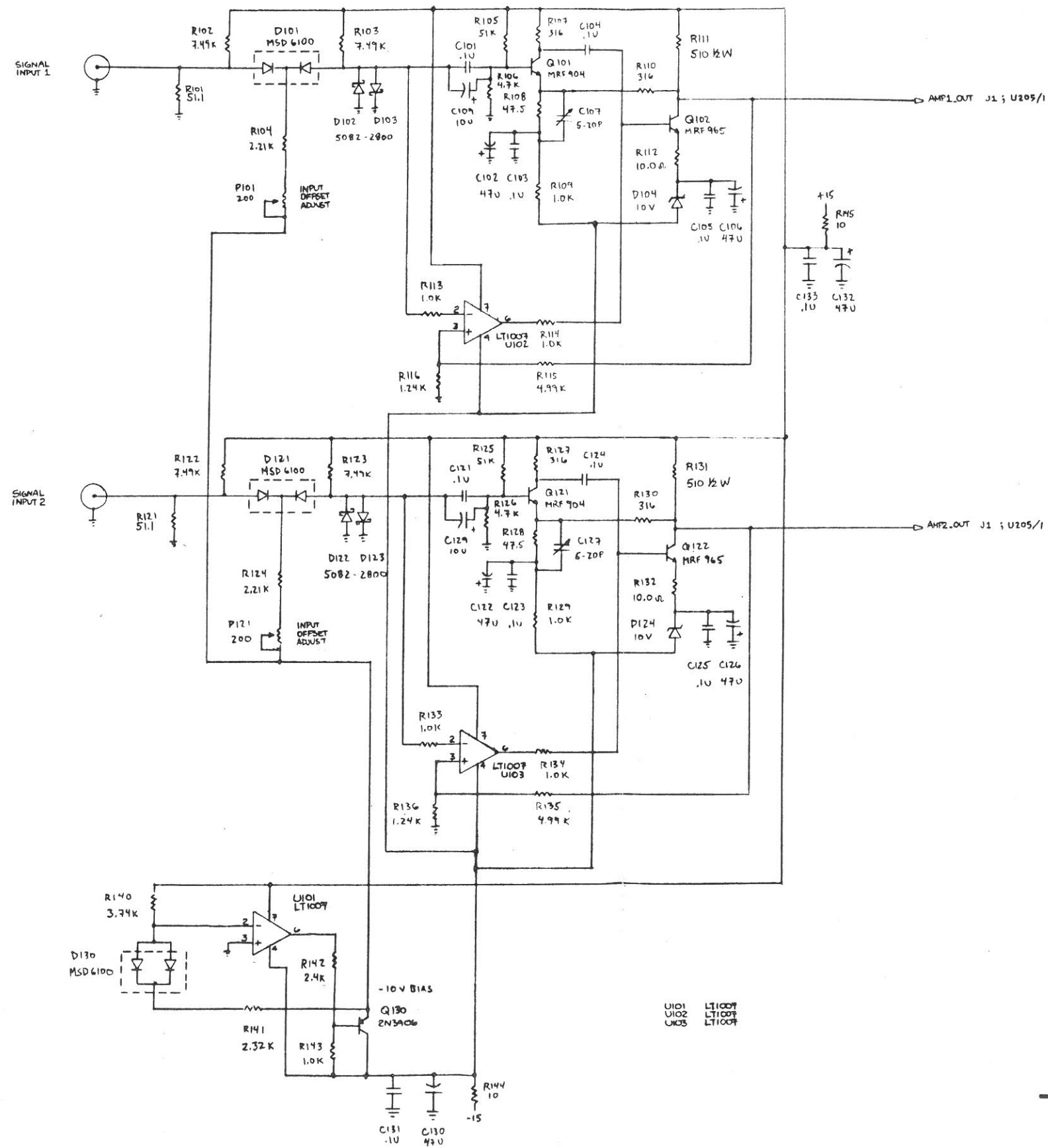
POWER SUPPLY

REF#	VALUE	DESCRIPTION
U 1306	UPD8253-5	Integrated Circuit
U 1307	UPD8253-5	Integrated Circuit
U 1308	UPD8253-5	Integrated Circuit
U 1309	DS75160A	Integrated Circuit
U 1310	DS75161A	Integrated Circuit
U 1311	TMS9914A	Integrated Circuit
U 1401	74HC374	Integrated Circuit
U 1402	74HC374	Integrated Circuit
U 1403	74HC374	Integrated Circuit
U 1404	74HC374	Integrated Circuit
U 1405	74HC374	Integrated Circuit
U 1406	74HC374	Integrated Circuit
U 1407	74HC374	Integrated Circuit
U 1501	DG211	Integrated Circuit
U 1502	AD7542JN	Integrated Circuit
U 1503	DG528	Integrated Circuit
U 1504	LF347	Integrated Circuit
U 1505	LH0070-0H	Integrated Circuit
U 1506	DG528	Integrated Circuit
U 1507	LF347	Integrated Circuit
U 1508	LF347	Integrated Circuit
Z0	M/F 3/16"	Standoff
Z0	M/F 5/16"	Standoff
Z0	#4 SPLIT	Washer, Split
Z0	1-1/2" #24	Wire #24
Z0	3-1/2" #24	Wire #24
Z0	4-40X1/4PF	Screw, Flathead Phillips
Z0	4-40X1/4PF	Screw, Black, All Types
Z0	TO-5	Heat Sinks
Z0	TO-18 HAT	Heat Sinks
Z0	TO-18	Insulators
Z0	6-32X1/2RP	Screw, Roundhead Phillips
Z0	6-32X5/16R	Screw, Roundhead Phillips
Z0	RIGHT FOOT	Hardware, Misc.
Z0	LEFT FOOT	Hardware, Misc.
Z0	6-32X1/4PF	Screw, Flathead Phillips
Z0	4-40X1/4PP	Screw, Panhead Phillips
Z0	4-40X3/8PP	Screw, Panhead Phillips
Z0	6-32X1/4PP	Screw, Panhead Phillips
Z0	HANDLE1	Hardware, Misc.
Z0	CLEAR 3	Window
Z0	KN1751B-1/4	Knobs
Z0	F1404	Knobs
Z0	6-32X1/4PF	Screw, Black, All Types
Z0	10-32X1/2PF	Screw, Black, All Types
Z0	4-40X1/8PP	Screw, Panhead Phillips
Z0	8-32X1/4PF	Screw, Black, All Types
Z0	BNC	Connector, BNC, Panel Mount
Z0	INS	Connector, BNC, Panel Mount
Z0	24	Socket, DIP
Z0	28	Socket, DIP
Z0	MINI 7 DI	Connector, Male
Z0	MINI 7 DIF	Connector, Female
Z0	4 PIN DI	Connector, Male
Z0	4 PIN DIF	Connector, Female
Z0	8 PIN DI	Connector, Male
Z0	8 PIN DIF	Connector, Female
Z0	SP-16	SOFTPOT
Z0	SR400-31	Front Panel
Z0	SR400-41	Printed Circuit Board
Z0	SR400-20	Fabricated Part
Z0	SR400-22	Fabricated Part
Z0	SR400-26	Fabricated Part
Z0	SR400-23	Fabricated Part
Z0	SR400-24	Fabricated Part
Z0	SR400-36	Fabricated Part
Z0	SR400-39	Fabricated Part
Z0	SR400-37	Fabricated Part
Z0	SR400-40	KEYPAD, CONDUCTIVE RUBBER
Z0	LM306B2C24	LCD Display

REF#	VALUE	DESCRIPTION
C 1601	470U	Capacitor, Misc.
C 1602	470U	Capacitor, Misc.
C 1603	100U	Capacitor, Electrolytic, 25V, 20%, Rad
C 1604	.22U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1605	.001U	Polypropylene Capacitor
C 1606	15P	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1607	330P	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1608	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1609	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1610	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1611	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1612	3300P	CERAMIC DISC TYPE DD, 1000 VDCW
C 1613	3300U	CAPACITOR, LINE VOLT., 6.3V
C 1614	15U	Capacitor, Tantalum, 35V, 20%, Rad
C 1615	3300U	CAPACITOR, LINE VOLT., 6.3V
C 1616	15U	Capacitor, Tantalum, 35V, 20%, Rad
C 1617	1000U	CAPACITOR, LINE VOLT., 6.3V
C 1618	15U	Capacitor, Tantalum, 35V, 20%, Rad
C 1619	470U	CAP. LINE VOLTAGE, 20V
C 1620	10U	Capacitor, Tantalum, 35V, 20%, Rad
C 1621	470U	CAP. LINE VOLTAGE, 20V
C 1622	10U	Capacitor, Tantalum, 35V, 20%, Rad
C 1623	2.2U	Capacitor, Tantalum, 35V, 20%, Rad
C 1624	2.2U	Capacitor, Tantalum, 35V, 20%, Rad
C 1625	.1U	Capacitor, Metallized Polyester
C 1626	1200P	CERAMIC DISC TYPE DD, 1000 VDCW
C 1627	.001U	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1628	.001U	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1629	.001U	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1630	.001U	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1631	.001U	Capacitor, Ceramic Disc, 50V, 10%, SL
C 1632	100U	Capacitor, Electrolytic, 25V, 20%, Rad
C 1633	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1634	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1635	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
C 1636	.1U	Capacitor, Ceramic Disc, 50V, 20%, Z5U
D 1601	MDA106G	Diode
D 1602	1N4004	Diode
D 1603	MUR140	Diode
D 1604	MUR110	Diode
D 1605	1N5821	Diode
D 1606	1N5821	Diode

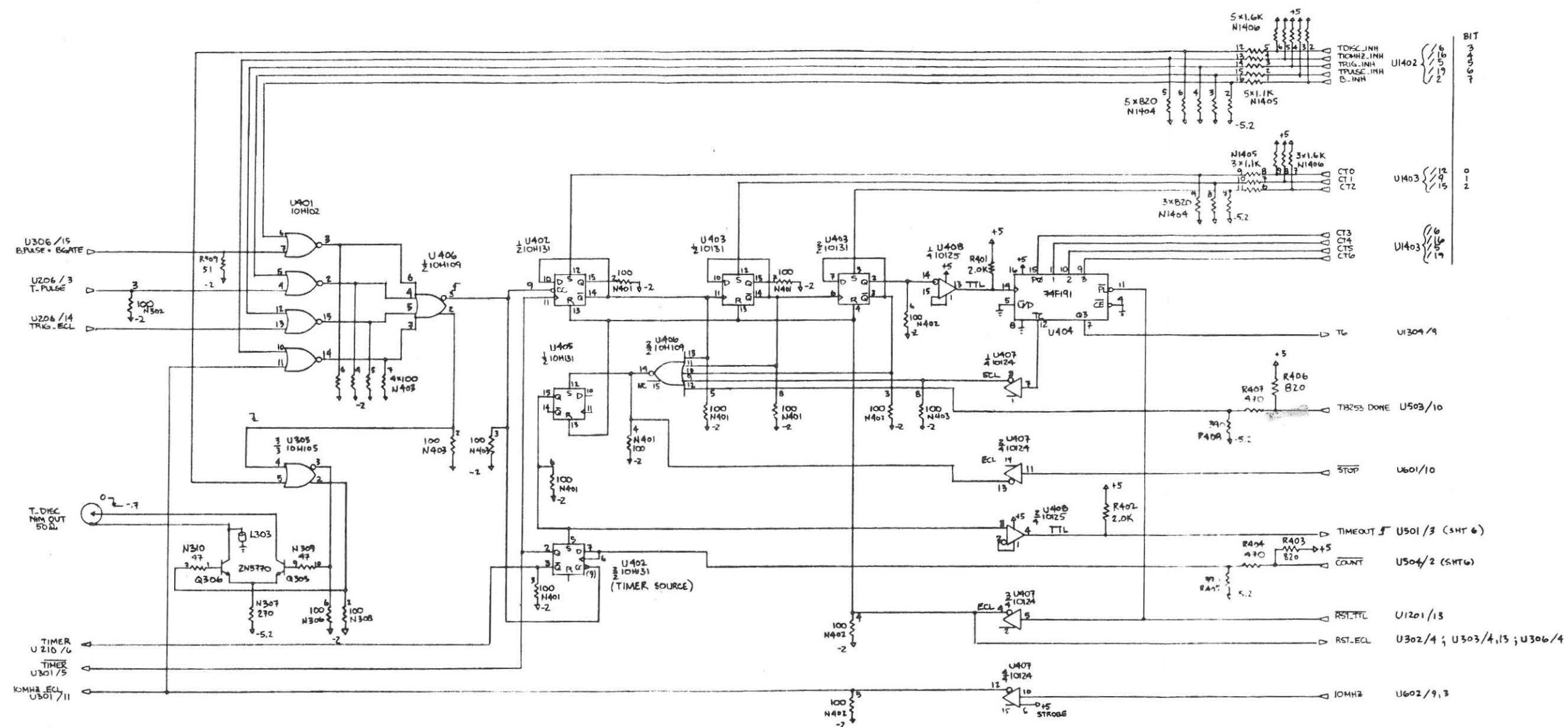
Parts List

REF#	VALUE	DESCRIPTION	REF#	VALUE	DESCRIPTION
D 1607	1N5820	Diode	R 1615	13.0K	Resistor, Metal Film, 1/8W, 1%
D 1608	MUR110	Diode	R 1616	1.0M	Resistor, Carbon Film, 1/4W, 5%
D 1609	MUR110	Diode	R 1617	10K	Resistor, Carbon Film, 1/4W, 5%
D 1610	1N4148	Diode	R 1618	21.5K	Resistor, Metal Film, 1/8W, 1%
L 1606	10MH	Inductor	R 1619	1.3K	Resistor, Carbon Film, 1/4W, 5%
LR1601	4.5UH	Inductor, Radial	R 1620	2.0K	Resistor, Carbon Film, 1/4W, 5%
LR1602	7.0UH	Inductor, Radial	R 1621	20K	Resistor, Carbon Film, 1/4W, 5%
LR1603	7.0UH	Inductor, Radial	R 1622	5.1K	Resistor, Carbon Film, 1/4W, 5%
LR1604	10UH	Inductor, Radial	R 1623	1.3K	Resistor, Carbon Film, 1/4W, 5%
LR1605	10UH	Inductor, Radial	R 1624	1.0	Resistor, Carbon Comp, 1/2W, 5%
P 1601	1.0K	Trim Pot, PC Mount, In-Line Leads	R 1627	2.0K	Resistor, Carbon Film, 1/4W, 5%
Q 1601	MPSA44	Transistor, TO-92 Package	R 1628	47	Resistor, Carbon Comp, 1/2W, 5%
Q 1602	MPSA44	Transistor, TO-92 Package	R 1629	47	Resistor, Carbon Comp, 1/2W, 5%
Q 1603	MTP2N60	Integrated Circuit	R 1630	47	Resistor, Carbon Comp, 1/2W, 5%
Q 1604	2N3906	Transistor, TO-92 Package	R 1631	47	Resistor, Carbon Comp, 1/2W, 5%
Q 1605	TIP31B	Transistor, TO-220 (TAB) Package	R 1632	47	Resistor, Carbon Comp, 1/2W, 5%
R 1601	1.0	Resistor, Carbon Comp, 1W, 10%	R 1633	50-30-25	Thermistor, PTC
R 1602	1.0	Resistor, Carbon Comp, 1W, 10%	R 1634	2.0K	Resistor, Carbon Film, 1/4W, 5%
R 1603	30K	Resistor, Wire Wound	R 1635	2.0K	Resistor, Carbon Film, 1/4W, 5%
R 1604	1.0M	Resistor, Carbon Comp, 1/2W, 5%	R 1636	2.0K	Resistor, Carbon Film, 1/4W, 5%
R 1605	1.0	Resistor, Carbon Comp, 1/2W, 5%	R 1637	68K	Resistor, Carbon Film, 1/4W, 5%
R 1606	1.0M	Resistor, Carbon Comp, 1/2W, 5%	R 1638	33K	Resistor, Carbon Film, 1/4W, 5%
R 1607	1.5M	Resistor, Carbon Comp, 1/2W, 5%	R 1639	200	Resistor, Carbon Comp, 1/2W, 5%
R 1608	1.0M	Resistor, Carbon Comp, 1/2W, 5%	S1601	DPDT	Switch, Panel Mount, Power, Push-button
R 1609	20K	Resistor, Carbon Comp, 1W, 10%	T 1601	ES-5861	Transformer
R 1610	1.0M	Resistor, Carbon Comp, 1/2W, 5%	T 1602	PL11-03	Transformer
R 1611	7.5K	Resistor, Carbon Film, 1/4W, 5%	U 1601	UC3840	Integrated Circuit
R 1612	20K	Resistor, Carbon Film, 1/4W, 5%	U 1602	LM340T-15	Integrated Circuit
R 1613	6.2K	Resistor, Carbon Film, 1/4W, 5%	U 1603	LM320T-15	Integrated Circuit
R 1614	34.8K	Resistor, Metal Film, 1/8W, 1%	U 1604	LM555	Integrated Circuit



STANFORD RESEARCH SYSTEMS		SIGNAL AMPLIFIERS			
DRAWN	SEPT 18, 1986	SIZE	FSCM NO.	DWG. NO.	REV.
ISSUED		SCALE		SHEET	A
BISHOP GRAPHICS, INC. REORDER NO. 1860					

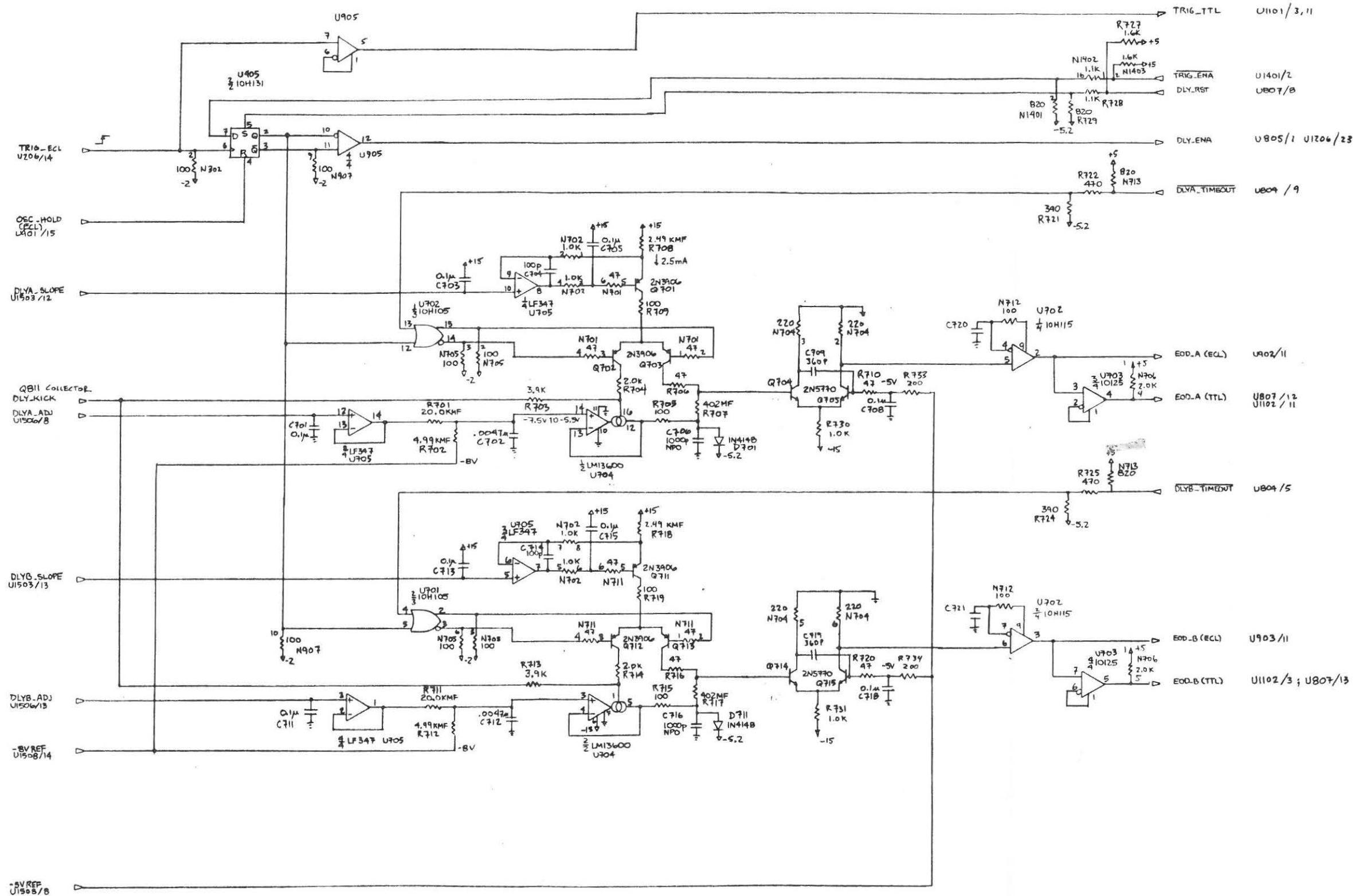




U401 MC10H102
 U402 MC10H101
 U403 MC10H101
 U404 74F1
 U405 MC10H101
 U406 MC10H102
 U407 MC10H101
 U408 MC10H102
 U305 MC10H105 4/4

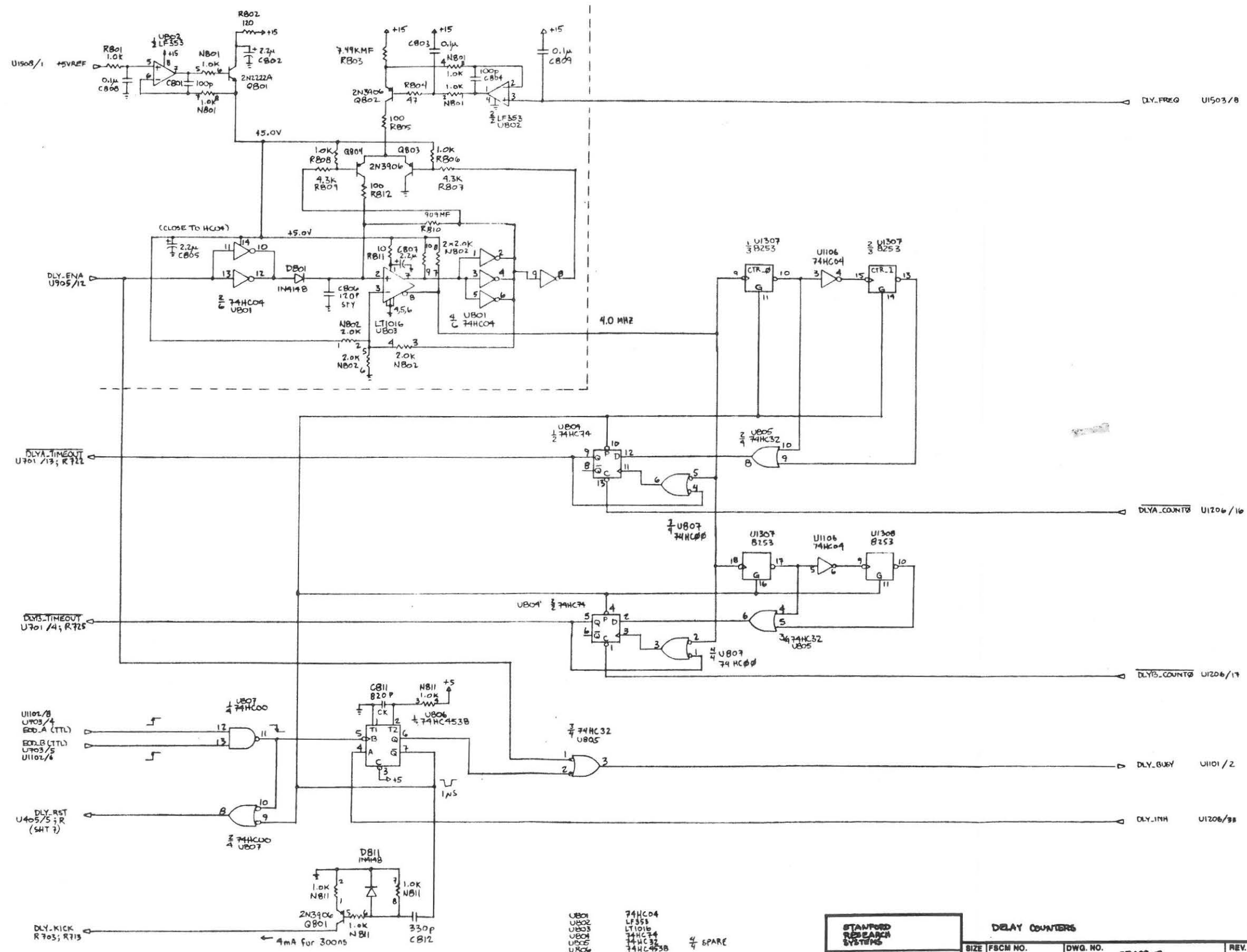
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BISHOP GRAPHICS, INC.
 REORDER NO. 1860



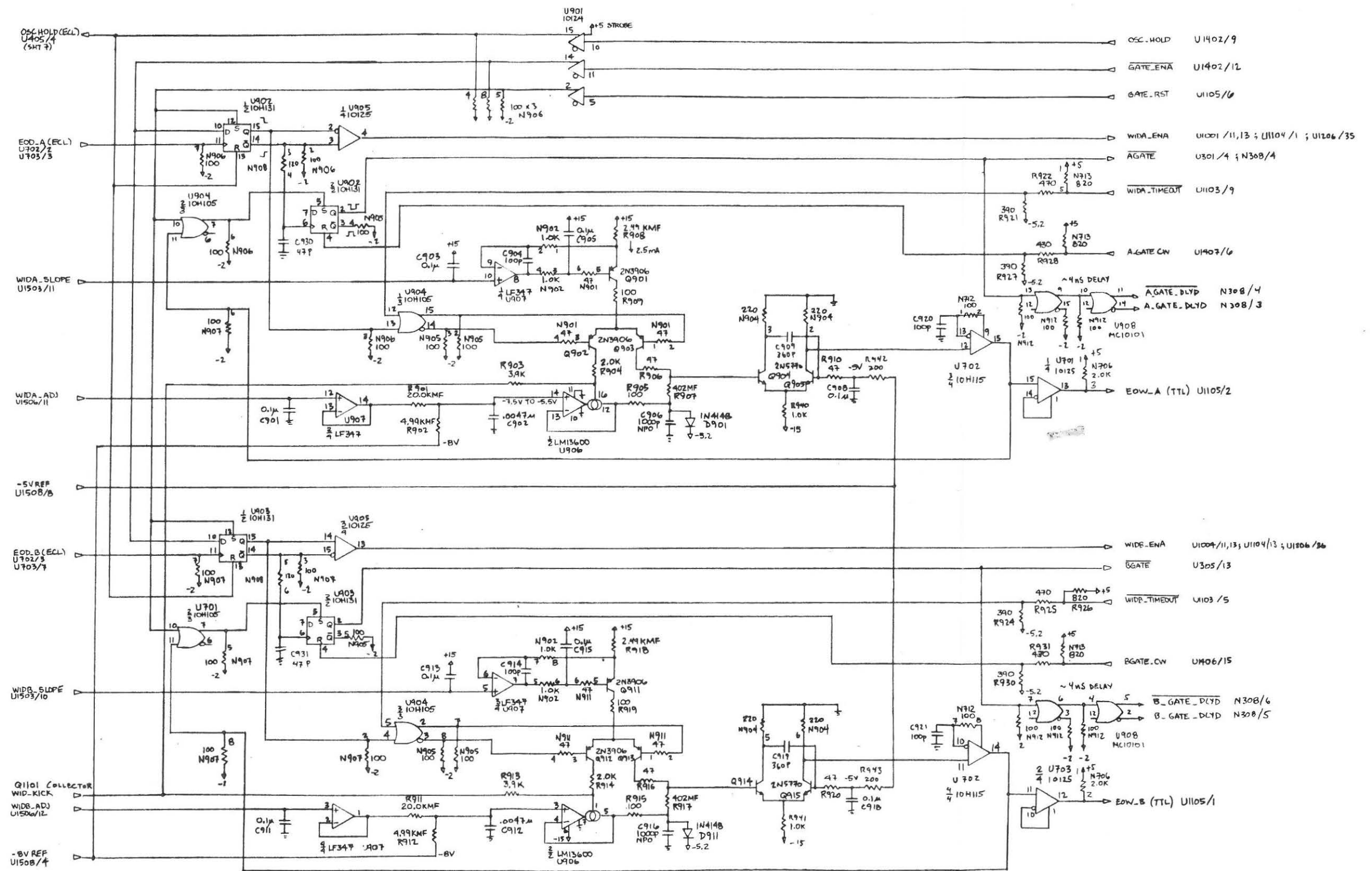
U701 MC10H105
 U702 MC10H115
 U703 MC10125
 U704 LM13600
 U705 LF347
 U405 MC10H131
 U905 MC10125

STANDARD REVISION SYSTEMS		DELAY RAMPS		
DRAWN 22 AUG 1986	SIZE	FSCM NO.	DWG. NO. 68400-7	REV. A
ISSUED	SCALE	DATE	SHEET	



STANFORD RESEARCH SYSTEMS		DELAY COUNTERS			
DRAWN	21 AUG 1986	SIZE	FSCM NO.	DWG. NO.	REV.
ISSUED		SCALE	JCW	SHEET	A

BISHOP GRAPHICS, INC.
REORDER NO. 1860

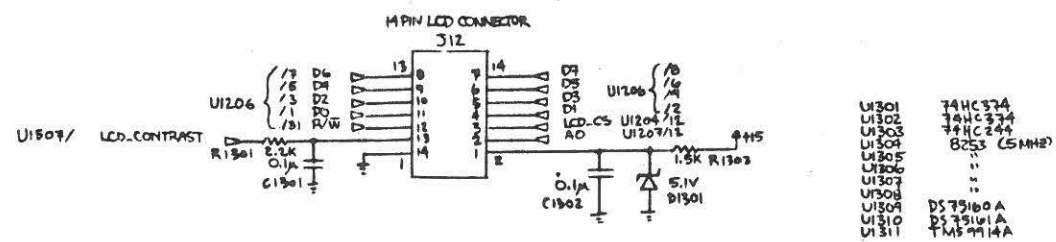
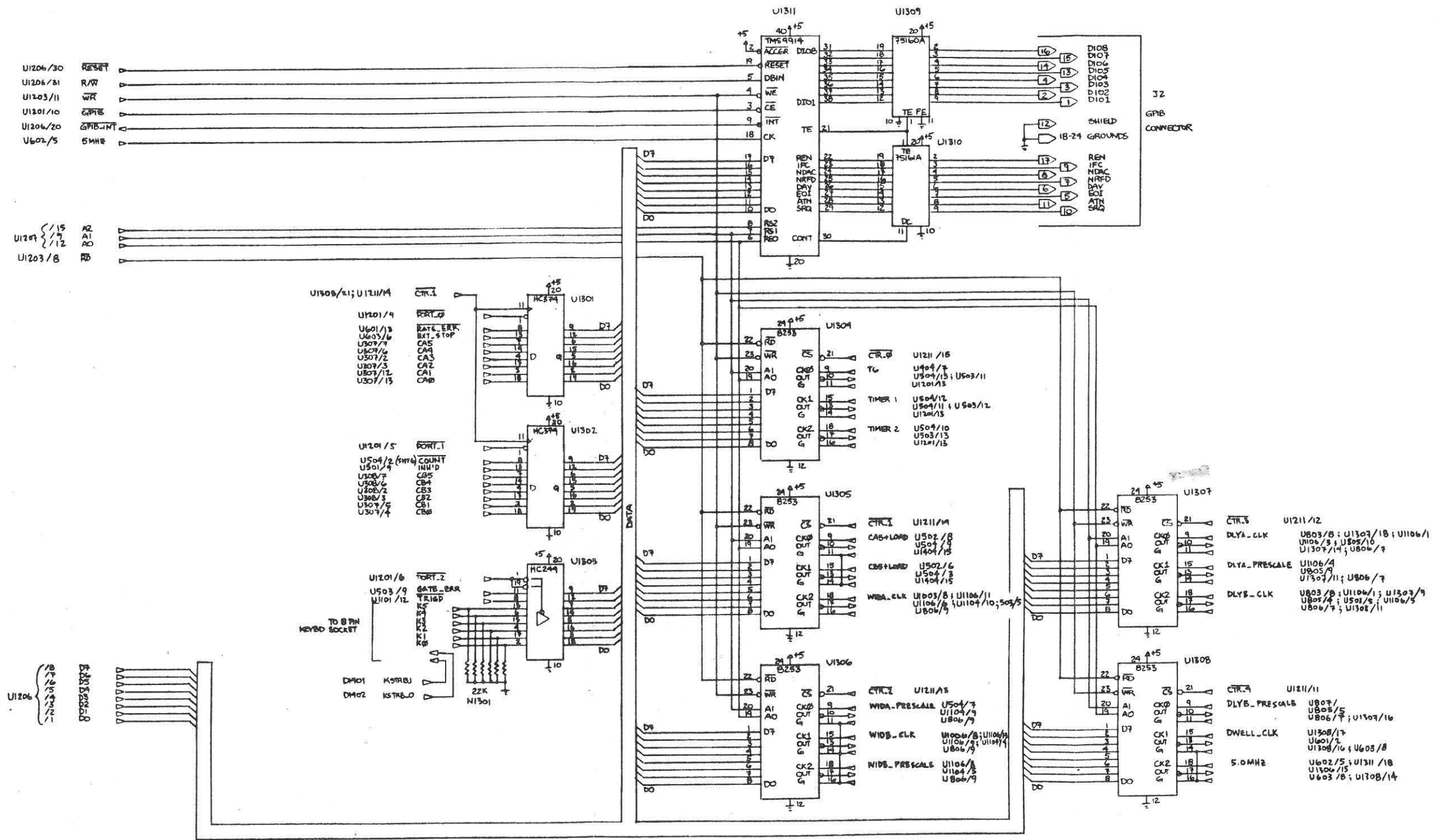


U901 MC10124
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STANFORD		GATE RAMPS		REV.	
SIZE	FSCM NO.	DWG. NO.	SR400-9	A	
ISSUED	SCALE	JUN	SHEET		

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STANDARD SYSTEMS		DIGITAL INTERFACE 1	
DRAWN	SEPT 15, 1986	SIZE	FSCM NO.
ISSUED	SCALE	DWG. NO.	64100-13
SHEET		REV. A	

